

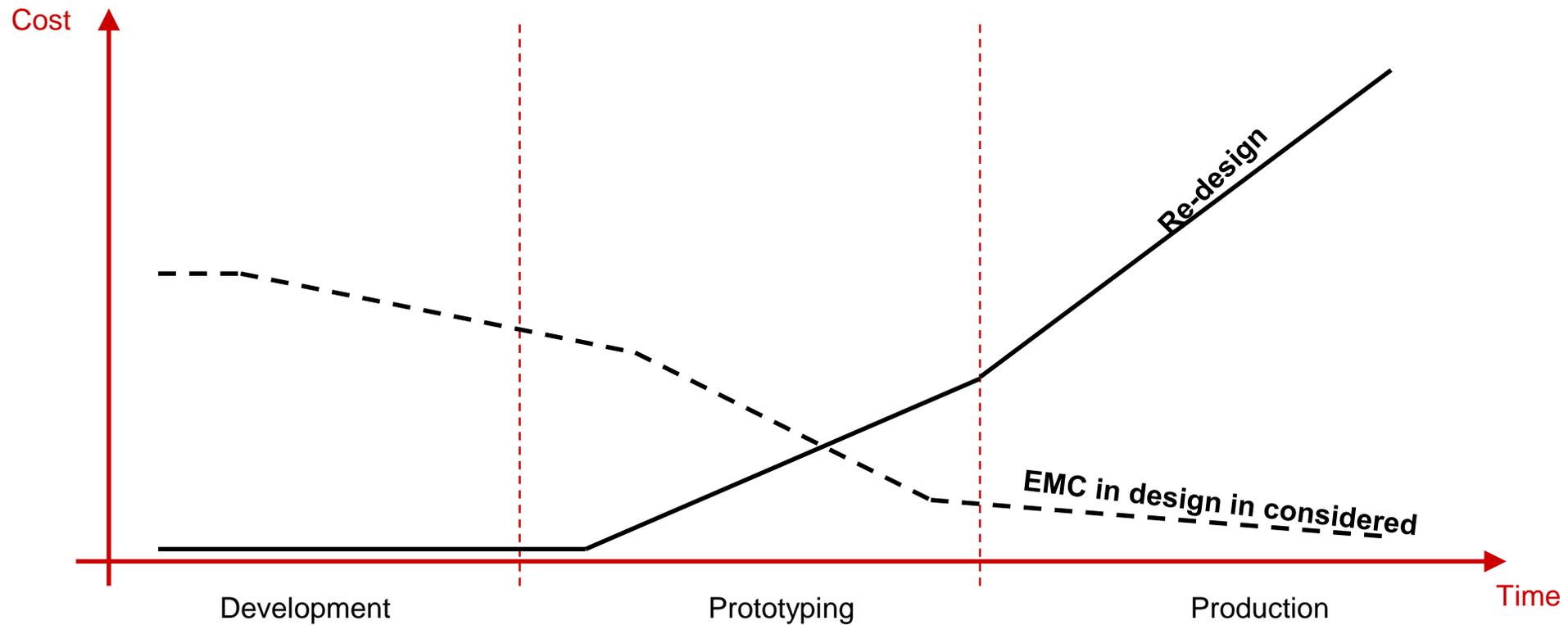
# EMC DELAVNICA 2023

Tomo Koželjnik; Field Application Engineer

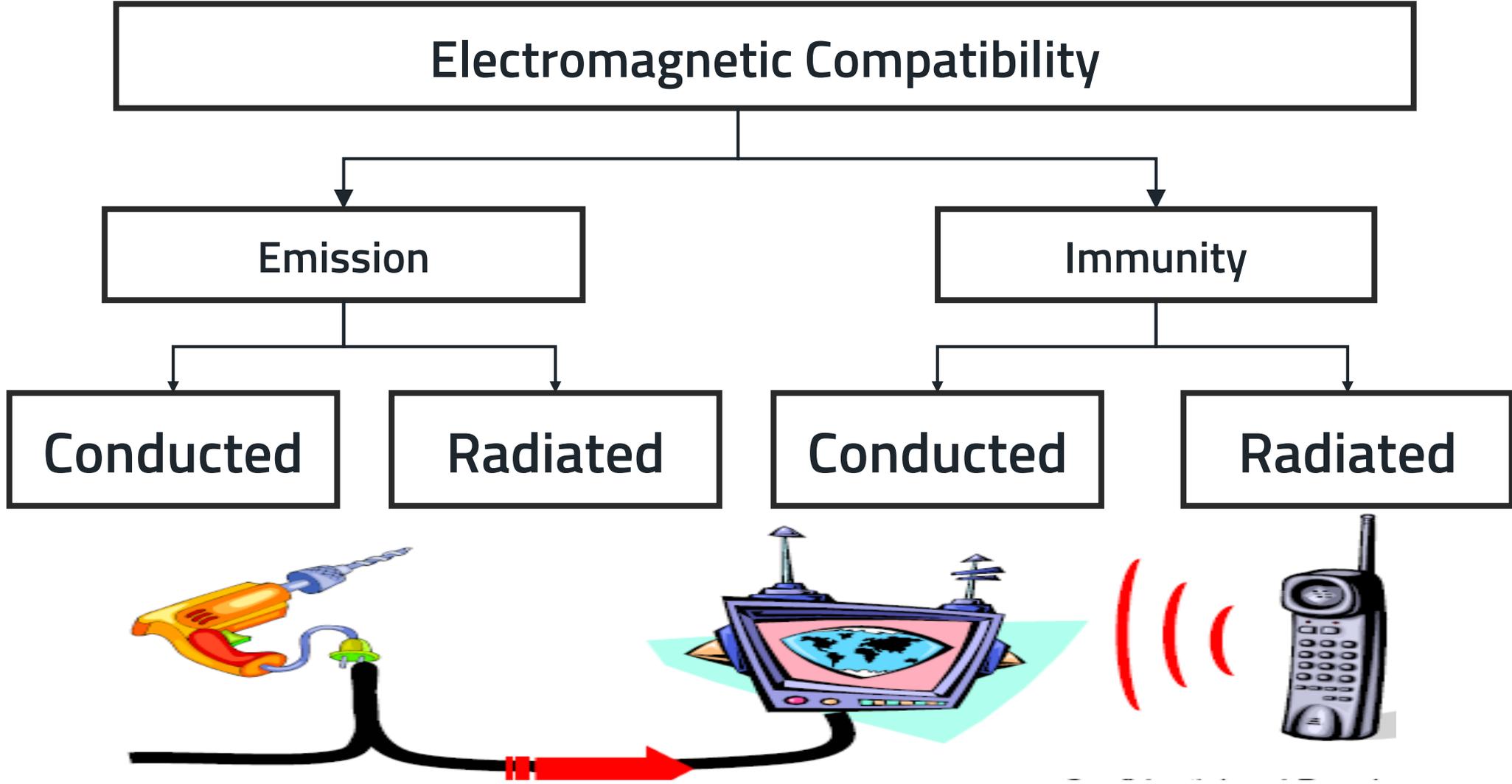
**WÜRTH ELEKTRONIK** MORE THAN YOU EXPECT

## DESIGN PHASE FOR EMC

- Economical point of view:
- Depends on you when will start to design EMC conform



# EMC – BASIC TESTS



# CONDUCTED EMISSION

- Conducted emission over wideband
- Caused by ripple current at input lines (common mode - / differential mode noise)
- EMC requirements for „*Conducted Emission*“ according ETSI, CEN, CENELEC
- E.g.: EN 55011: 2016 (Industrial, scientific and medical (ISM) radio-frequency equipment)

66 - 56dB $\mu$ V @ 150<KHz<500KHz (QP)

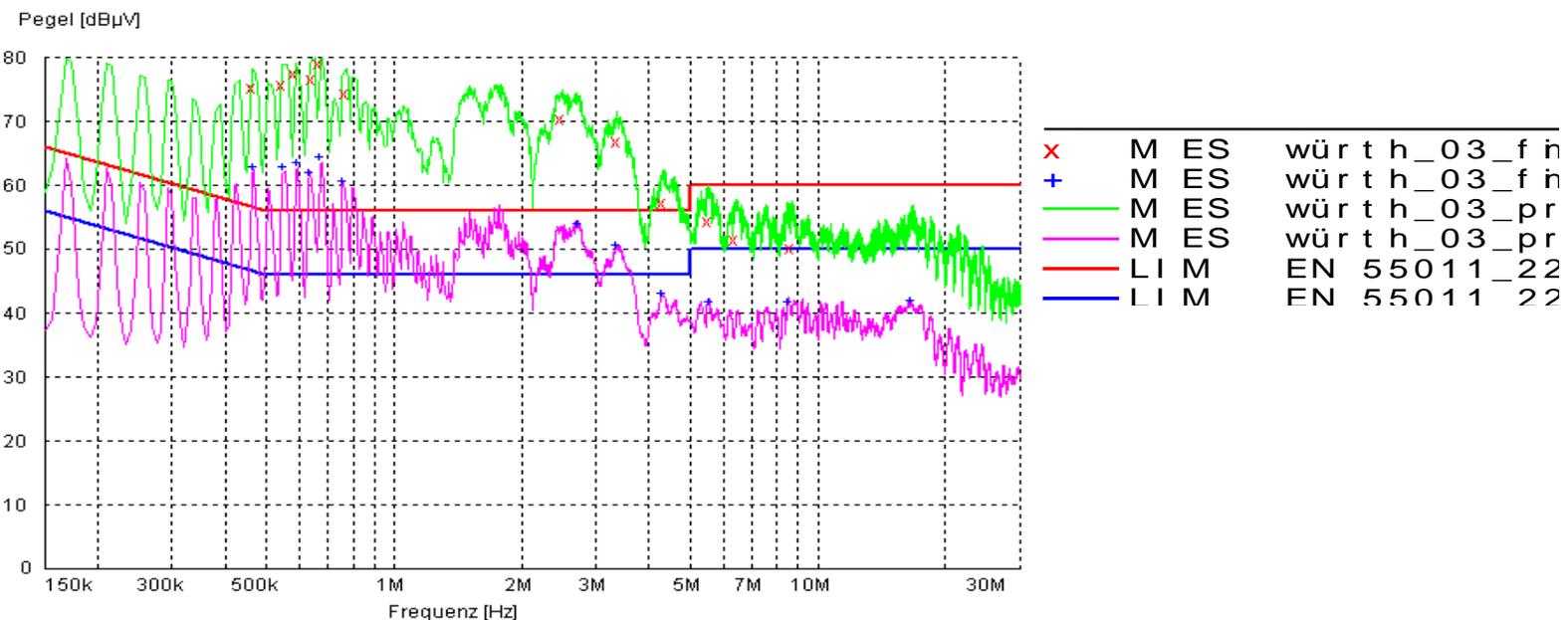
56 - 46dB $\mu$ V @ 150<KHz<500KHz (Av)

56dB $\mu$ V @ 0,5<MHz<5 (QP)

46dB $\mu$ V @ 0,5<MHz<5 (Av)

60dB $\mu$ V @ 5<MHz<30 (QP)

50dB $\mu$ V @ 5<MHz<30 (Av)



# RADIATED EMISSION

- Radiated emission over wideband
- Caused by:
  - Power traces on PCB
  - Power choke of DC/DC converter
- EMC requirements for „*Radiated Emission*“ according ETSI, CEN, CENELEC

- **EN 61000-6-3** : 2011-09 (Home)

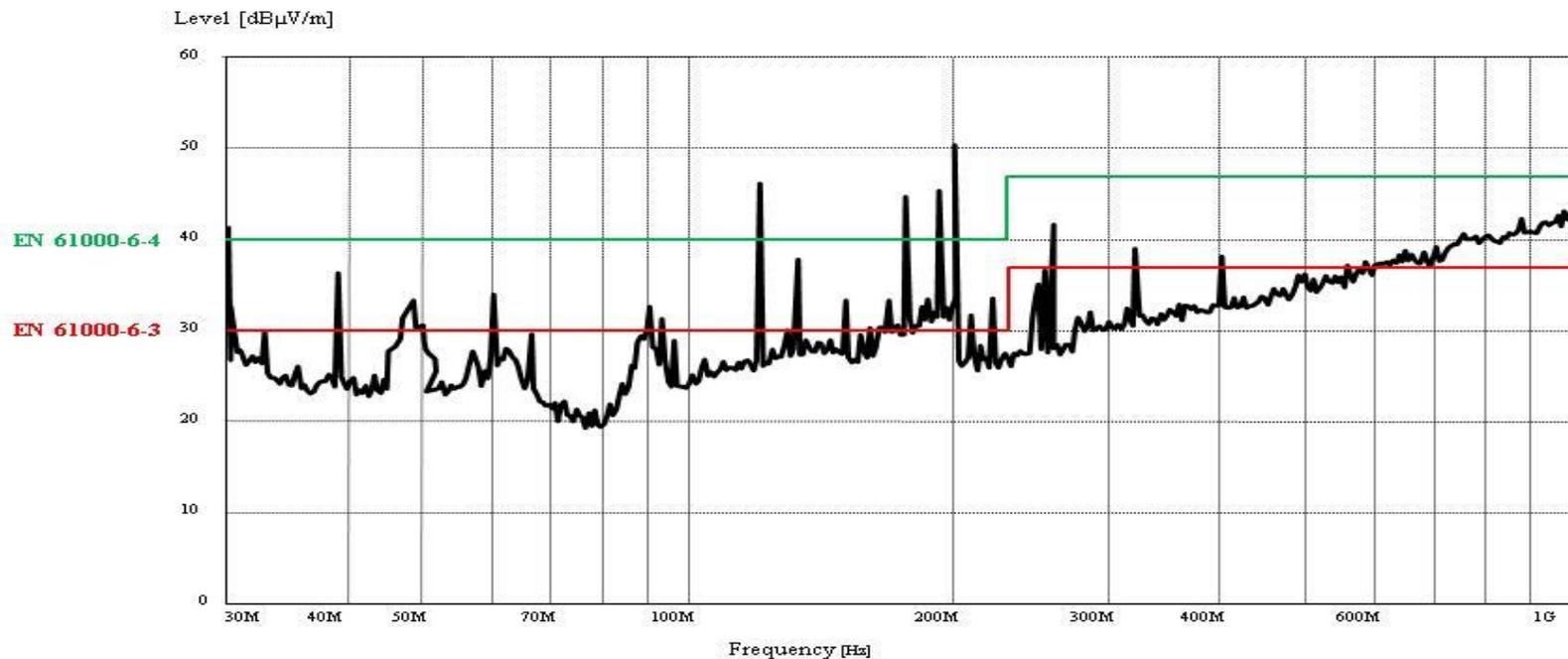
30dB @ 30MHz~230MHz  $\mu\text{V}/\text{m}$

37dB @ 230MHz~1GHz  $\mu\text{V}/\text{m}$

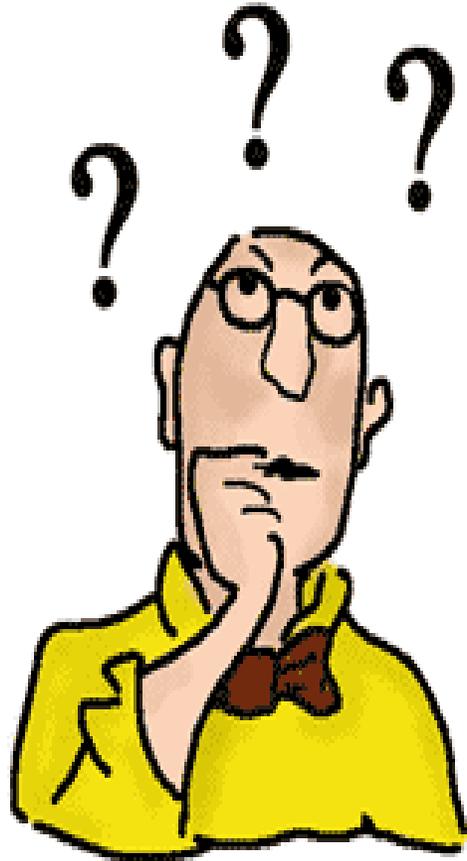
- **EN 61000-6-4** : 2011-09 (Industrial)

