

Switched Mode Power Supply with high efficiency and best EMI design



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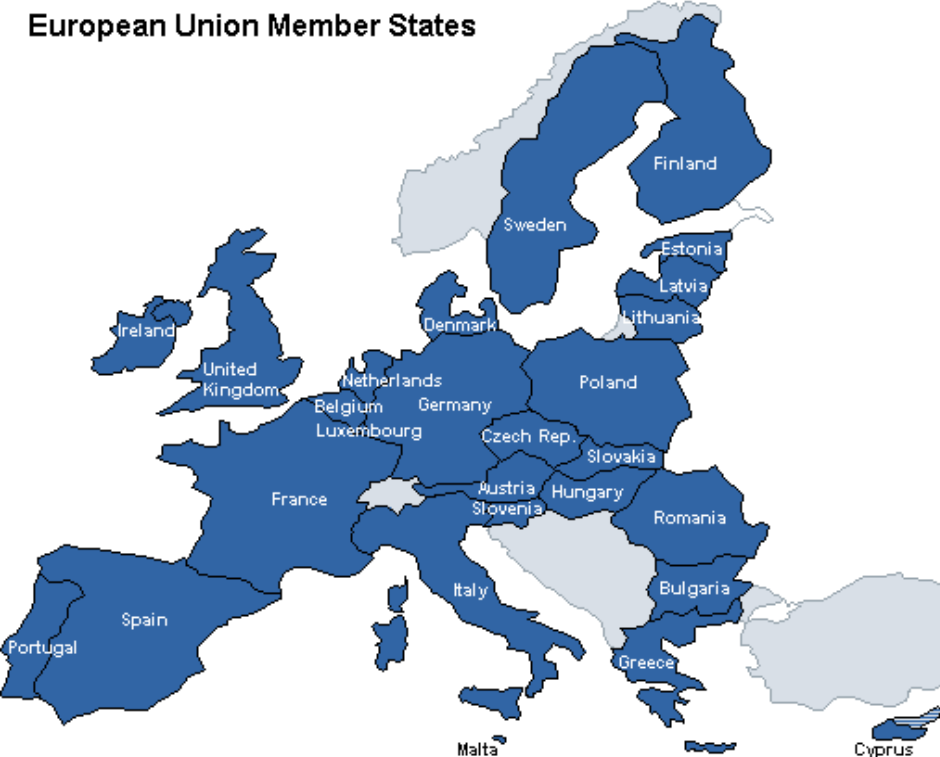
REQUIREMENTS IN EMC

EMC - Standards

- EN 61000-3-2 Limits for harmonic current emissions (equipment input current up to and including 16 A per phase)
- EN 61000-3-3 Limitation of voltage changes, voltage fluctuations and flicker in public low-voltage supply systems
- EN 55011 ISM Equipment (**I**ndustrial, **S**cientific and **M**edical) also known as CISPR-11
- EN 55013 Audio and Broadcast receiver equipment
- EN 55014-1 House hold appliances, electric tools and similar apparatus
- EN 55015 Limits and methods of measurement of radio disturbance characteristics of electrical lighting and similar equipment
- EN 55022 ITE (**I**nformation **T**echnology **E**quipment), also known as CISPR-22
- EN 61000-6-1 Generic immunity standard for residential, commercial and light industry environments
- EN 61000-6-2 Generic immunity standard for industrial environments
- EN 61000-6-3 Generic emission standard for residential, commercial and light industry environments
- EN 61000-6-4 Generic emission standard for industrial environments
- EN 61000-4-2 Electrostatic discharge immunity test (ESD)

CE Marking

- With the formation of the single European market, standardization was required to remove technical barriers to trade.
- New Approach Directives were introduced to remove these barriers to trade
- 22 New Approach Directives
 - Electro Magnetic Compatibility (EMC)
 - Low Voltage Directive (LVD)
 - Medical Devices Directive (MDD)



What is the meaning of EMC ?



What's all the fuss about EMC?

- In Europe, we have a mechanism called CE Marking
- It is applicable to any electrical/electronic product
- **EMC Directive** , regulation to ensure that intentional RF transmission signals are not interfered with
- Ensures that Electrical/Electronic devices continue to operate as intended in a Electro Magnetic Environment
- **Failure to comply with the law can be an offence, either criminal, civil or both**



Other International EMC approval marks



- **Federal Communications Commission**



- **Voluntary Control Council for Interference**



- **Australian Communications and Media Authority**



Conducted Emission

- Conducted emission over wideband
- Caused by ripple current at input lines (common mode - / differential mode noise)
- EMC requirements for „*Conducted Emission*“ according ETSI, CEN, CENELEC
- E.g.: EN 55013 : 2006 (Radio & TV broadcast receivers and associated equipment)

66 - 56dB μ V @ 150<KHz<500KHz (QP)