40dB @ 30MHz~230MHz  $\mu\text{V}/\text{m}$

47dB @ 230MHz~1GHz  $\mu\text{V}/\text{m}$



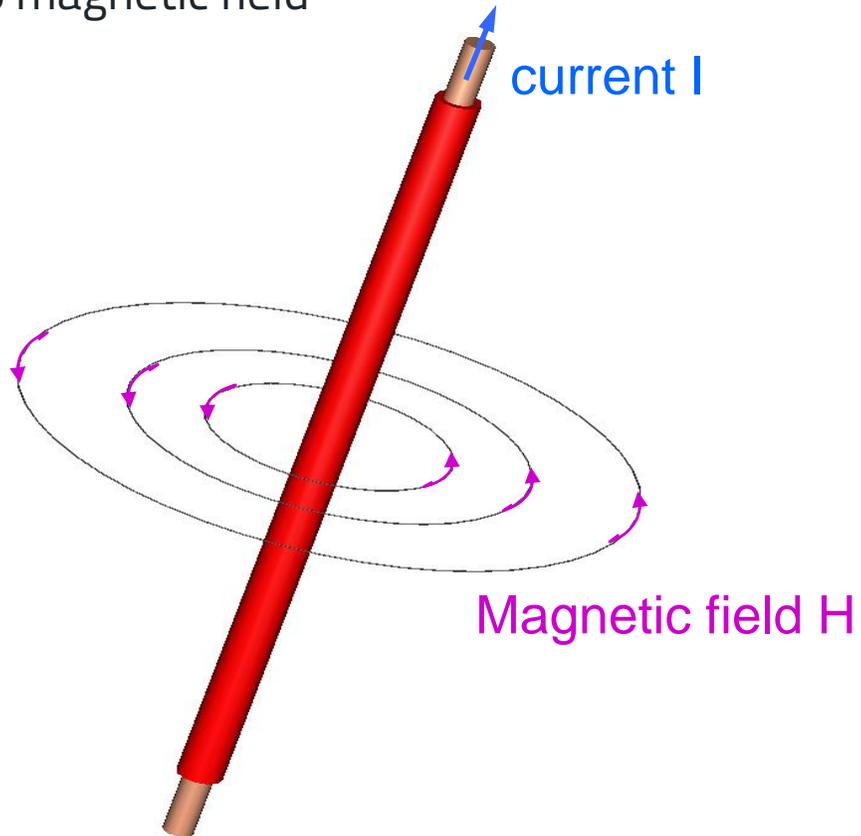
# HOW CAN WE CHECK THE EMC ?



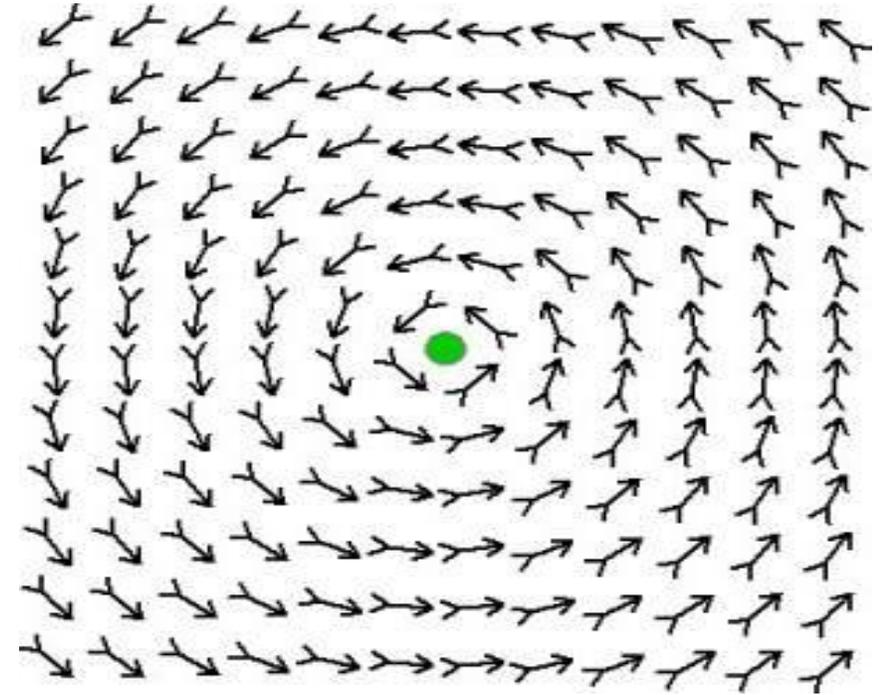
# Magnetic and Material Basics

# THE MAGNETIC FIELD

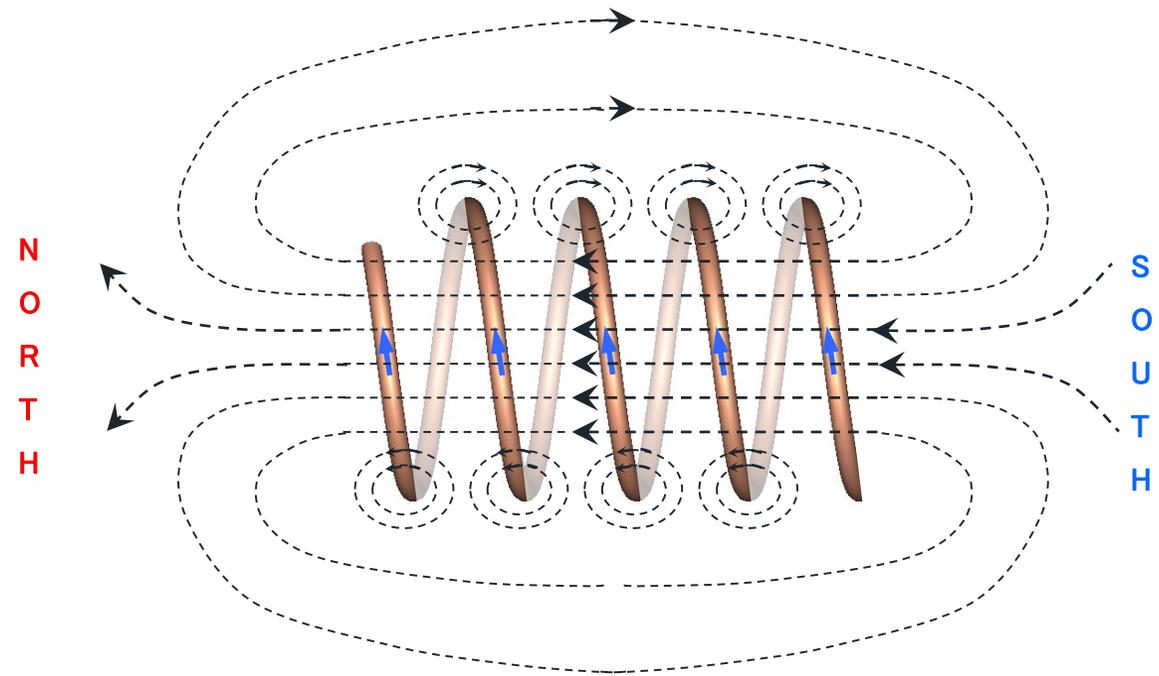
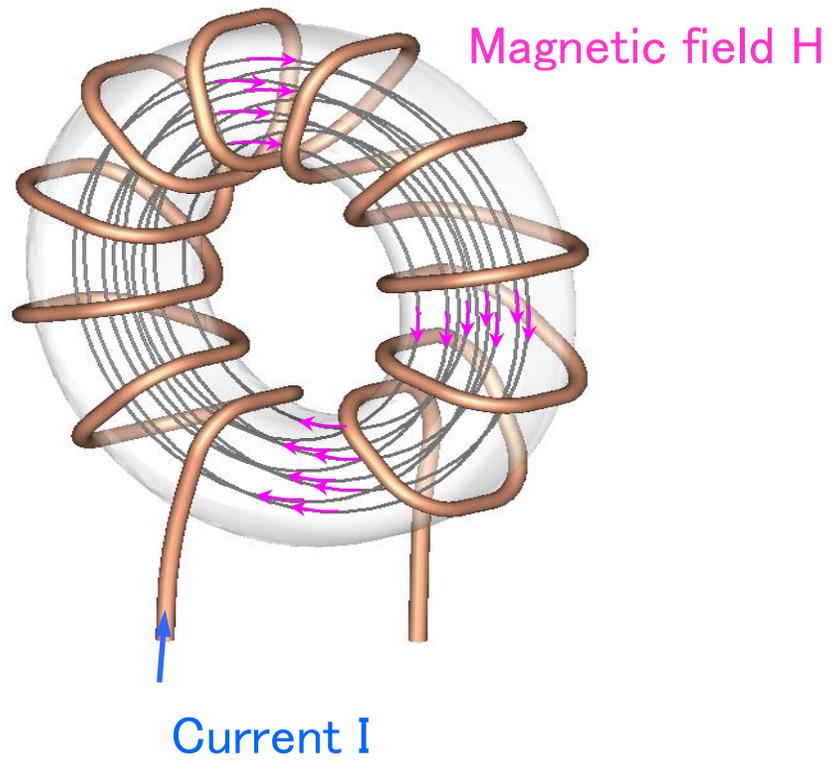
- Each electric powered wire generates an electro magnetic field



- Field model

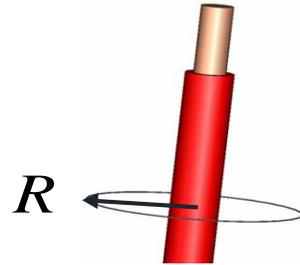


# THE MAGNETIC FIELD – FIELD MODEL



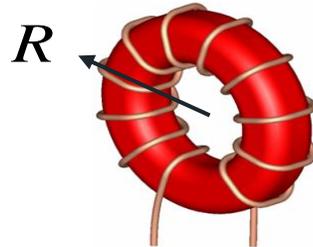
# MAGNETIC FIELD – MAGNETIC FIELD STRENGTH

Straight wire



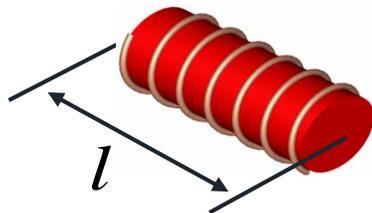
$$H = \frac{I}{2 \cdot \pi \cdot R}$$

Toroidal core



$$H = \frac{N \cdot I}{2 \cdot \pi \cdot R}$$

Rod core



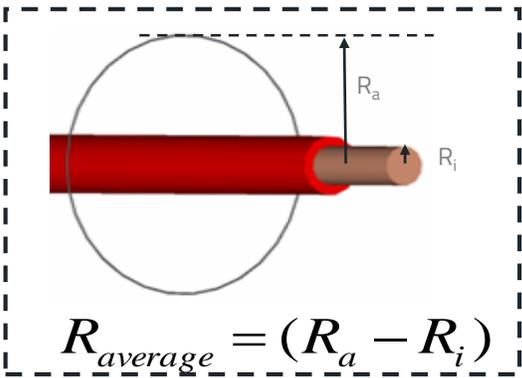
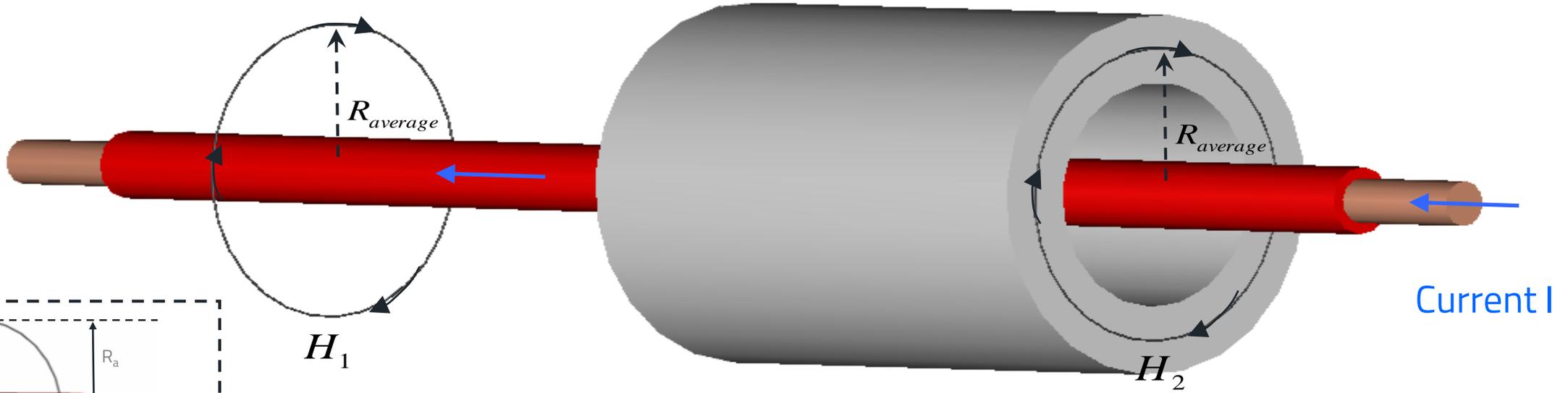
$$H = \frac{N \cdot I}{l}$$

The magnetic field strength is dependent from:

- No. of turns
- current
- dimension
- and

**NOT FROM MATERIAL**

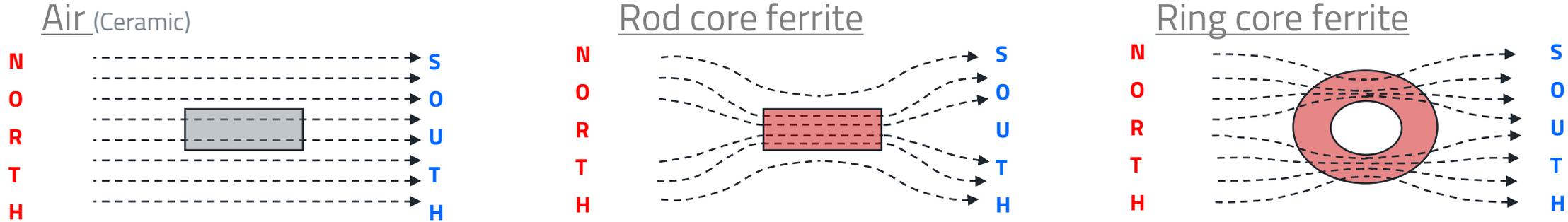
# MAGNETIC FIELD- MAGNETIC FIELD STRENGTH



$$H_1 = H_2 = H = \frac{I}{2 \cdot \pi \cdot R_{average}}$$

$B_1 \neq B_2$   
 $?$   
 $=$

# THE MAGNETIC FIELD



Induction in air:

$$B = \mu_0 \cdot H$$

linear function, because  $\mu_r = 1 \Rightarrow$  constant!

Induction in a ferrite:

$$B = \mu_0 \cdot \mu_r \cdot H$$

**material-  
frequency-  
temperature-  
current-  
pressure-**

**-dependant parameter**

**The relative permeability is a:**

# PERMEABILITY

- Typical value ranges of *WE* common-mode chokes:
  - Nickel zinc ferrite ( $\text{Ni}_x\text{Zn}_{1-x}\text{Fe}_2\text{O}_4$ ):  $\mu_{r,i} = 400 \dots 800$
  - Manganese zinc ferrite ( $\text{Mn}_x\text{Zn}_{1-x}\text{Fe}_2\text{O}_4$ ):  $\mu_{r,i} = 2000 \dots 10000$
  - Nanocrystalline material ( $\text{Fe}_{nc}$ ):  $\mu_{r,i} = 5000 \dots 95000$
- The *ExB* series is a combination of different core materials.
  - NiZn:  $\mu_{r,i} = 400$
  - MnZn:  $\mu_{r,i} = 6000$



# PERMEABILITY – CORE MATERIAL PARAMETER

## Temperature influence

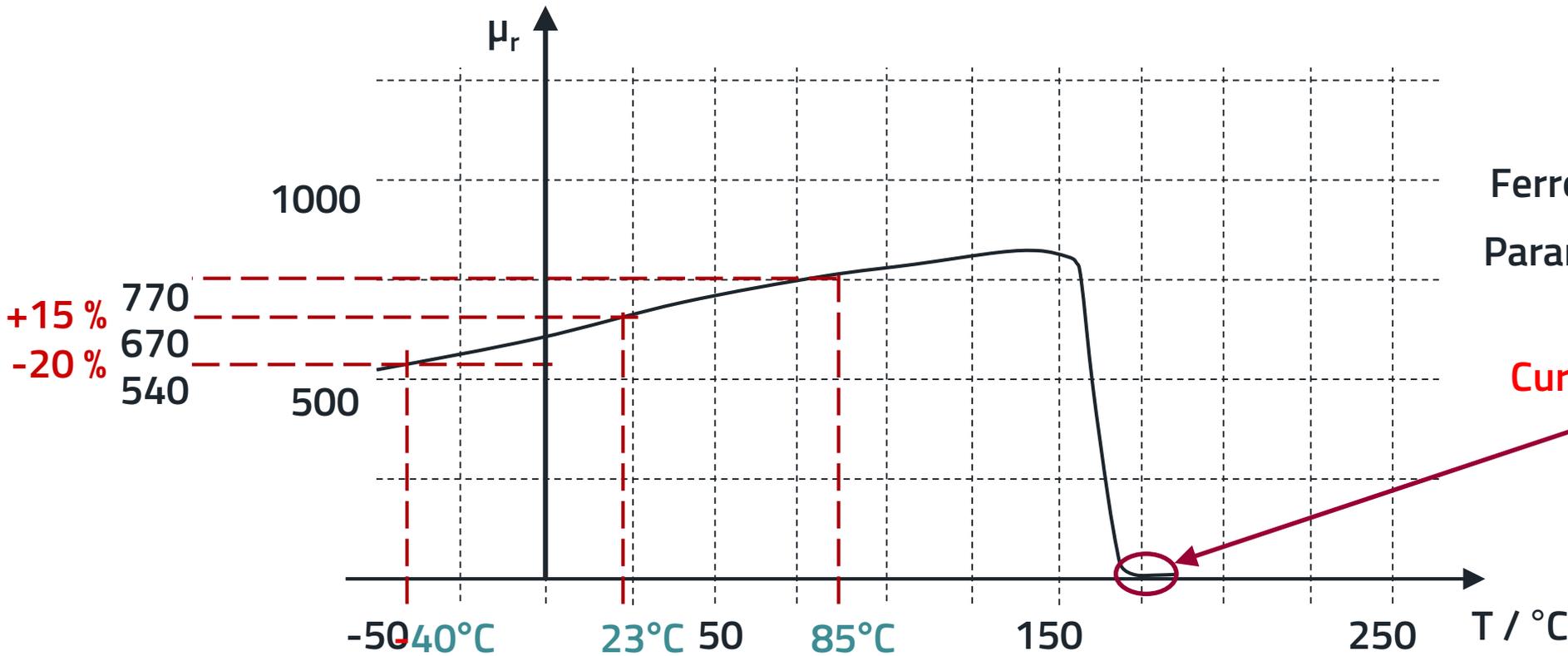
- The magnetization depends from the temperature



Alignment of elementary magnets



Ferromagnetic change to Paramagnetic



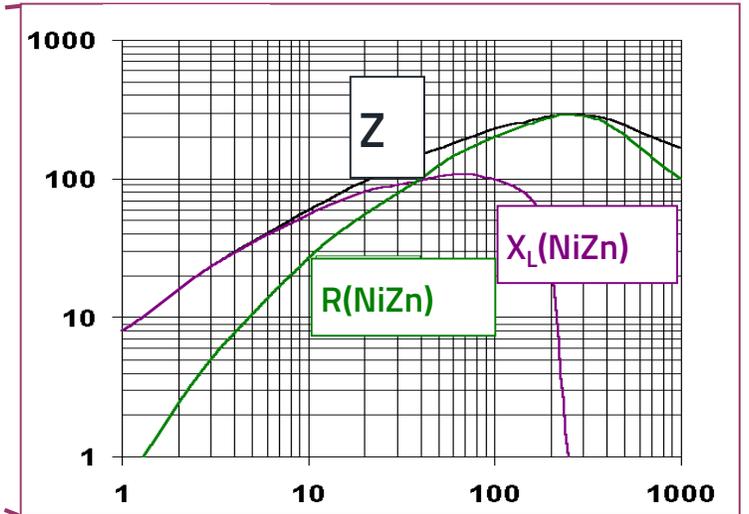
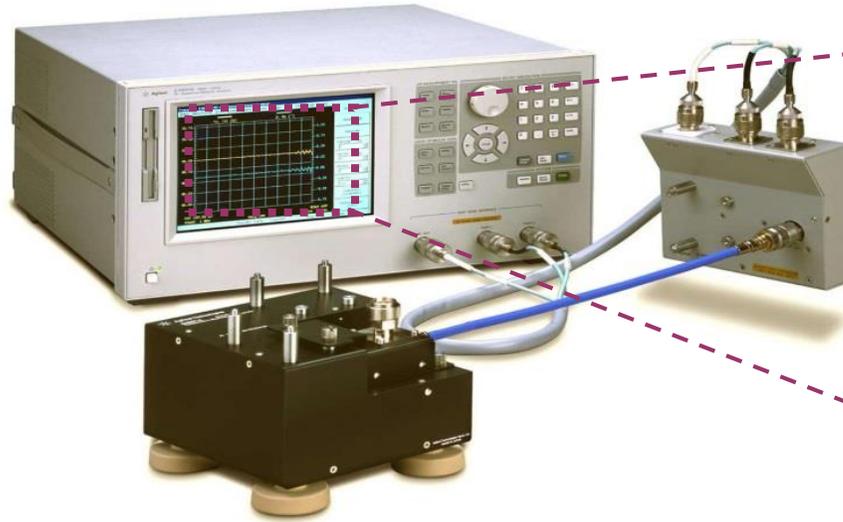
Curie-temperature

$$\mu_r = ? 1$$

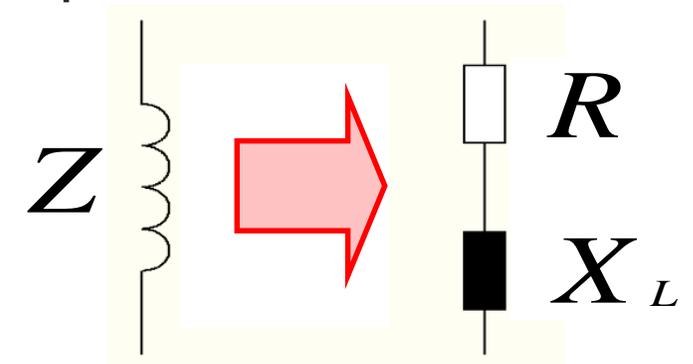
# PERMEABILITY – CORE PARAMETER



=1 turn



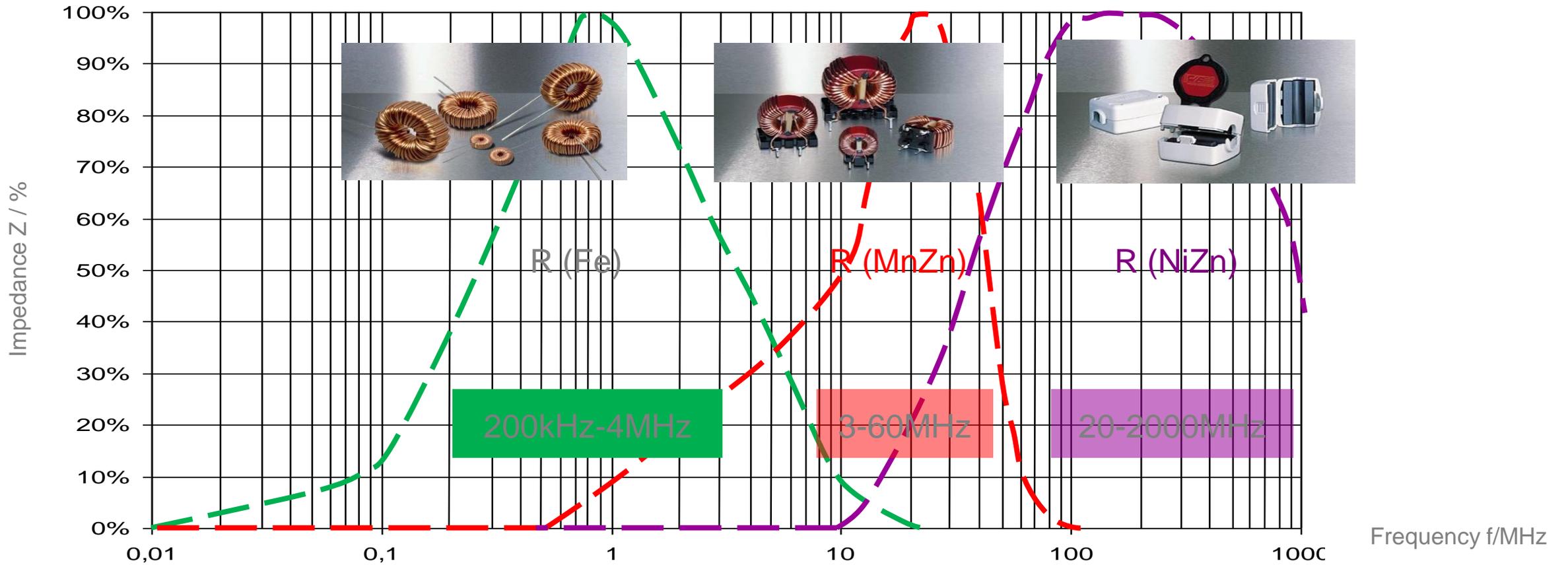
Core material-Parameter  
Replacement circuit



$$Z = \sqrt{R^2 + X_L^2}$$

# CORE MATERIALS- CHOKES (FILTERING)

Noise frequency range must be known



## RECOGNIZING THE COUPLING MODE

- common mode noise ?
- differential mode noise ?



## COMMON MODE OR DIFFERENTIAL MODE?

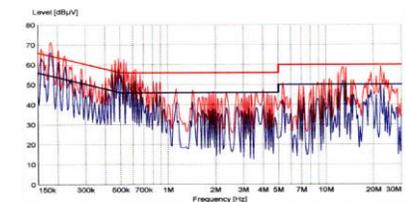
Take a Snap Ferrite and fix it on the cable  
(both lines e.g. VCC and GND)

if noise is reduced or  
noise immunity increase

you have Common Mode Interference

If not

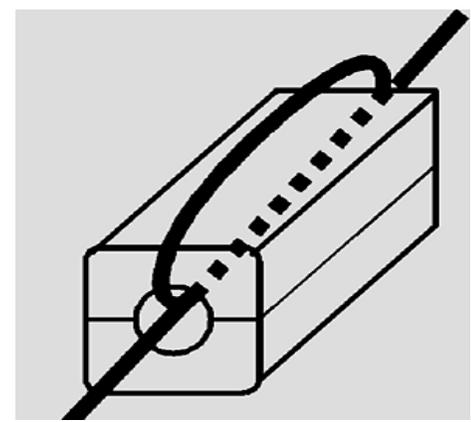
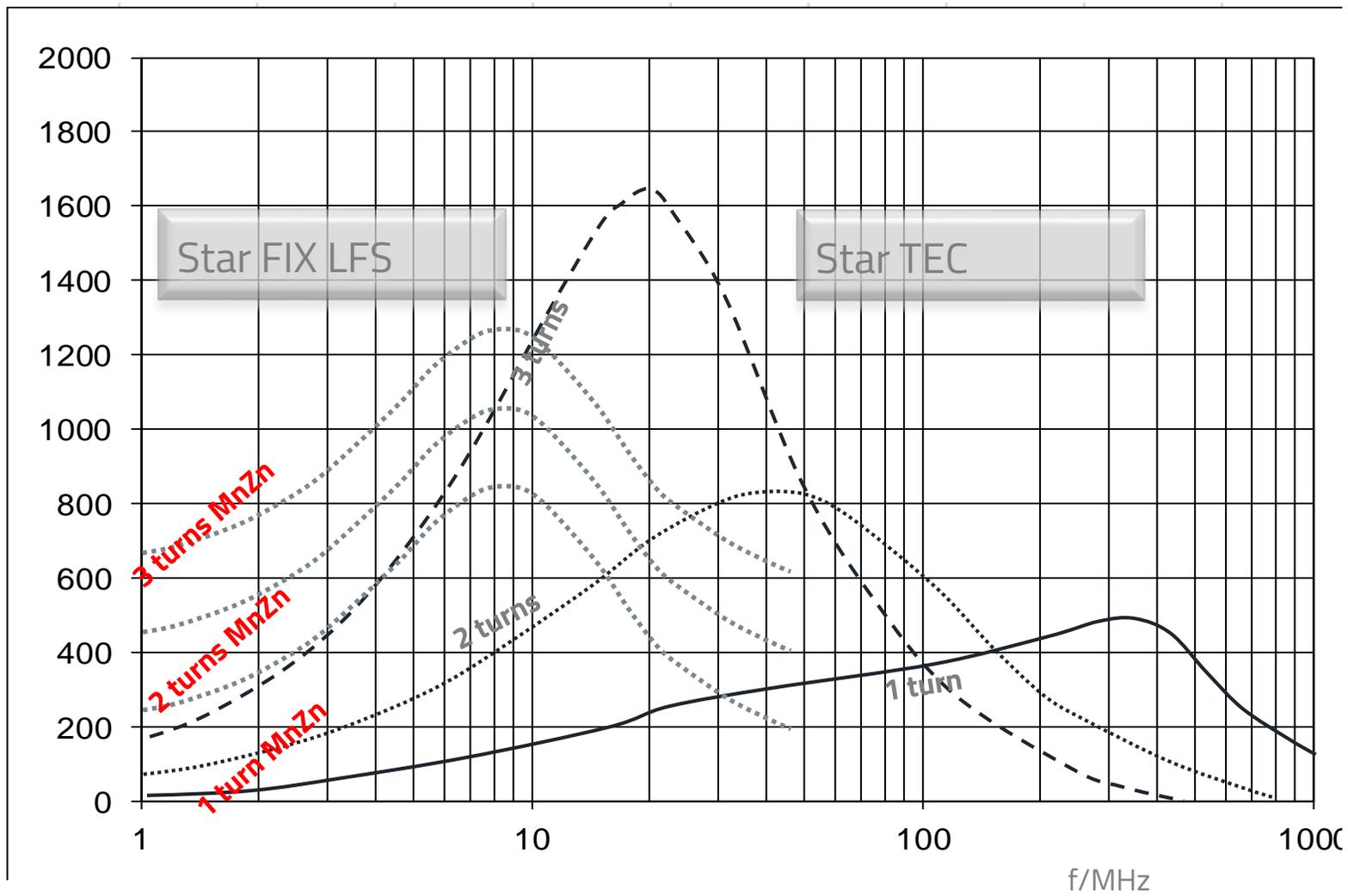
you have Differential Mode Interference



e.g. Common mode  
choke

e.g. chip bead ferrite

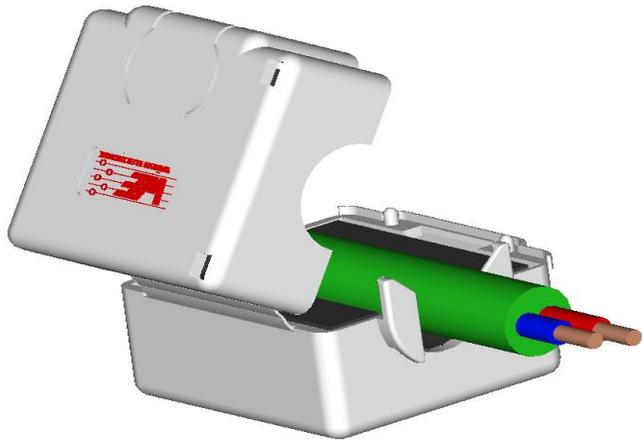
# SNAP ON FERRITE – TYPICAL BEHAVIOR



Increase the no. of turns means:

## SNAP ON FERRITE - CONSTRUCTION

- Snap on ferrite acts as an CMC
- Absorbs common mode Interferences
- Comparable with bifilar winding CMC