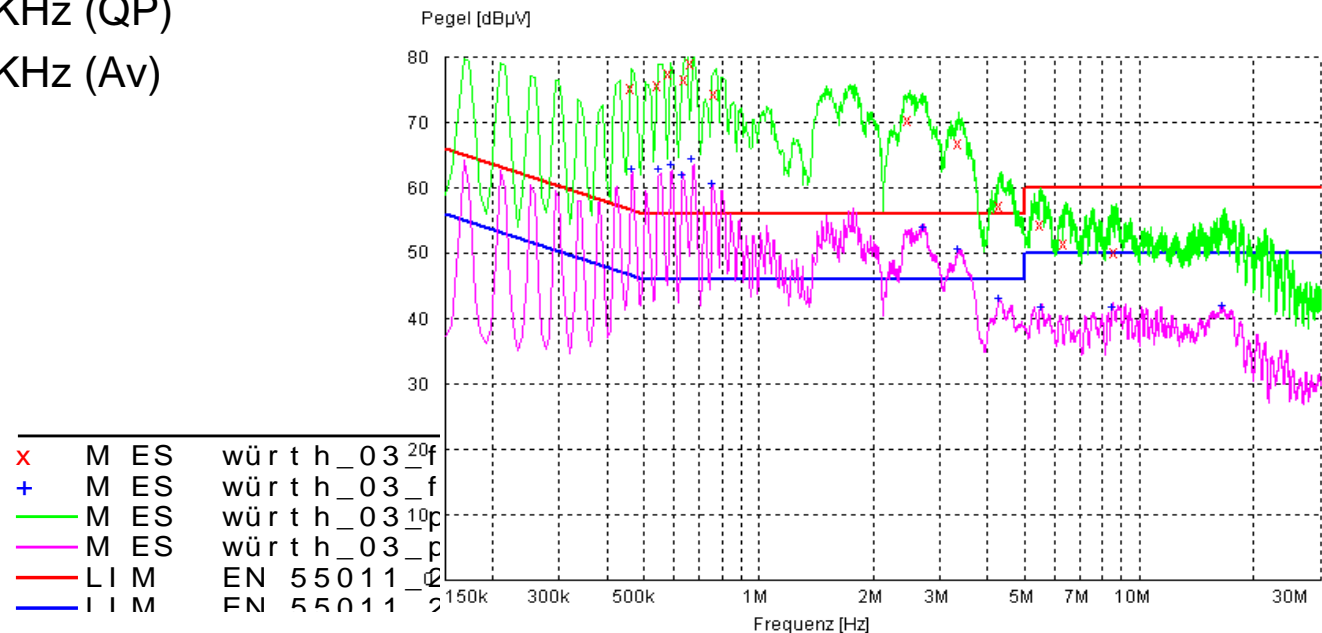
56 - 46dB μ V @ 150<KHz<500KHz (Av)

56dB μ V @ 0,5<MHz<5 (QP)

46dB μ V @ 0,5<MHz<5 (Av)

60dB μ V @ 5<MHz<30 (QP)

50dB μ V @ 5<MHz<30 (Av)



Radiated Emission

- Radiated emission over wideband
- Caused by:
 - Power traces on PCB
 - Power choke of DC/DC converter
- EMC requirements for „*Radiated Emission*“ according ETSI, CEN, CENELEC

- **EN 61000-6-3** : 2007 (Home)