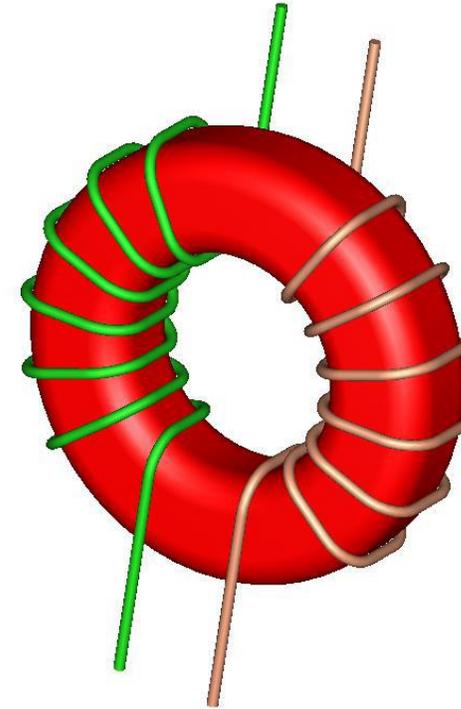
## COMMON MODE CHOKE – WINDING STYLE

### ■ bifilar



- Power supply: low voltage application
- Signal: low speed signal and High speed Signal

### ■ sectional



- Power supply: „high“ voltage application and low voltage DC applications; according to IEC60938 or UL1283

## COMMON MODE FILTER – HOW IT WORKS

It is a Bi-directional filter

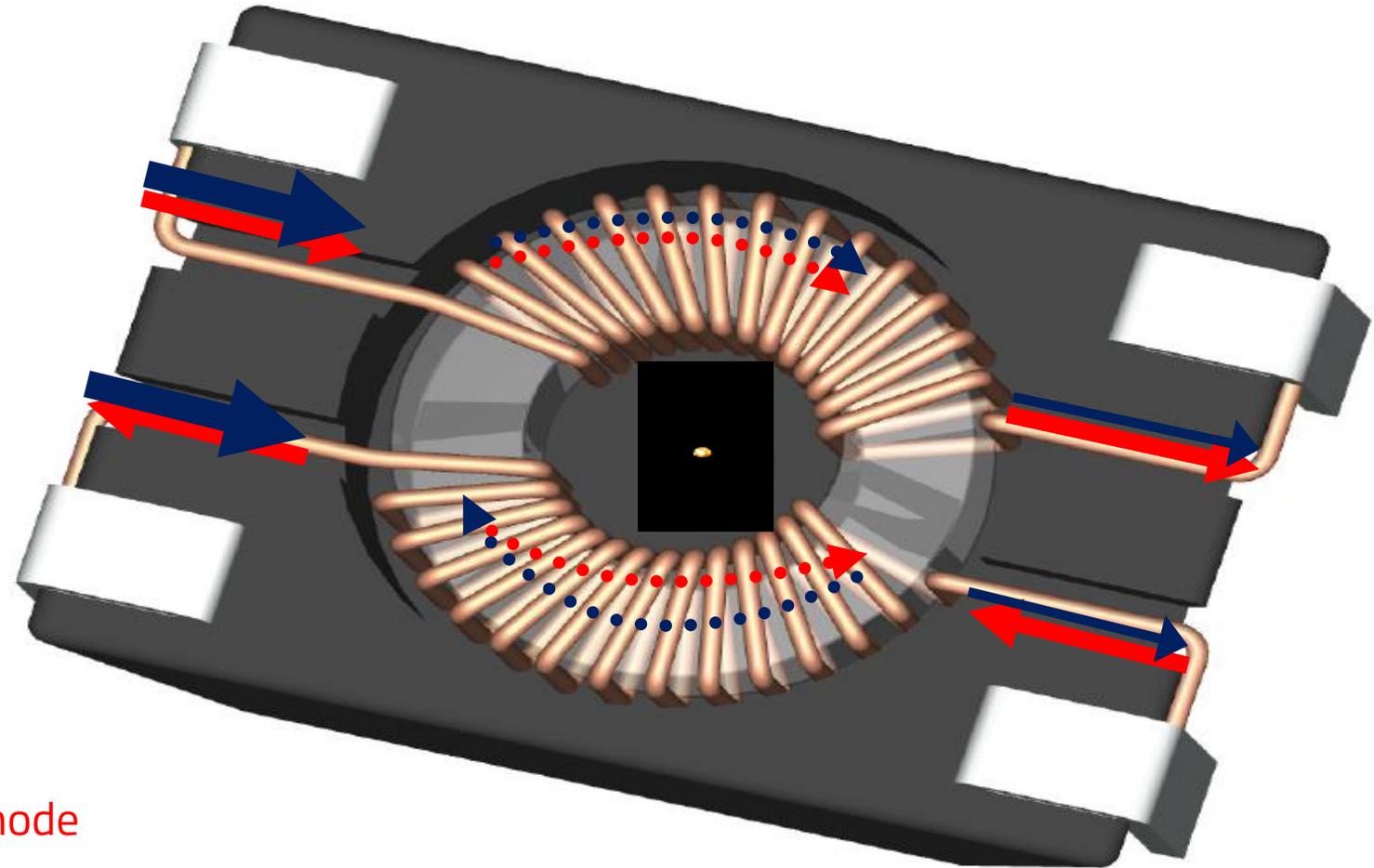
- From device to outside environment
- From outside environment to inside device

Intended Signal - **Differential mode**

Interference Signal (noise) – **Common Mode**

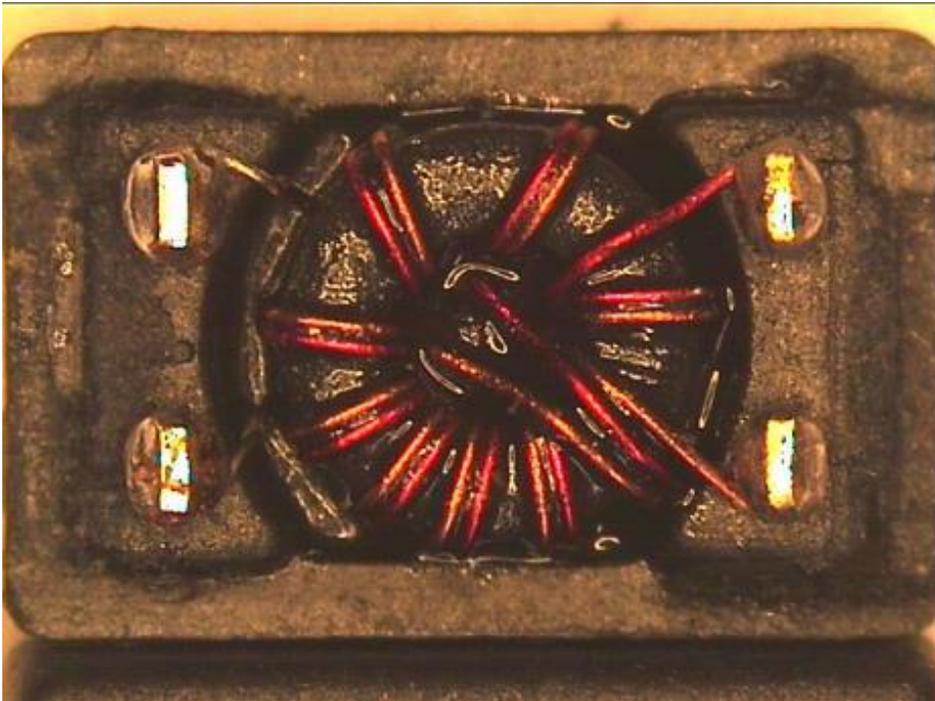
Conclusion:

- “almost” no affect the signal - **Differential mode**
- high attenuation to the interference signal (noise) – **Common Mode**



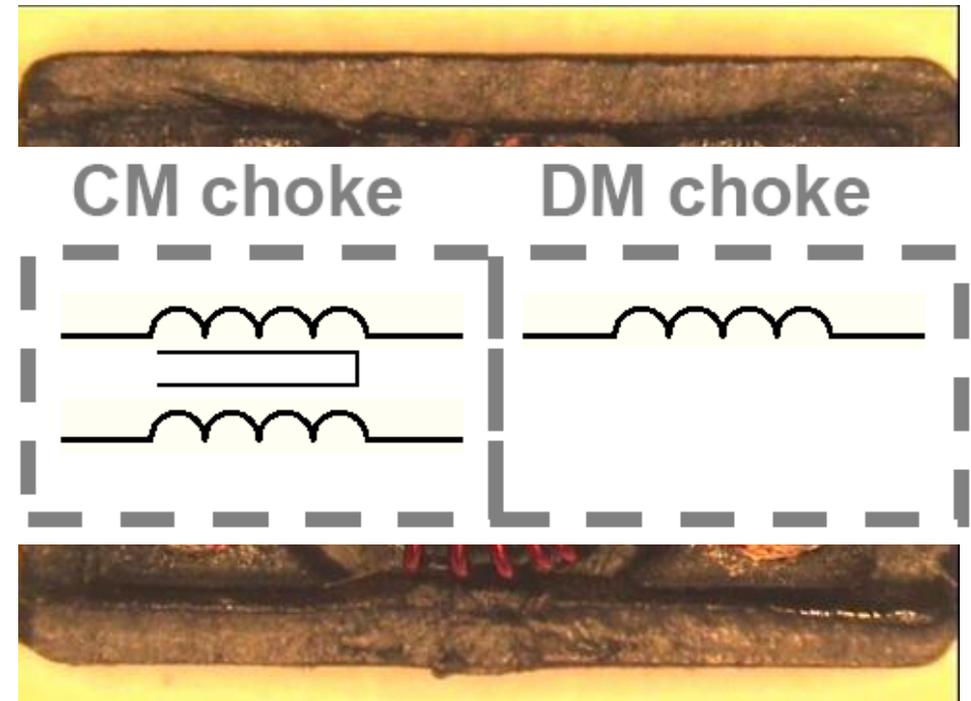
## COMMON MODE CHOKE – WINDING STYLE

- bifilar WE-SL2 744226



- low leakage inductor

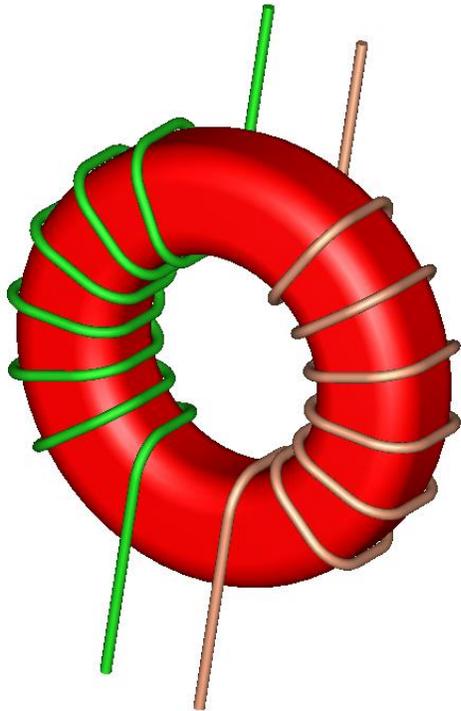
- sectional WE-SL2 744226S



- high leakage inductor

# LEAKAGE INDUCTANCE – WINDING STYLE

Sectional



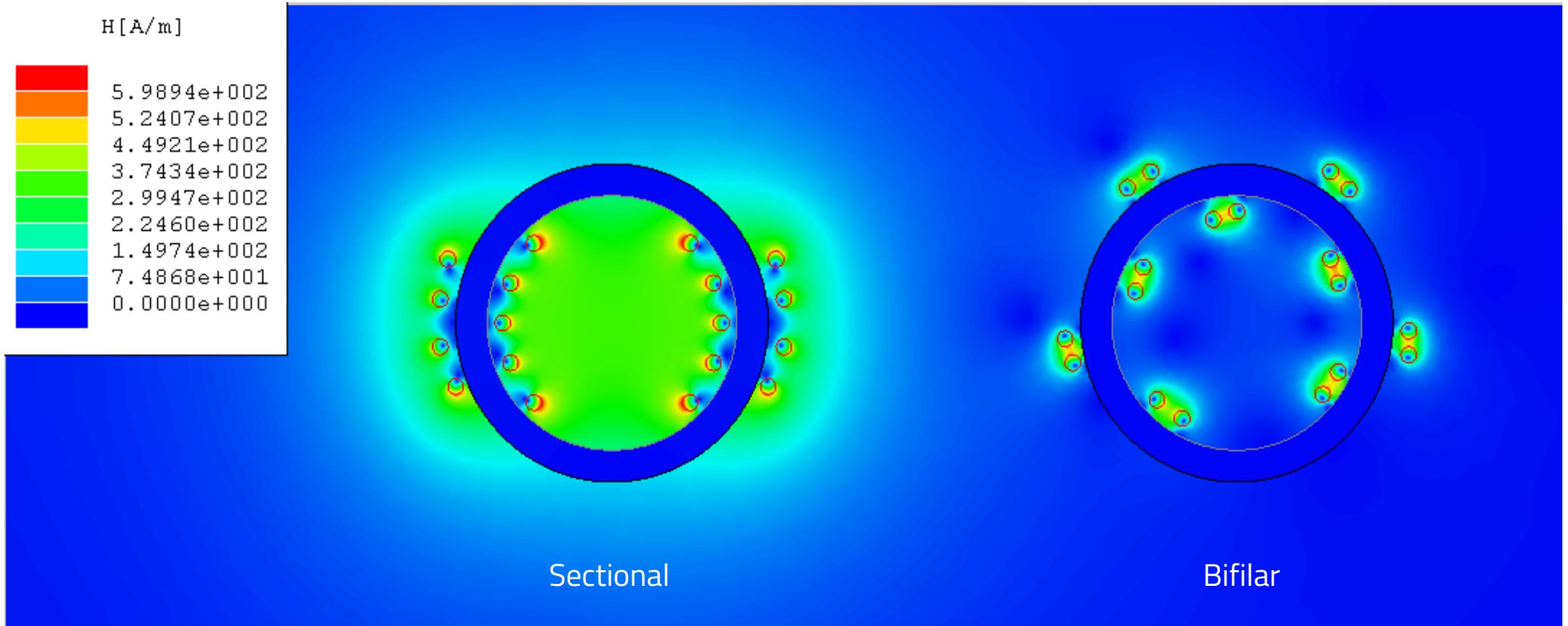
$$L_{\text{leak}} = (0,005 \dots 0,05) \cdot L$$

Bifilar



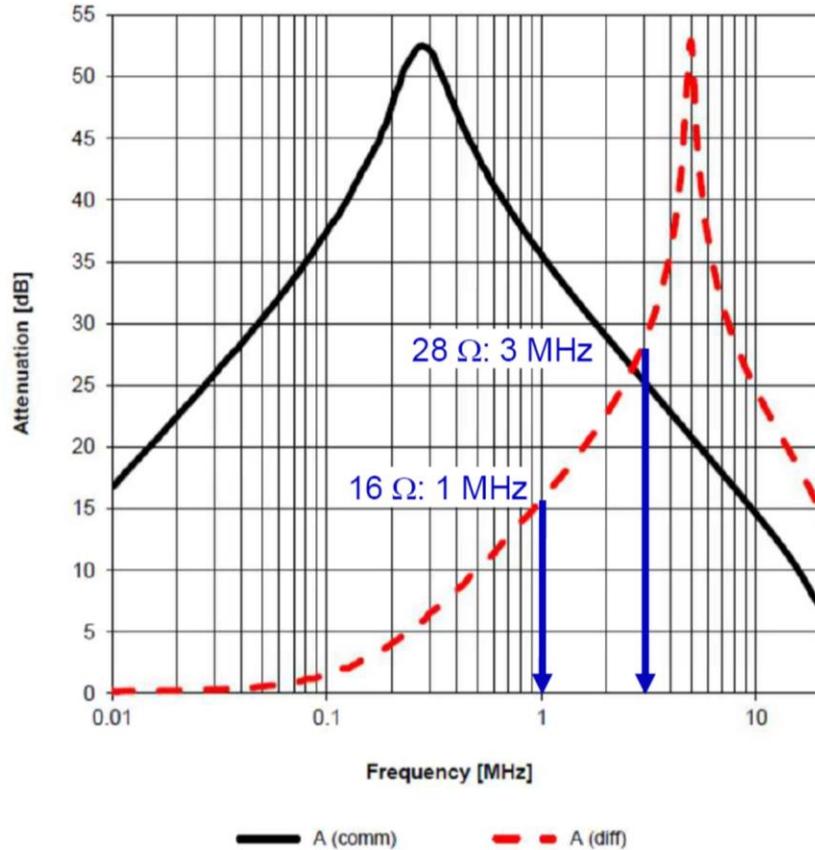
$$L_{\text{leak}} = (0,001 \dots 0,002) \cdot L$$

# LEAKAGE INDUCTANCE – WINDING STYLE



# LEAKAGE INDUCTANCE – WINDING STYLE

- Calculation approach
- Example of leakage inductance



Properties	Test conditions		Value	Unit	Tol.
Inductance	10 kHz/ 0.1 mA	L	2x10	mH	±30
Rated Current	@ 70 °C	I <sub>R</sub>	0.7	A	max.
DC Resistance	@ 20 °C	R <sub>DC</sub>	2x350	mΩ	max.
Rated Voltage	50 Hz	U <sub>R</sub>	250	V (AC)	max.
Insulation Test Voltage	50 Hz/ 5 mA/ 2 sec.	U <sub>T</sub>	1500	V (AC)	

$$L_{\text{diff}} = Z_f / \omega$$

$$L_{\text{diff\_1MHz}} = 16 \Omega / (2\pi \times 10^6) = 2,5 \mu\text{H}$$

$$L_{\text{diff\_3MHz}} = 28 \Omega / (2\pi \times 3 \times 10^6) = 1,5 \mu\text{H}$$



# COMPARISON OF THE CM ATTENUATION

744821201 ✕

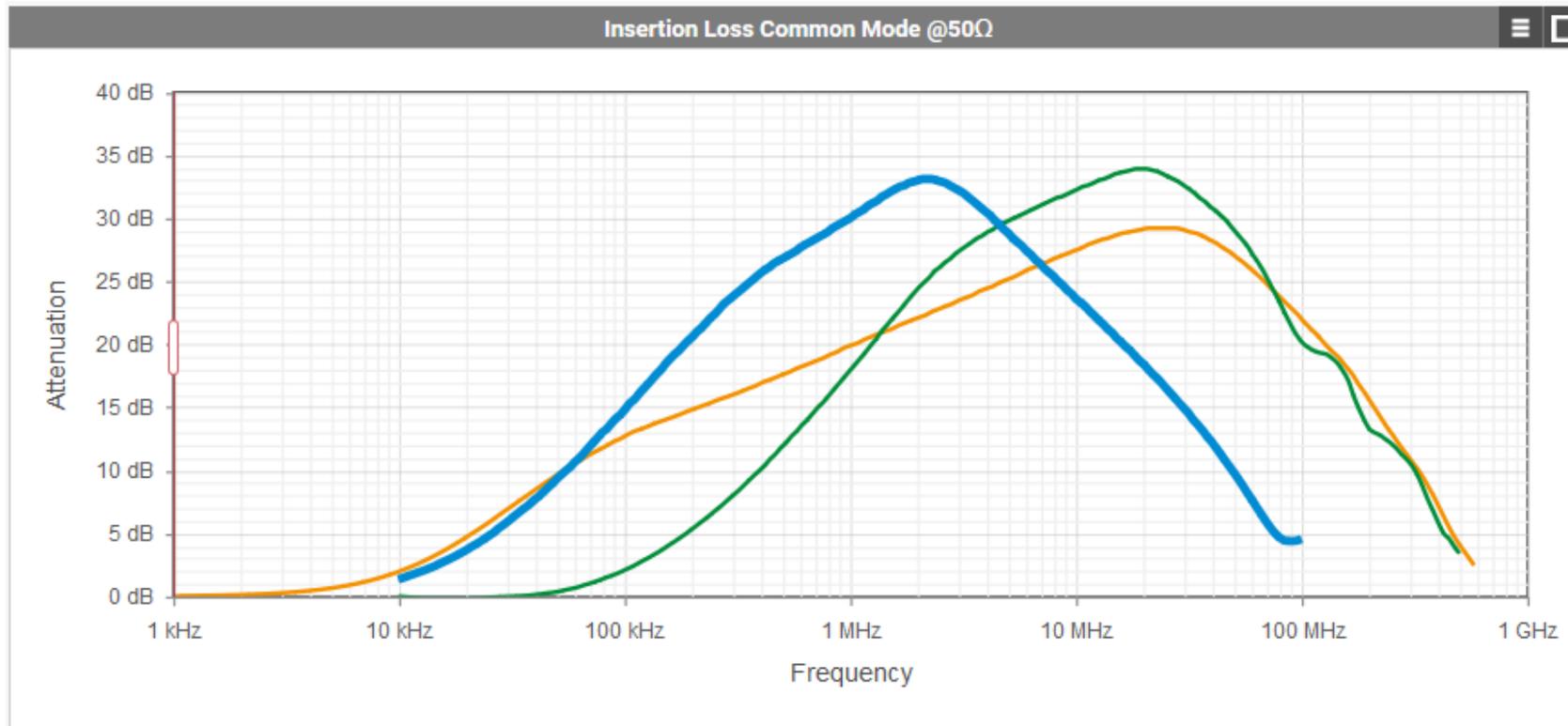
WE-CMB · XS  
1,00 mH · 2,00 A

7448012501 ✕

WE-CMBNC · XS  
1,00 mH · 2,50 A

744841210 ✕

WE-CMB NiZn · XS  
100 µH · 1,50 A



MnZn



NK



NiZn

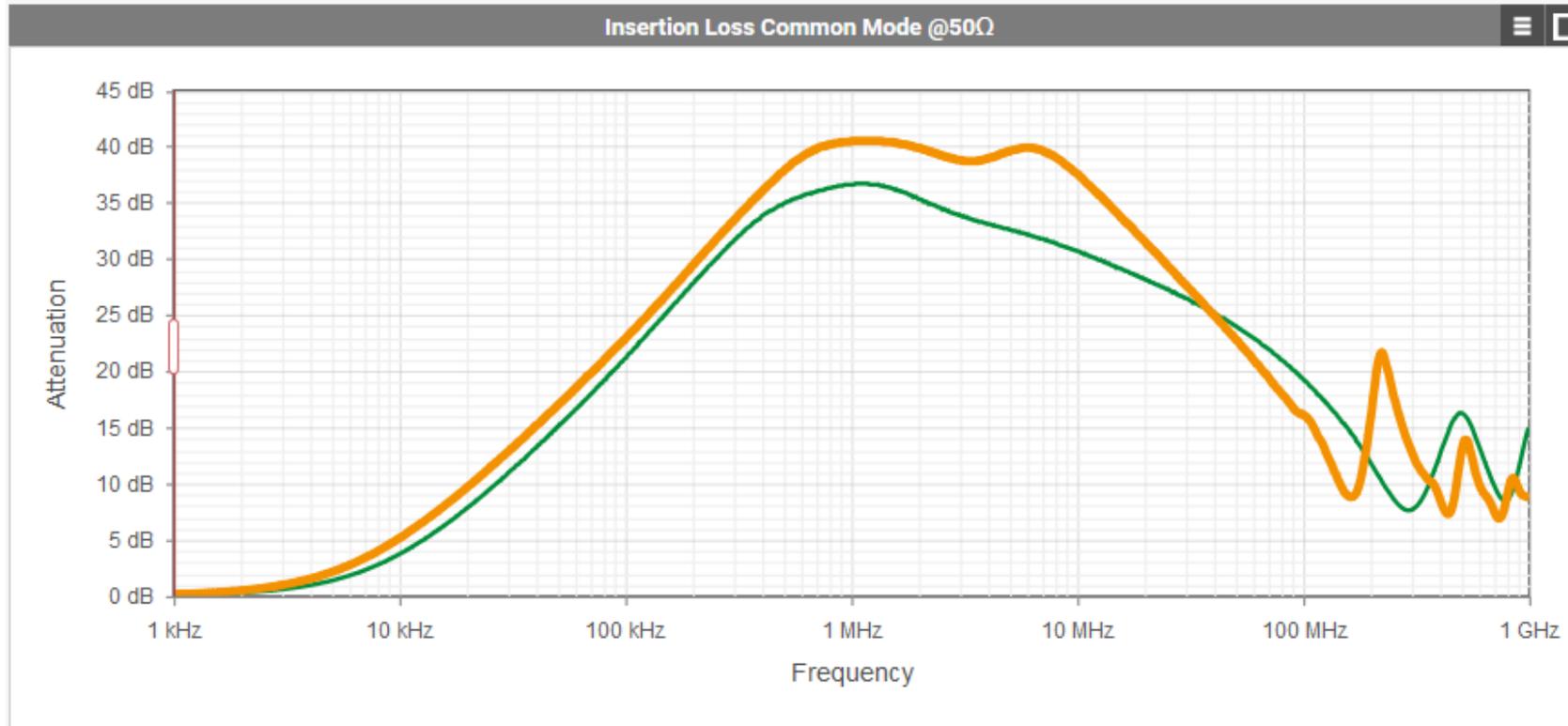
# COMPARISON OF THE CM ATTENUATION

744844102 ✕

WE-ExB · L  
1,00 mH · 4,50 A

744834101 ✕

WE-CMBH · L  
1,00 mH · 10,0 A



MnZn+NiZn



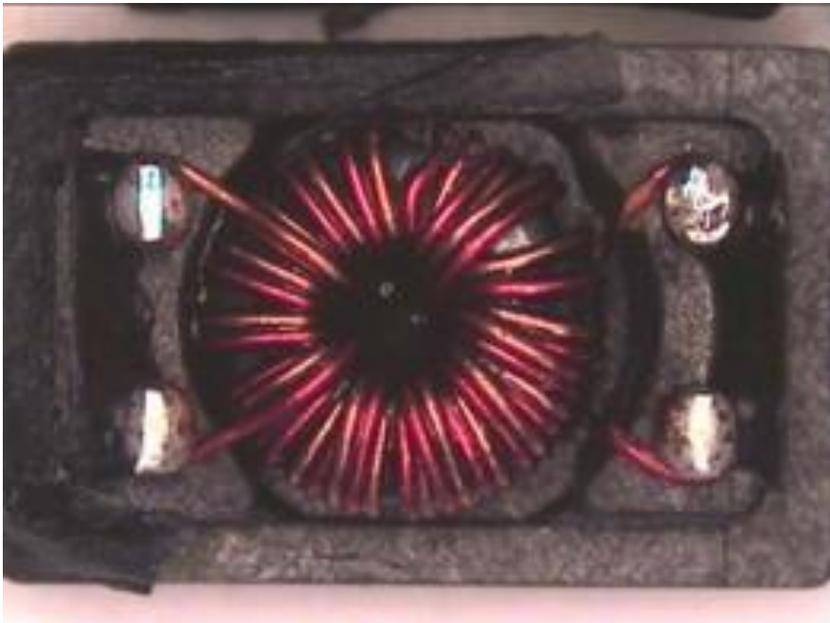
MnZn

## IMPEDANCE – WINDING STYLE

WE-SL2 744227

$$L_{cm} = 51\mu\text{H}$$