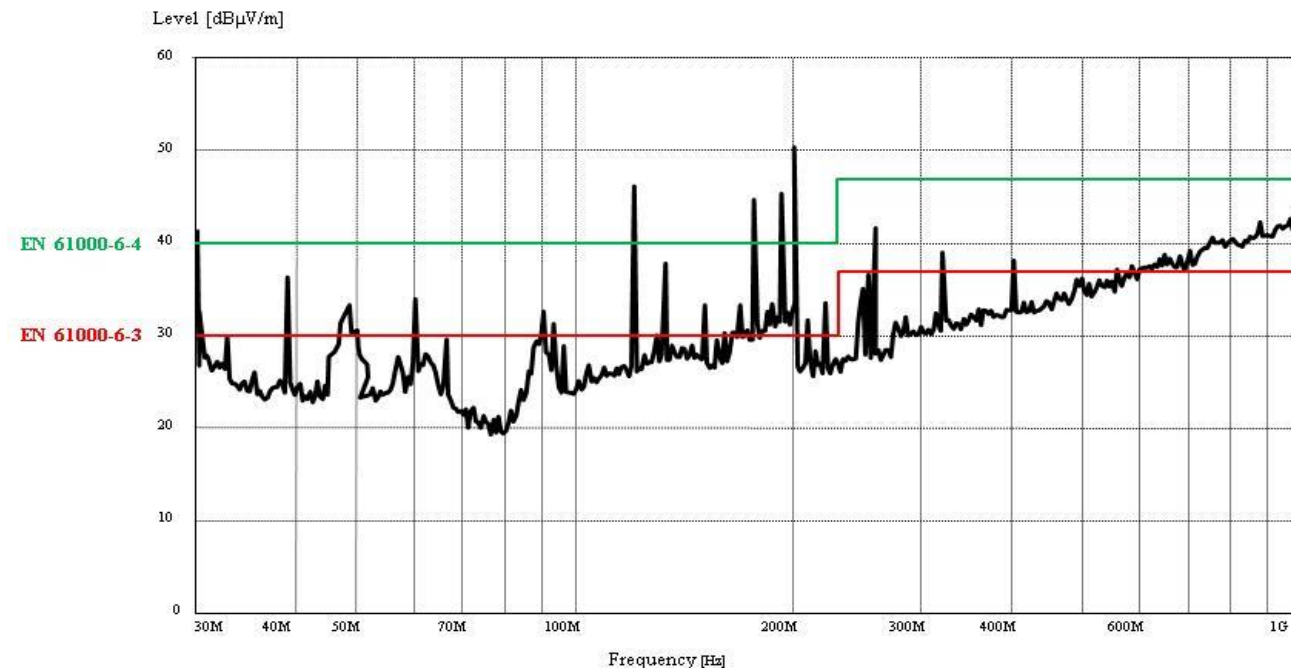
30dB @ 30MHz~230MHz $\mu\text{V}/\text{m}$

37dB @ 230MHz~1GHz $\mu\text{V}/\text{m}$

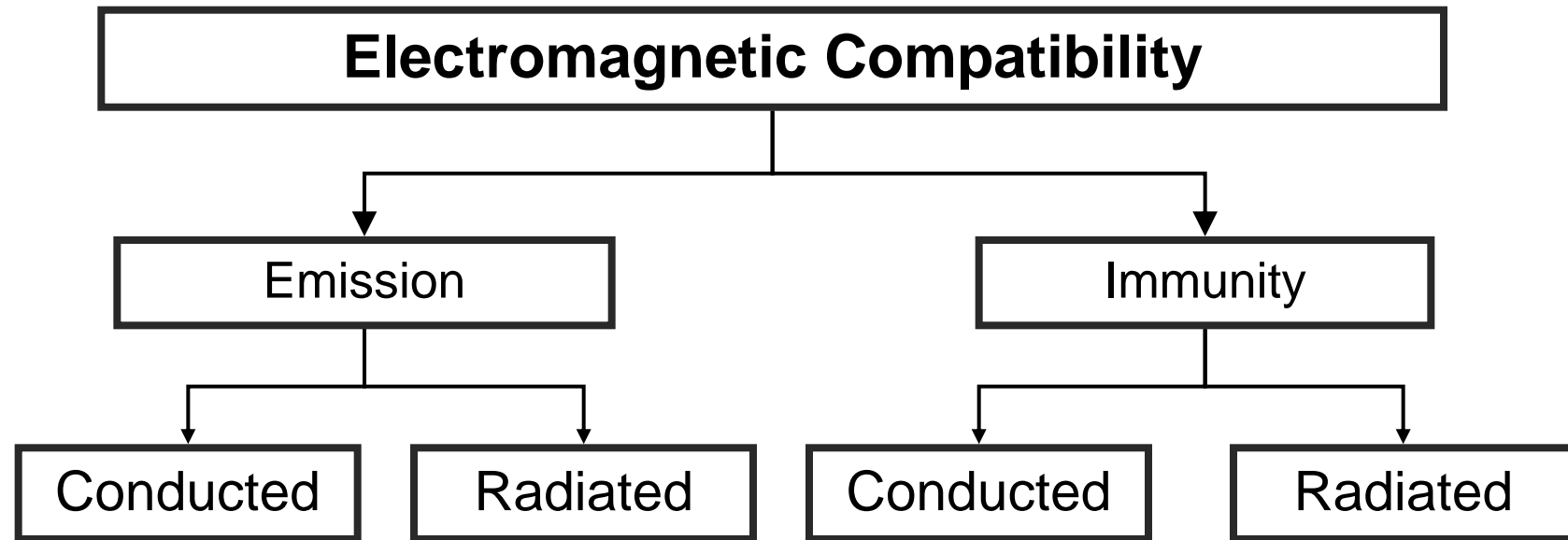
- **EN 61000-6-4** : 2007 (Industrial)

40dB @ 30MHz~230MHz $\mu\text{V}/\text{m}$

47dB @ 230MHz~1GHz $\mu\text{V}/\text{M}$



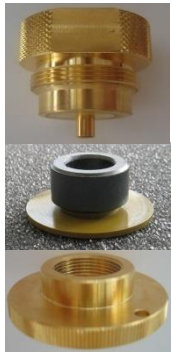
EMC – Basic Phenomena



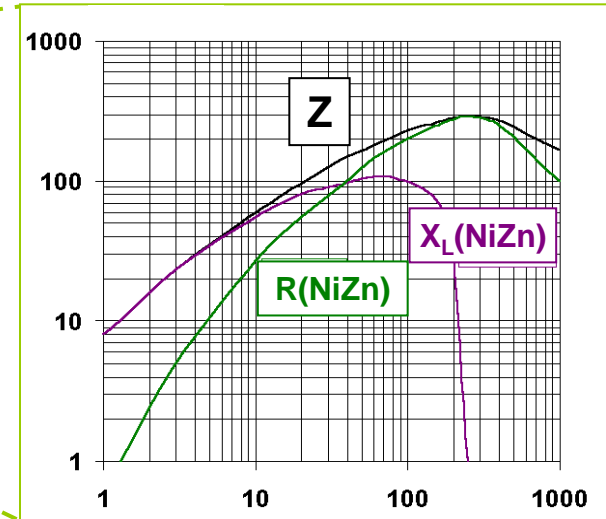
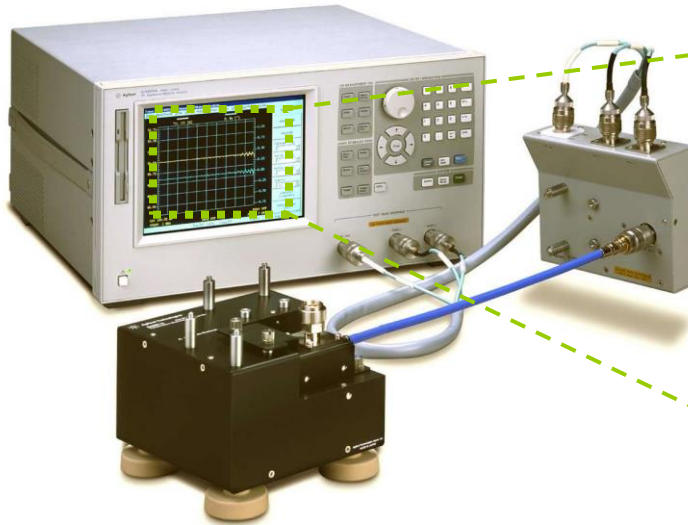