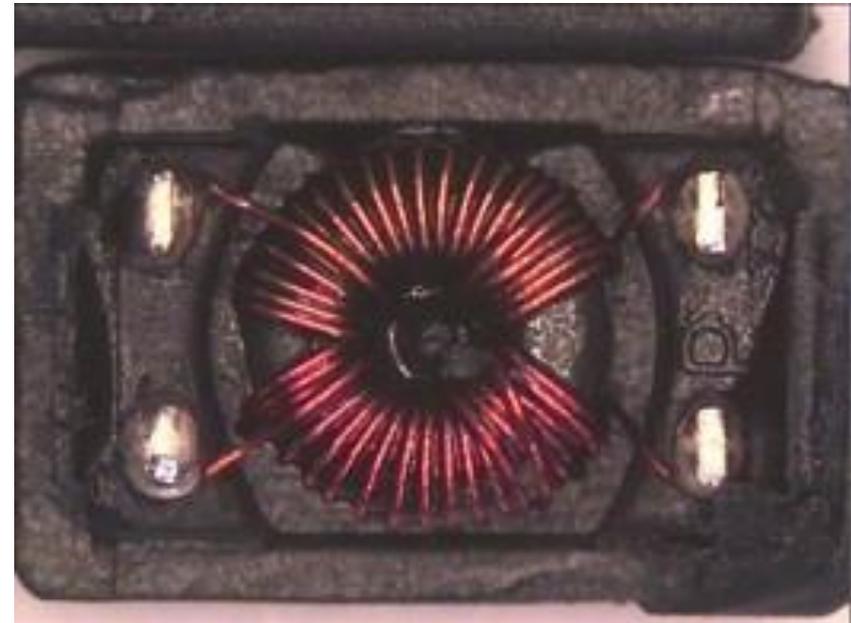
**Bifilar** winding



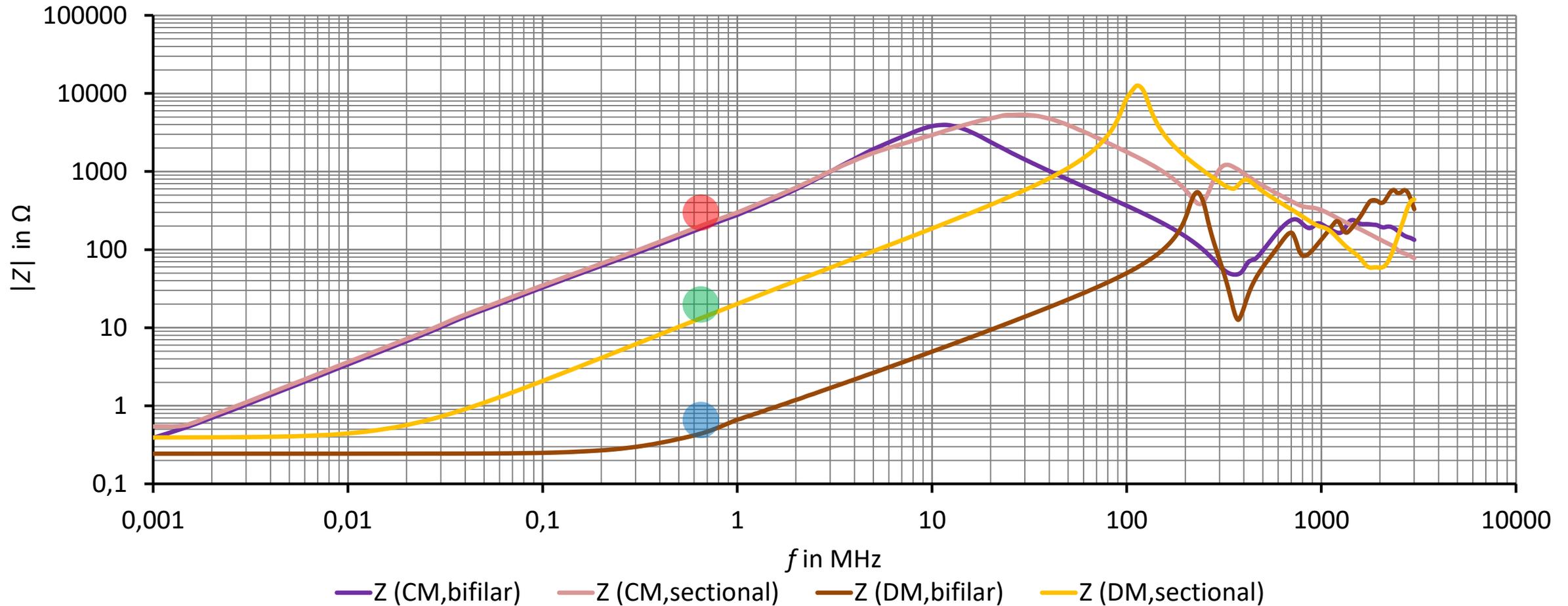
WE-SL2 744227S

$$L_{cm} = 51\mu\text{H}$$

**Sectional** winding

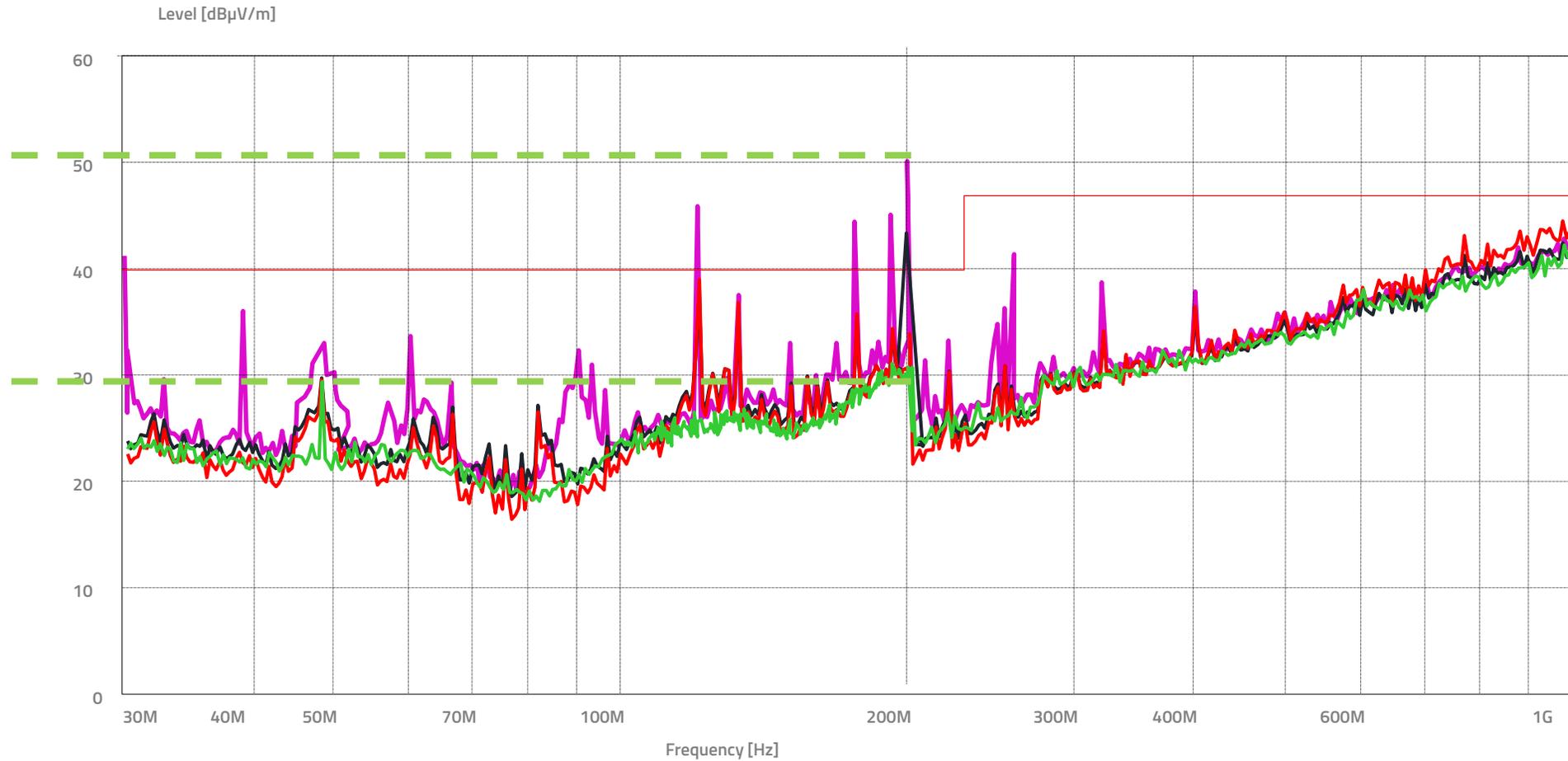


# IMPEDANCE – WINDING STYLE

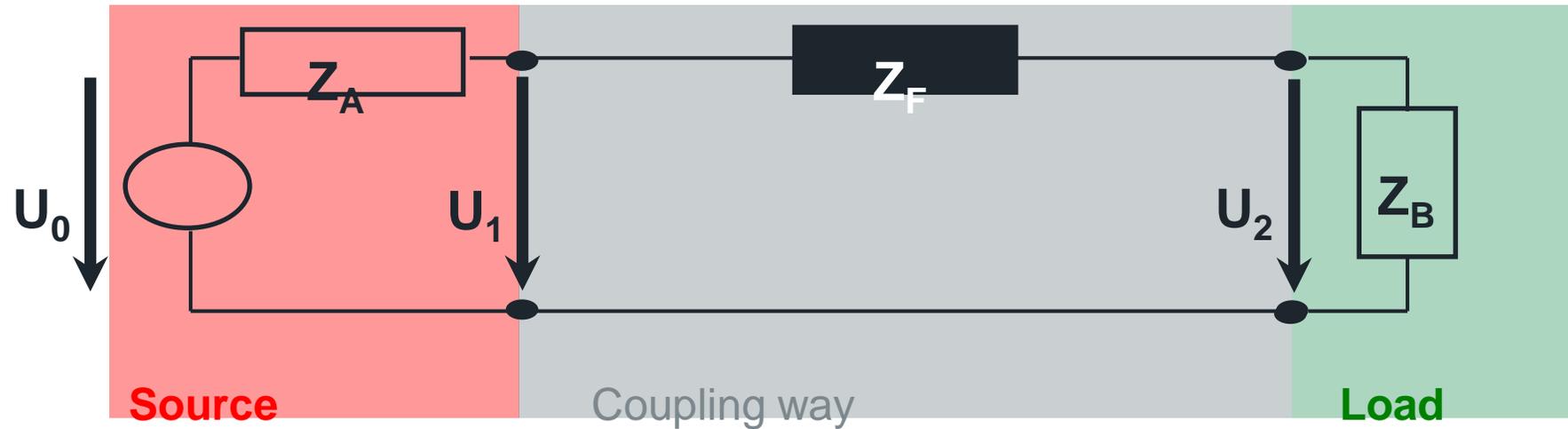


# CHECK THE RESULTS IN THE EMC LAB

→ Measuring the emission and compare with the solution



## INSERTION LOSS – MATHEMATICAL DEFINITION



- System attenuation

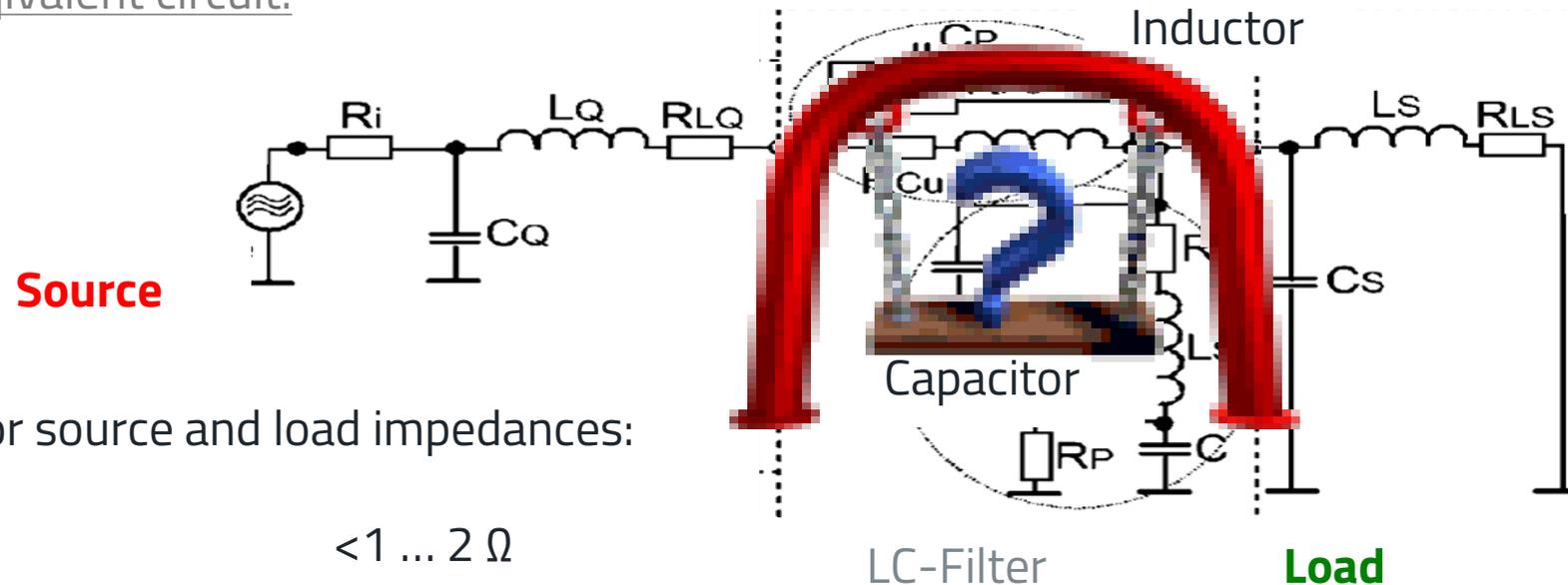
$$A = 20 \cdot \log \frac{Z_A + Z_F + Z_B}{Z_A + Z_B} \quad \text{in (dB)}$$

- Impedance

$$Z_F = \left[ 10^{\frac{A}{20}} \cdot (Z_A + Z_B) \right] - (Z_A + Z_B) \quad \text{in } (\Omega)$$

# INSERTION LOSS - DEFINITION

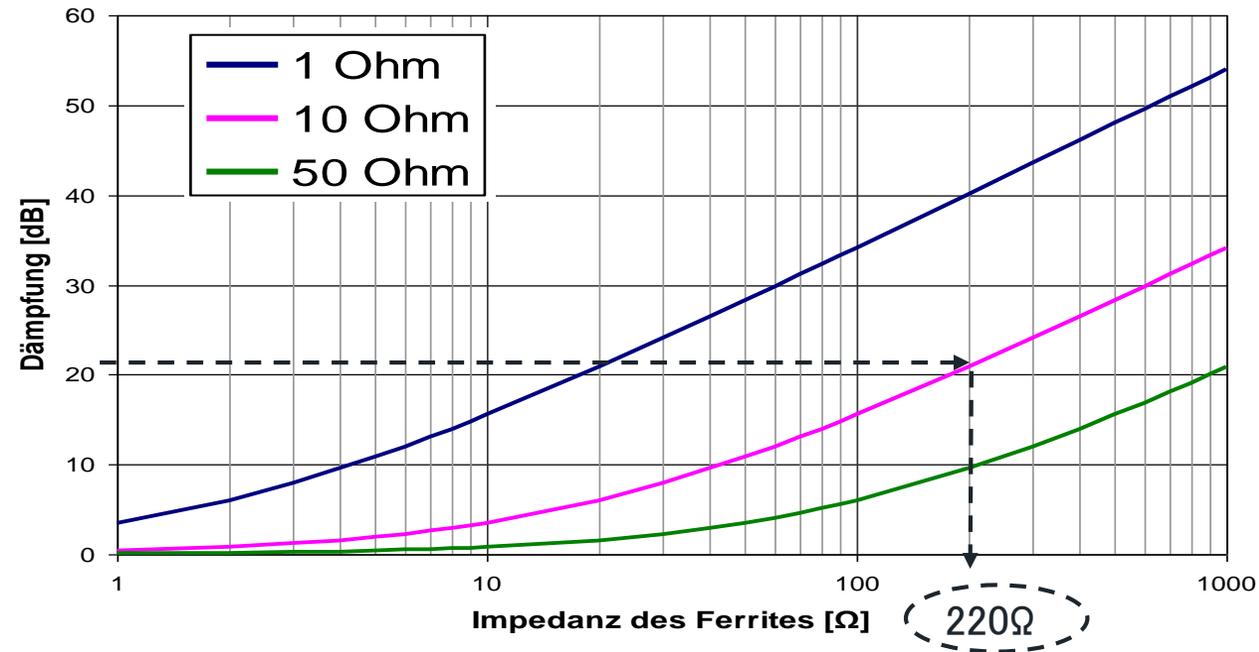
Equivalent circuit:



- Practical values for source and load impedances:

→ Ground planes	<1 ... 2 $\Omega$
→ Vcc distribution	10 ... 20 $\Omega$
→ Video- /Clock- /Data line	50 ... 90 $\Omega$
→ long data lines	90 ... >150 $\Omega$

# HOW TO CALCULATE THE RIGHT CHIP BEAD FERRITE?



Example: power supply

(1) Required insertion loss of ferrite: 22dB @ 200 MHz

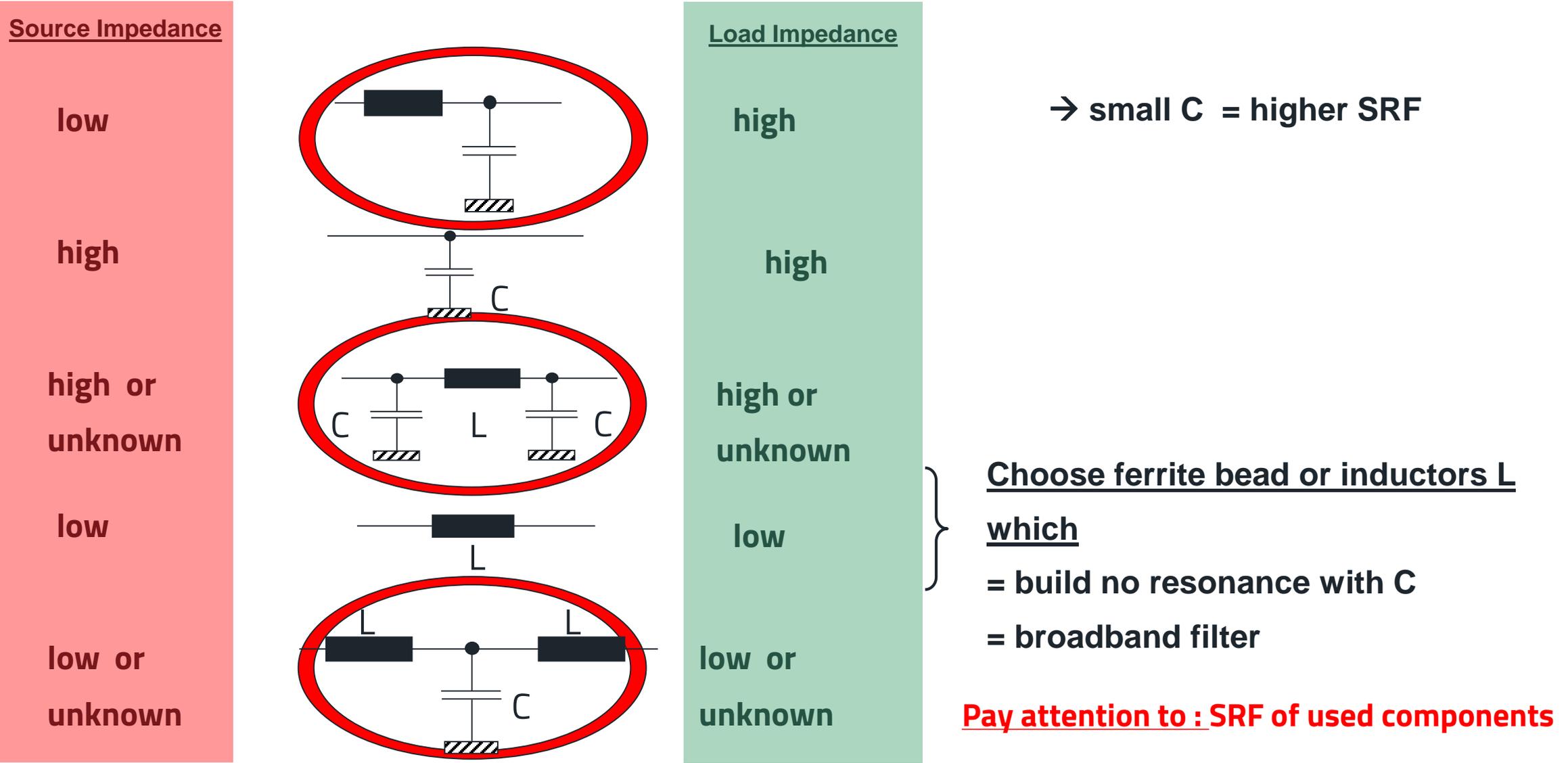
(2) System impedance for power supplies:  $Z < 10 \Omega$