Magnetic and Material Basics

Permeability – complex permeability

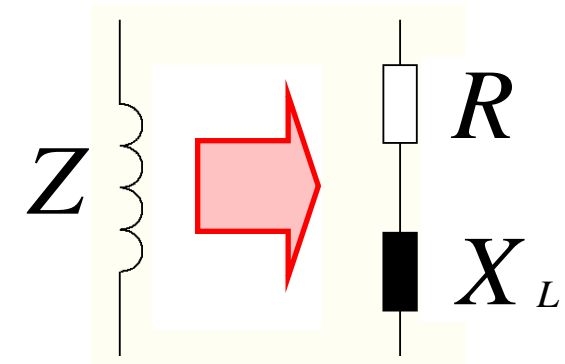


=1
turn



Core material-Parameter

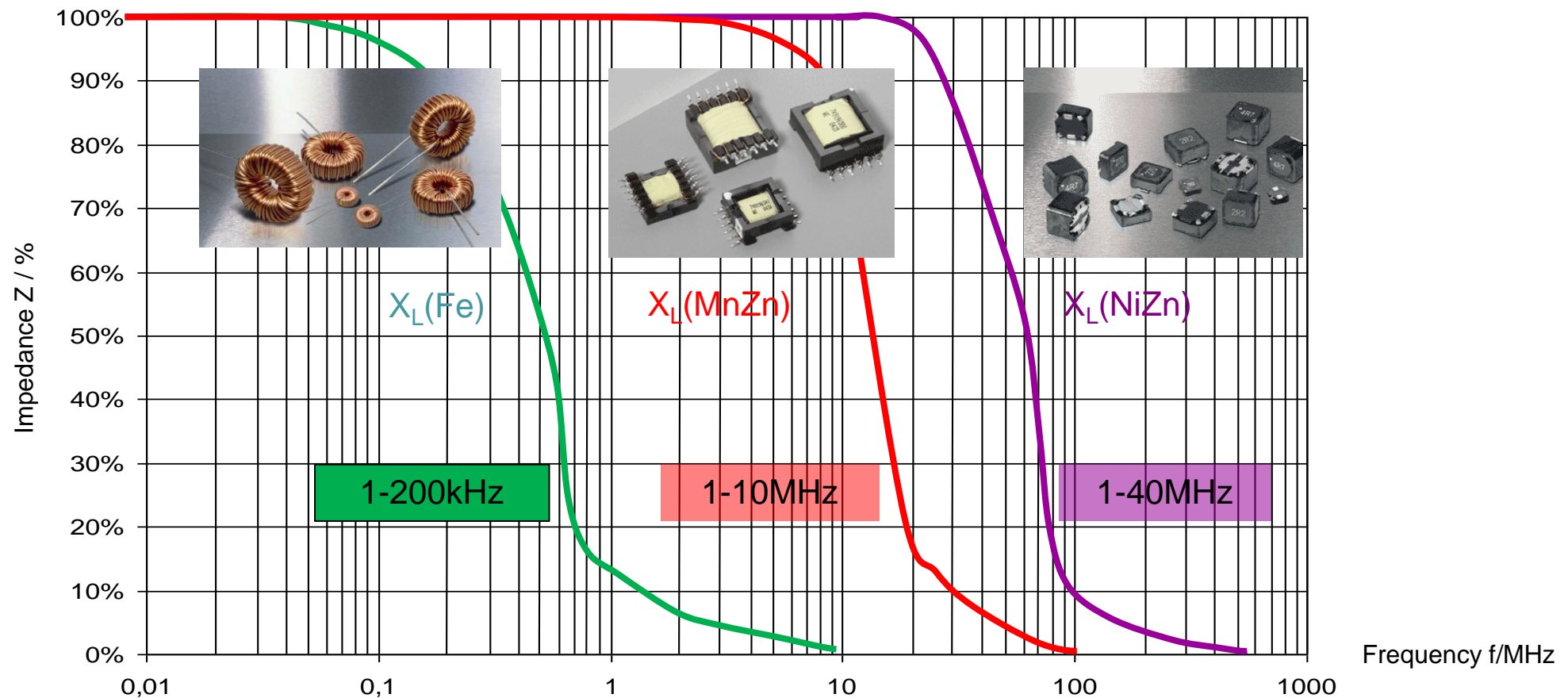
Replacement circuit



$$Z = \sqrt{R^2 + X_L^2}$$

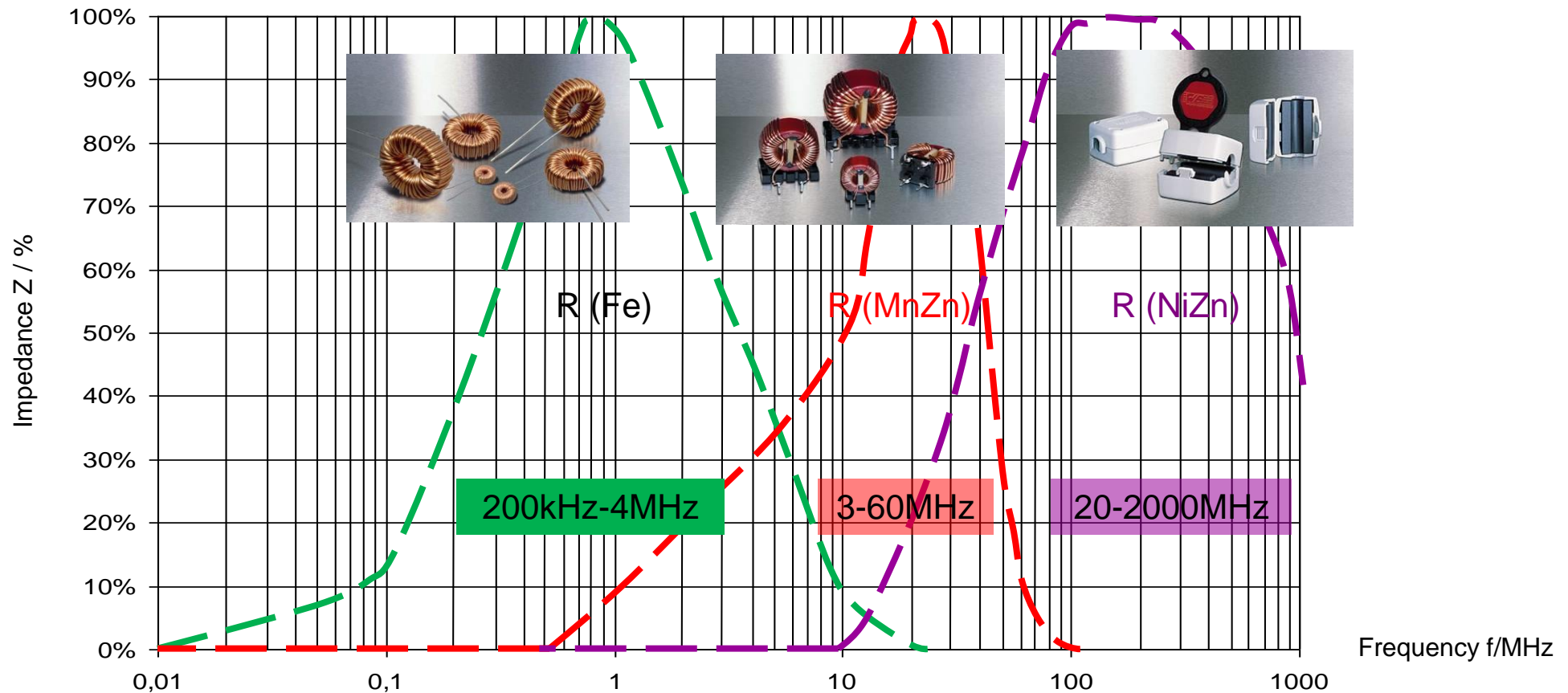
Core materials - Inductors (Energy storage)

Which switching frequency do you use?



Core materials- Chokes (filtering)

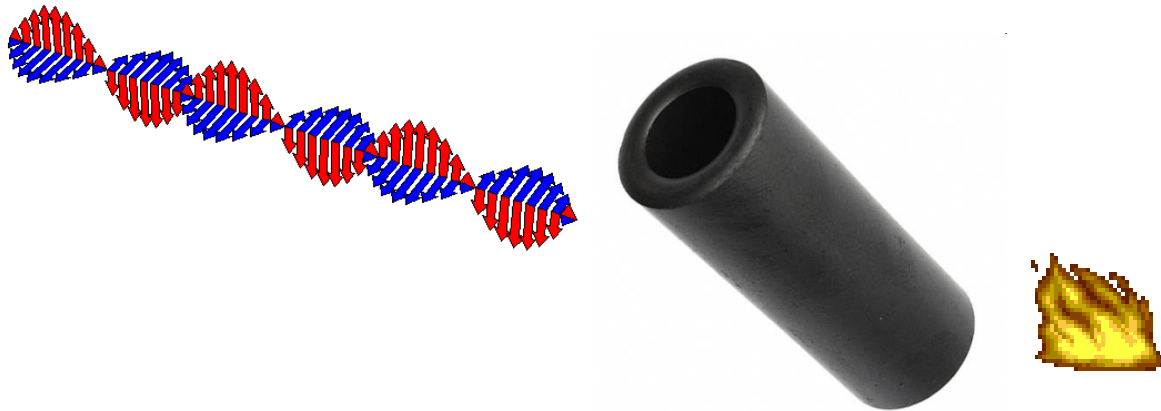
Noise frequency range must be known



Core Losses

Electro Magnetic energy cannot disappear, it will be just transformed into other energy form → energy conservation law

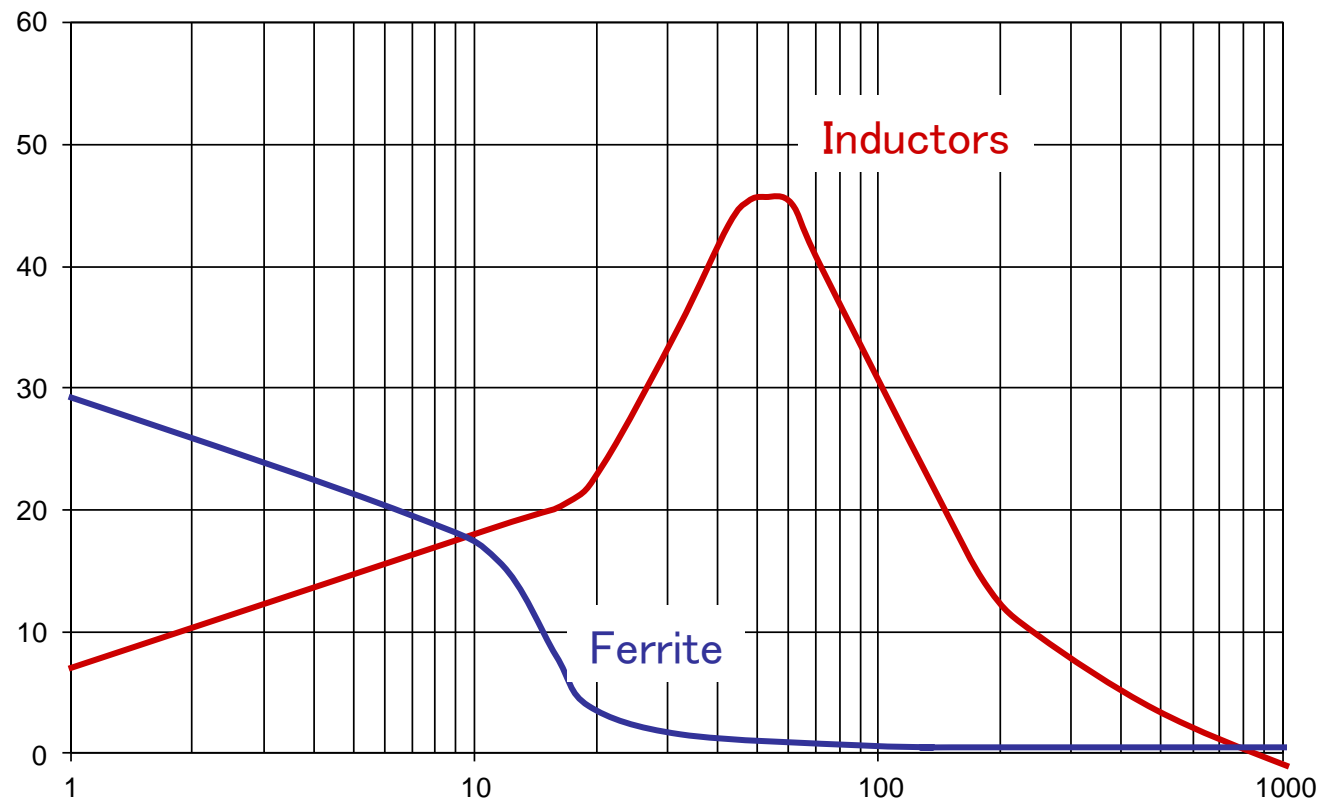
e.g. electrical energy transformed into → thermal energy