(3)  $Z_{\text{ferrite}} = 220 \Omega$

(4) 742792022

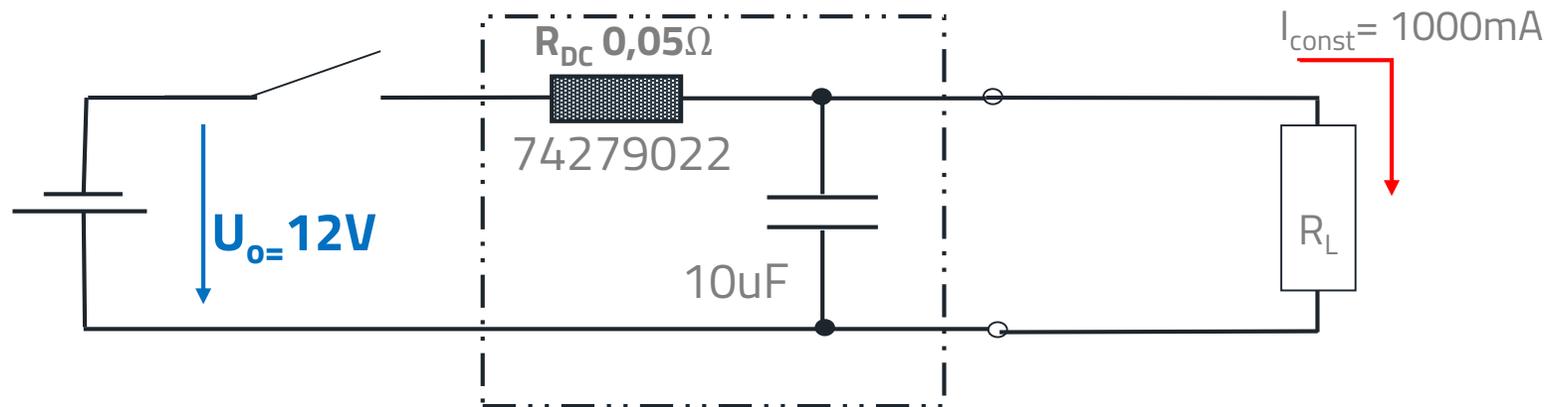


# INSERTION LOSS – RECOMMENDED FILTER TOPOLOGY



## CHIP BEAD FERRITE – PEAK CURRENT BEHAVIOR

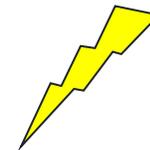
Ferrite is destroyed due to over current/in-rush current



$$I_o = U_o / (R_{DC \text{ ferrite}} + R_{ESR \text{ capacity}})$$

$$= 12V / (0.05\Omega + 0.5\Omega) = 22A$$

→ 11 times higher current



**Ferrite can be destroyed, might not fail directly => "creeping process"**

at 22A...you can smell it!

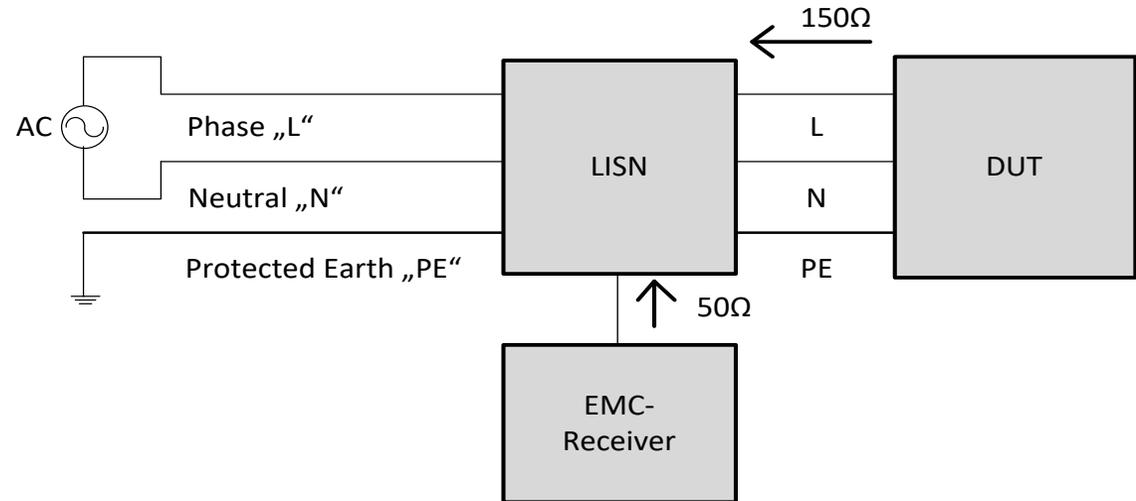


## SOLUTION: WE-MPSB

# PRACTICAL PART - CONDUCTED EMISSIONS MEASUREMENT

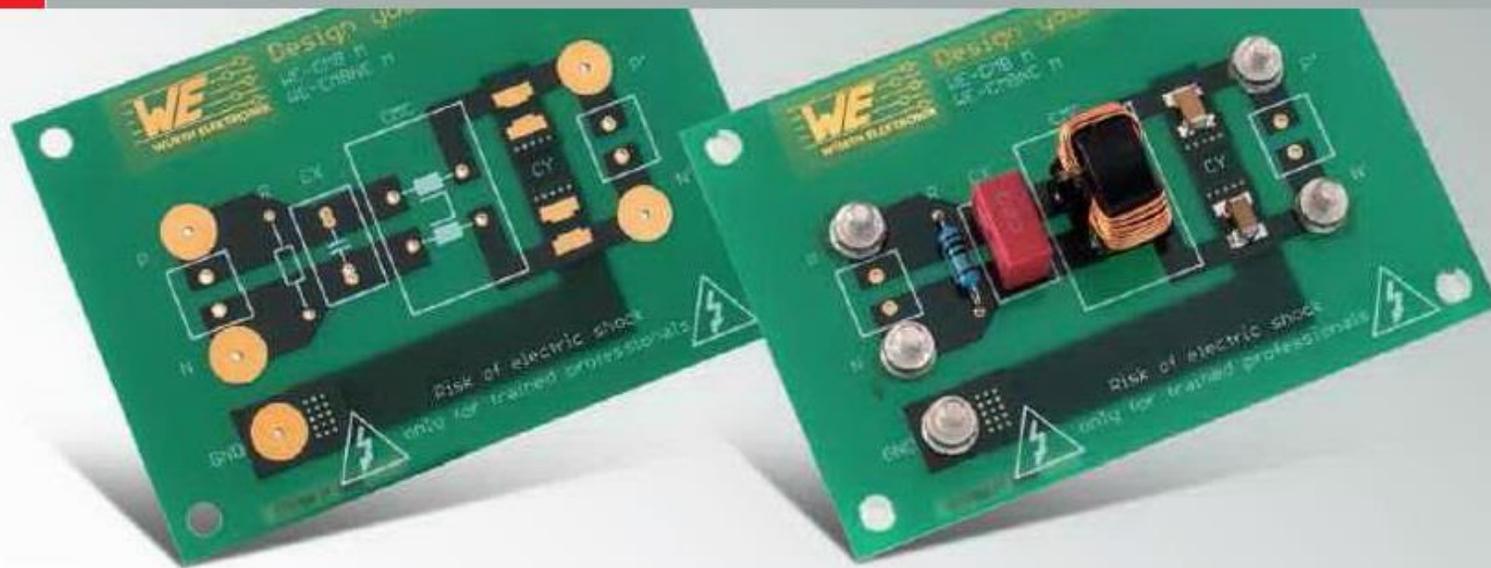
## Test setup

- LISN: „Line Impedance Stabilization Network“
  - Create known impedance on power lines for DUT
  - Filter mains voltage and cut higher frequency
  - Transfer conducted emission noise to EMC-Receiver
- EMC-chamber is recommended but not required



# DESIGN KIT

## DESIGN KIT Design your EMC Filter



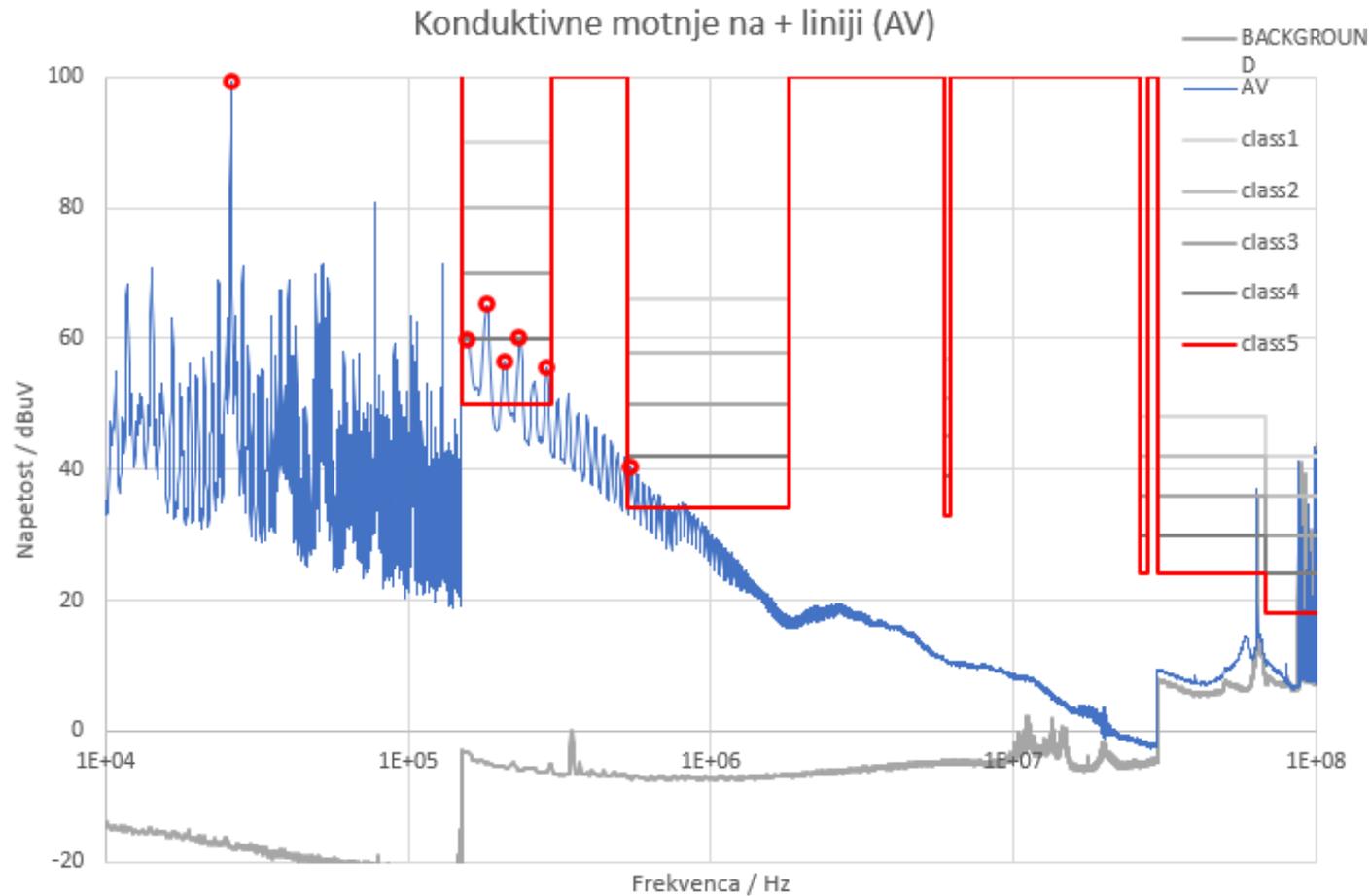
### CONTENT:

- Common Mode Chokes
- Capacitors
- Terminal Blocks
- SMD Power Elements
- Resistors
- Test Boards
- Self-Retaining Spacer

Order Code 744 998  
Version 2.0



# MEASUREMENT



# REDEXPERT®: WHAT IS IT?

Online platform for easy component selection

Select | Compare | Simulate

- Live updates
- Easy component comparison
- Several design tools integrated
- Share results and searches easily online

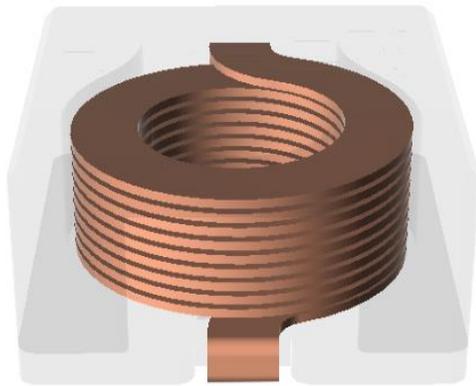


**WE**  
WÜRTH ELEKTRONIK

**REDEXPERT**  
No Calculation  
Smart Determination

# WORLD'S MOST ACCURATE AC LOSS MODEL

$$P_{\text{total}} = P_{\text{dc}} + P_{\text{ac}} = P_{\text{Cu,dc}} + (P_{\text{Cu,ac}} + P_{\text{core,ac}})$$



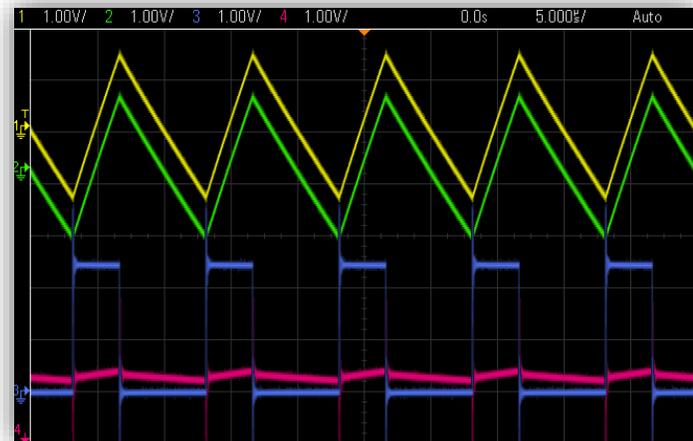
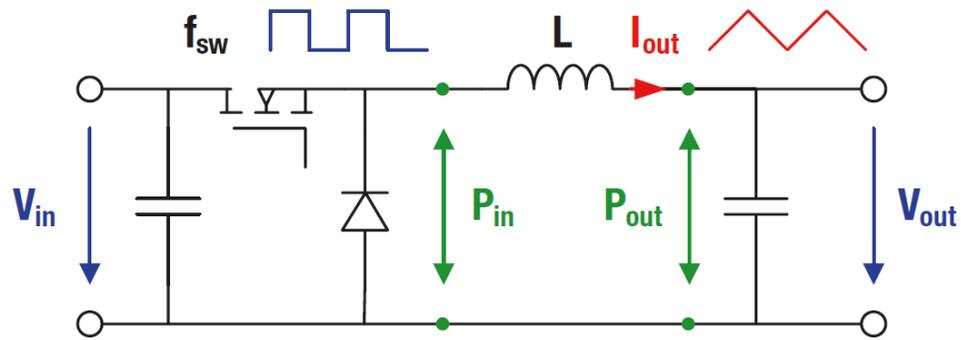
- Winding losses:
  - DC losses
  - AC losses (skin effect, proximity effect)