the core losses from ferrite transform the noise energy into heat

Core material (Inductor / EMC Ferrite)

■ Compare the Q



$$Q = \frac{X_L}{R}$$



Transmission Modes & Filter Topologies

EMC - Coupling

→ Primary procedure

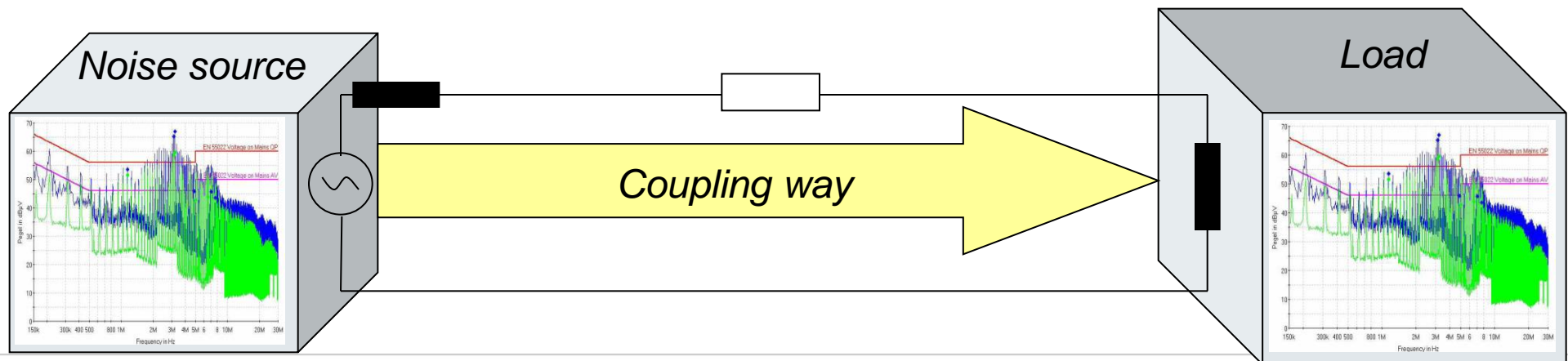
...to aim at source a low noise

→ Secondary procedure

... eliminate the noise thru interrupting the coupling way

→ Tertiary procedure

... increase the noise immunity at load



Recognizing the coupling mode

- **common mode noise ?**
- **differential mode noise ?**



Common mode or differential mode?

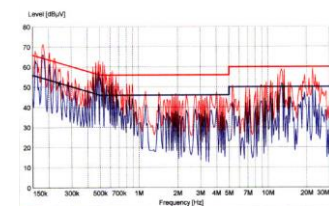
Take a Snap Ferrite and fix it on the cable
(both lines e.g. VCC and GND)