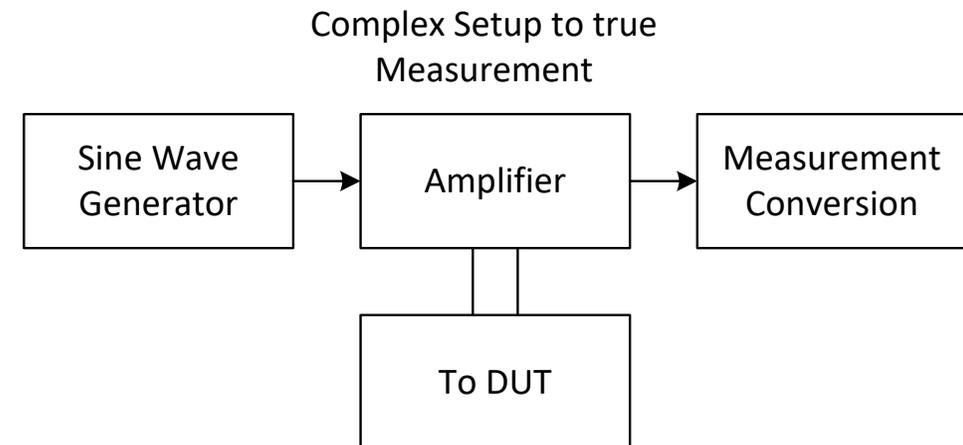
- Core losses:
  - Hysteresis losses
  - Eddy current losses

# WORLD'S MOST ACCURATE AC LOSS MODEL

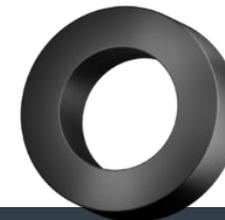
## Würth model



## Steinmetz model

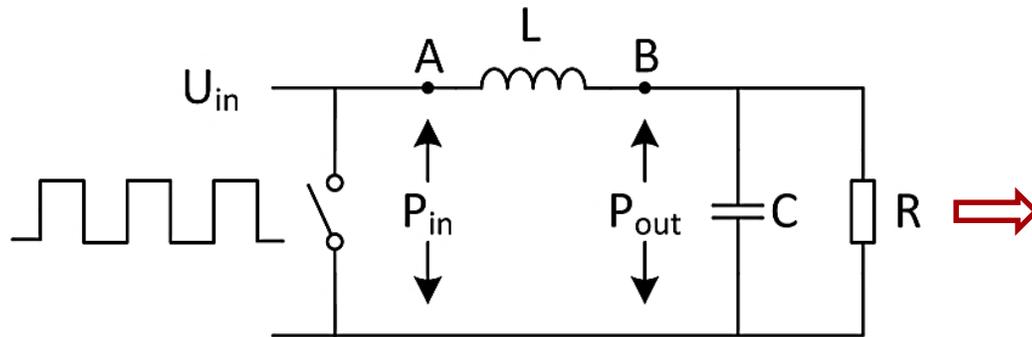


Limitation on Permeability



# WÜRTH ELEKTRONIK AC LOSS MODEL

- Description & Set up:



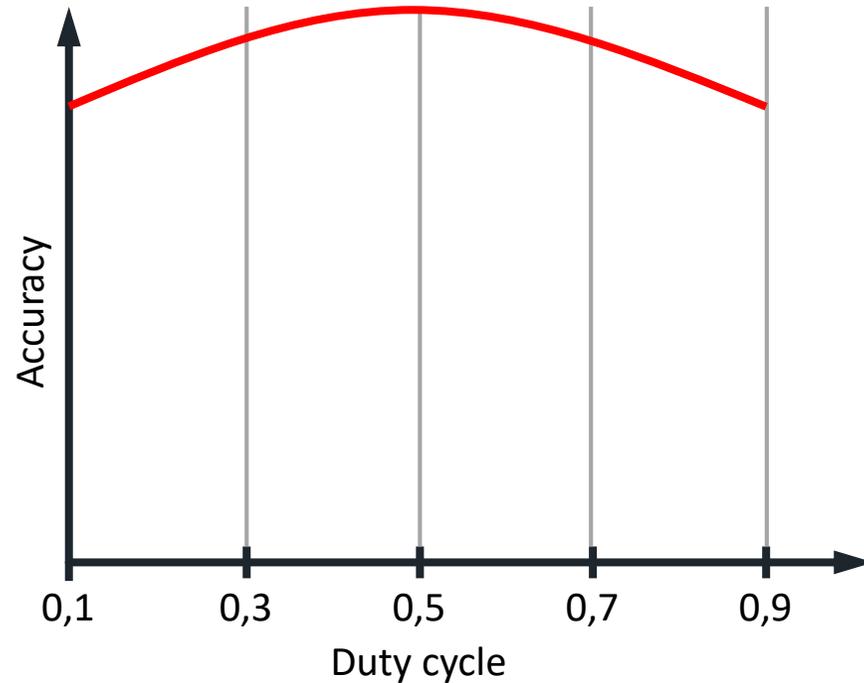
- A pulsating input voltage is applied over the Inductor
- $(P_{in} - P_{out})$  is the power loss in the Inductor

Fig 7 & 8 : Practical DC-DC converter set-up & resulting scope shots

- A pulsating input voltage is applied over the Inductor
- $(P_{in} - P_{out})$  is the power loss in the Inductor
- Separate losses of the Inductor into AC & DC loss

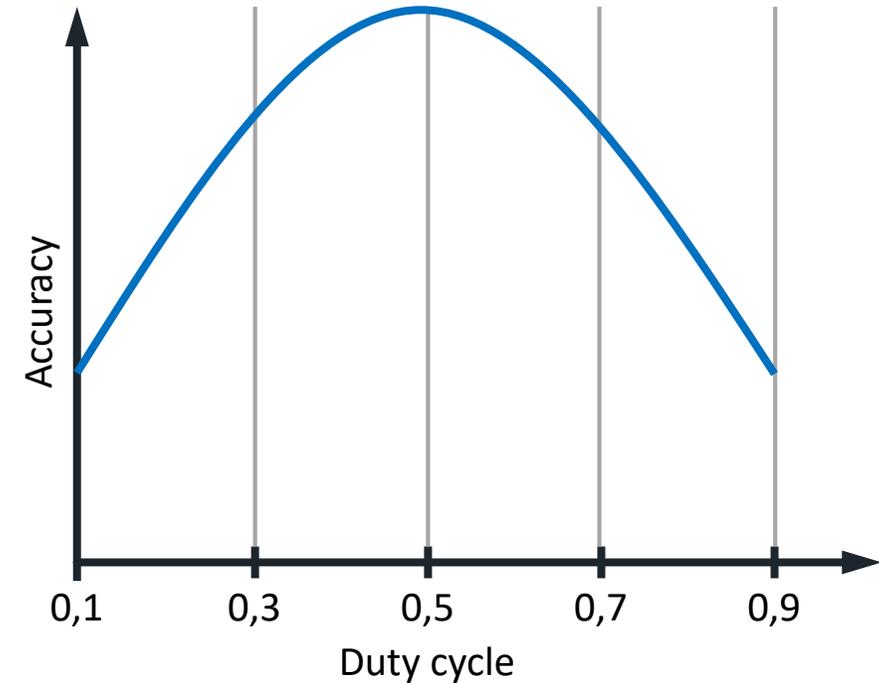
# WORLD'S MOST ACCURATE AC LOSS MODEL

**Würth model**



- Highest accuracy over wide d.c. range

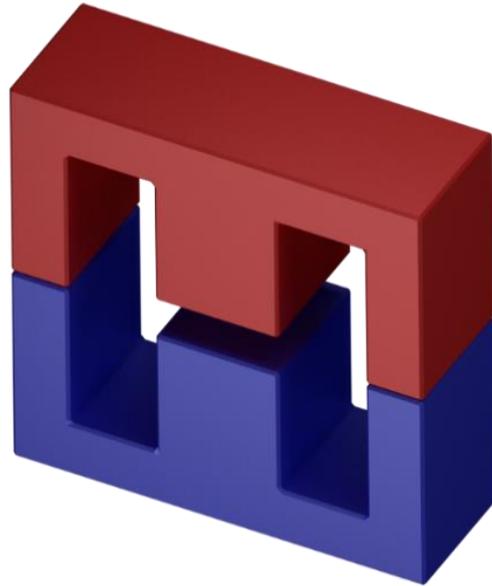
**Steinmetz model**



- Acceptable accuracy for ring cores at 50%, worse for other d.c.

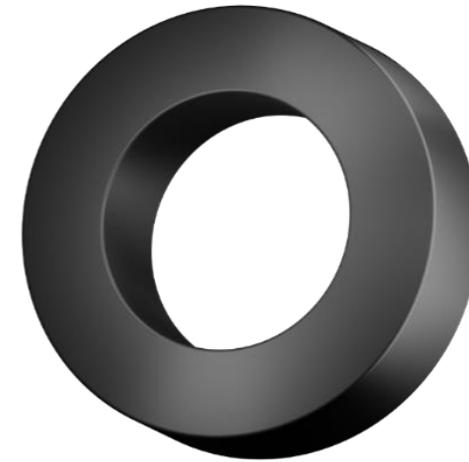
# WORLD'S MOST ACCURATE AC LOSS MODEL

**Würth model**



- Material combination supported, i.e. NiZn, MnZn, iron powder, metal alloy, etc.

**Steinmetz model**



- Only single material, mainly for NiZn, MnZn
- Not applicable for iron powder & metal alloy

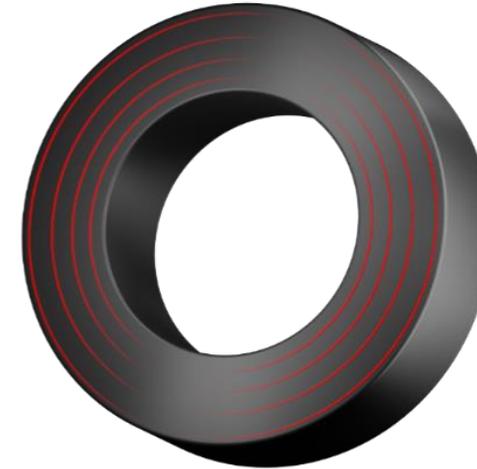
# WORLD'S MOST ACCURATE AC LOSS MODEL

## Würth model



- Consideration of:
  - Real core shapes
  - Winding structure
  - Winding losses
- Losses due to air gap (fringing effects)

## Steinmetz model



- Consideration of:
  - Toroidal cores without an air gap

# START SCREEN



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Product table

Order Code	Series	Size	L <sub>0</sub>	R <sub>DC,typ</sub>	I <sub>R</sub>	I <sub>sat</sub>	L	W	H <sub>Max</sub>	T <sub>Op</sub>	Shielded	Q+ / AEC-Q	Material	Assy
74438343022	WE-MAPI	2010	2.20 µH	225 mΩ	1.10 A	2.50 A	2.00 mm	1.60 mm	1.00 mm	125°C	Shielded		Metal Alloy	SMT
744383210047	WE-MAPI	2506	470 nH	76.0 mΩ	2.20 A	3.70 A	2.50 mm	2.00 mm	0.600 mm	125°C	Shielded		Metal Alloy	SMT
74438313015	WE-MAPI	1610	1.50 µH	189 mΩ	950 mA	2.70 A	1.60 mm	1.60 mm	1.00 mm	125°C	Shielded		Metal Alloy	SMT
74438313022	WE-MAPI	1610	2.20 µH	337 mΩ	850 mA	2.50 A	1.60 mm	1.60 mm	1.00 mm	125°C	Shielded		Metal Alloy	SMT

Product tray

74438343022 WE-MAPI - 2010 2.20 µH - 225 mΩ

744383210047 WE-MAPI - 2506 470 nH - 76.0 mΩ

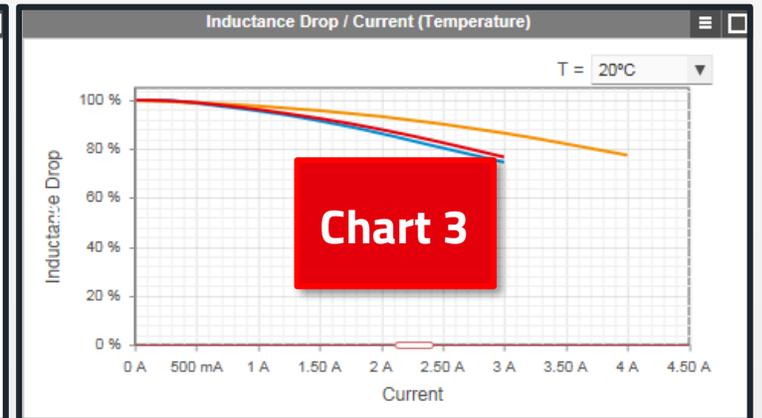
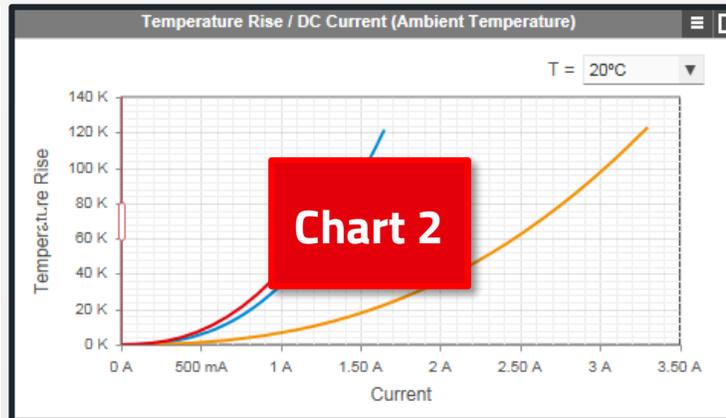
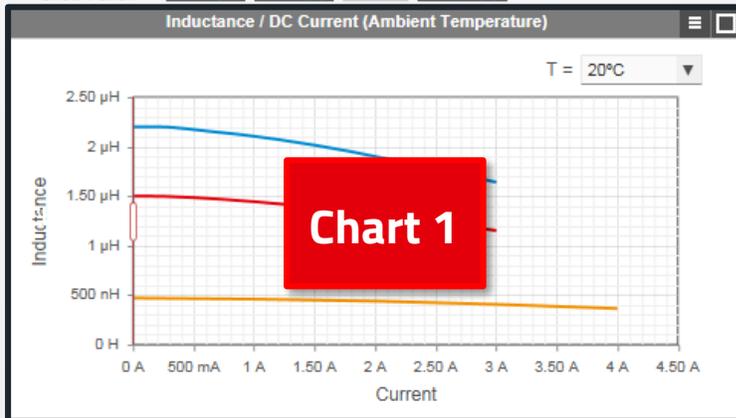
74438313015 WE-MAPI - 1610 1.50 µH - 189 mΩ

Drop Order Codes in the tray to add

Add to Cart

More...

Show Panel: L vs. I(T) | K vs. I(T) | Z vs. F | LD vs. I(T)



# PRODUCT TABLE

## Filter summary line:

- Updates are highlighted red for a second



Filters: Type = Single ✕  $I_R \geq 2.00 \text{ A}$  ✕  $I_{sat} \geq 2.40 \text{ A}$  ✕  $12.7 \mu\text{H} \leq L_{20.0^\circ\text{C}@2.22 \text{ A}} \leq 23.6 \mu\text{H}$  ✕ URL 71 items

Order Code	Series	Size	Sp...	Type	$L_0$	$L_{20.0^\circ\text{C}@2.22 \text{ A}}$	$R_{DC,typ}$	$I_R$	$I_{sat}$	L	W
7443551151	WE-HCI	1365	PDF	Single	15.4 $\mu\text{H}$	14.2 $\mu\text{H}$	14.0 m $\Omega$	9.00 A	8.00 A	13.2 mm	12.8 mm
7443551181	WE-HCI	1365	PDF	Single	18.0 $\mu\text{H}$	16.5 $\mu\text{H}$	22.0 m $\Omega$	7.50 A	7.50 A	13.2 mm	12.8 mm
7443641500	WE-HCF	2818	PDF	Single	15.0 $\mu\text{H}$	14.3 $\mu\text{H}$	2.40 m $\Omega$	30.0 A	26.0 A	28.0 mm	28.5 mm
7443551181	WE-HCI	1365	PDF	Single	18.0 $\mu\text{H}$	16.5 $\mu\text{H}$	22.0 m $\Omega$	7.50 A	7.50 A	13.2 mm	12.8 mm

## Filters are set by:

- Drop-down menu
- Design tools
- Chart marker

- Sort & filter each column
- Live application of changes

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