if noise is reduced or
noise immunity increase

you have Common Mode Interference

If not

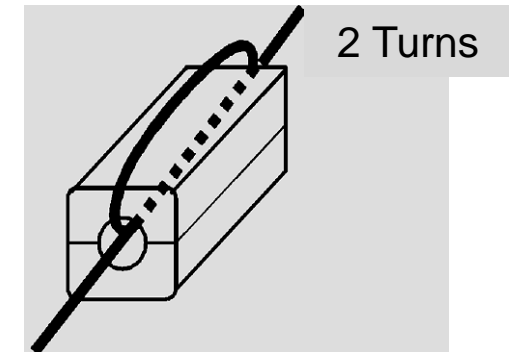
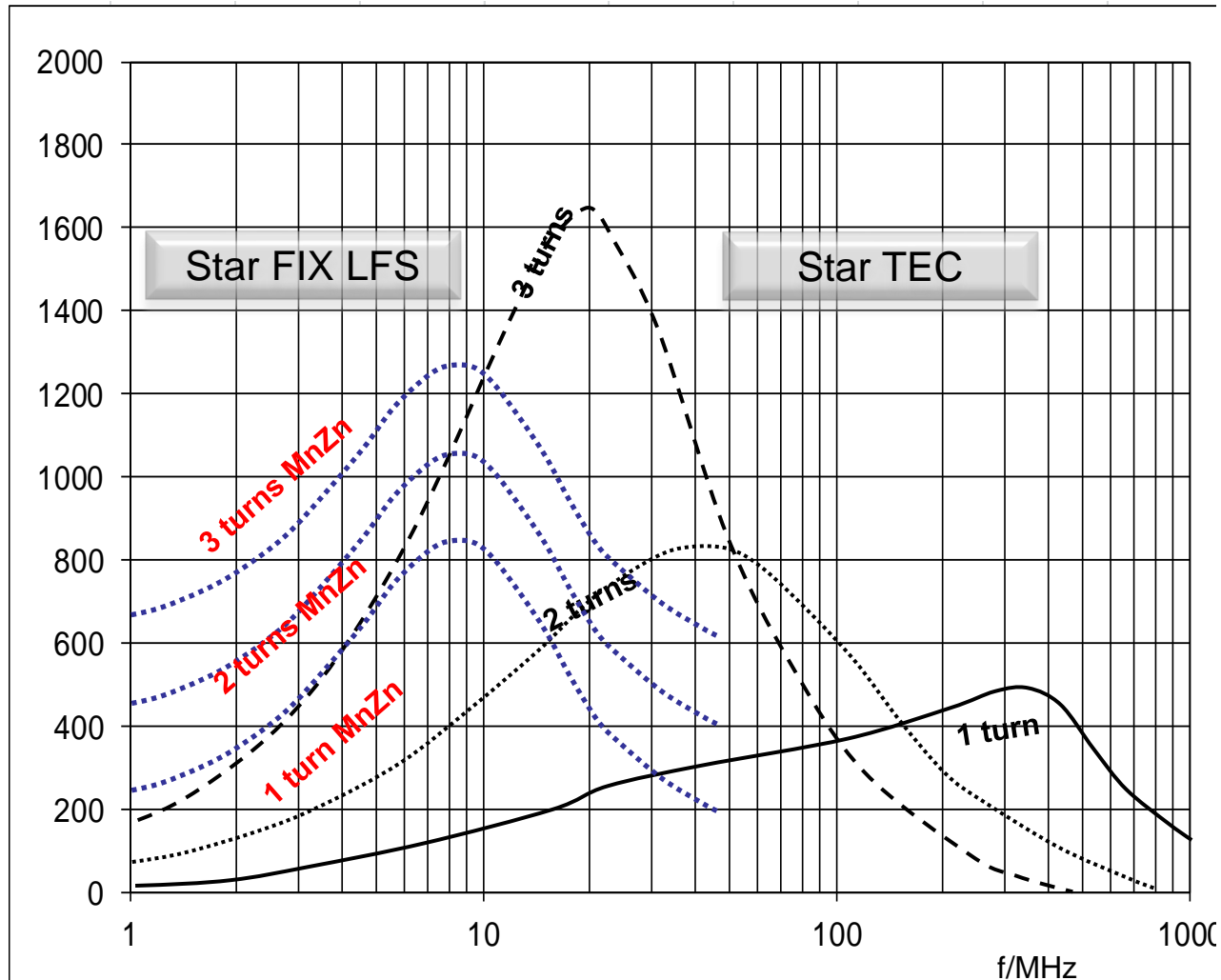
you have Differential Mode Interference



e.g. Common mode
choke

e.g. chip bead ferrite

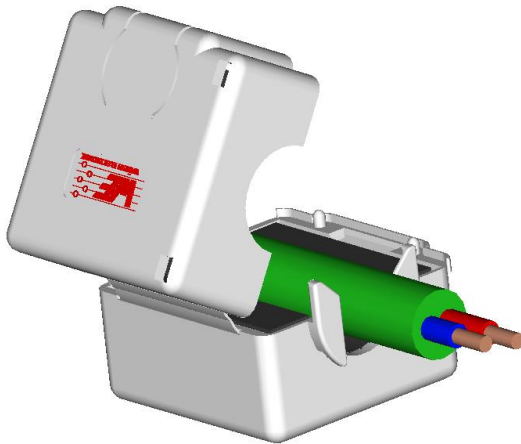
Snap on ferrite – typical behavior



Increase the no. of turns means:

Snap on ferrite - Construction

- Snap on ferrite acts as an CMC
- Absorbs common mode Interferences
- Comparable with bifilar winding CMC



Common Mode Filter – How it works

It is a Bi-directional filter

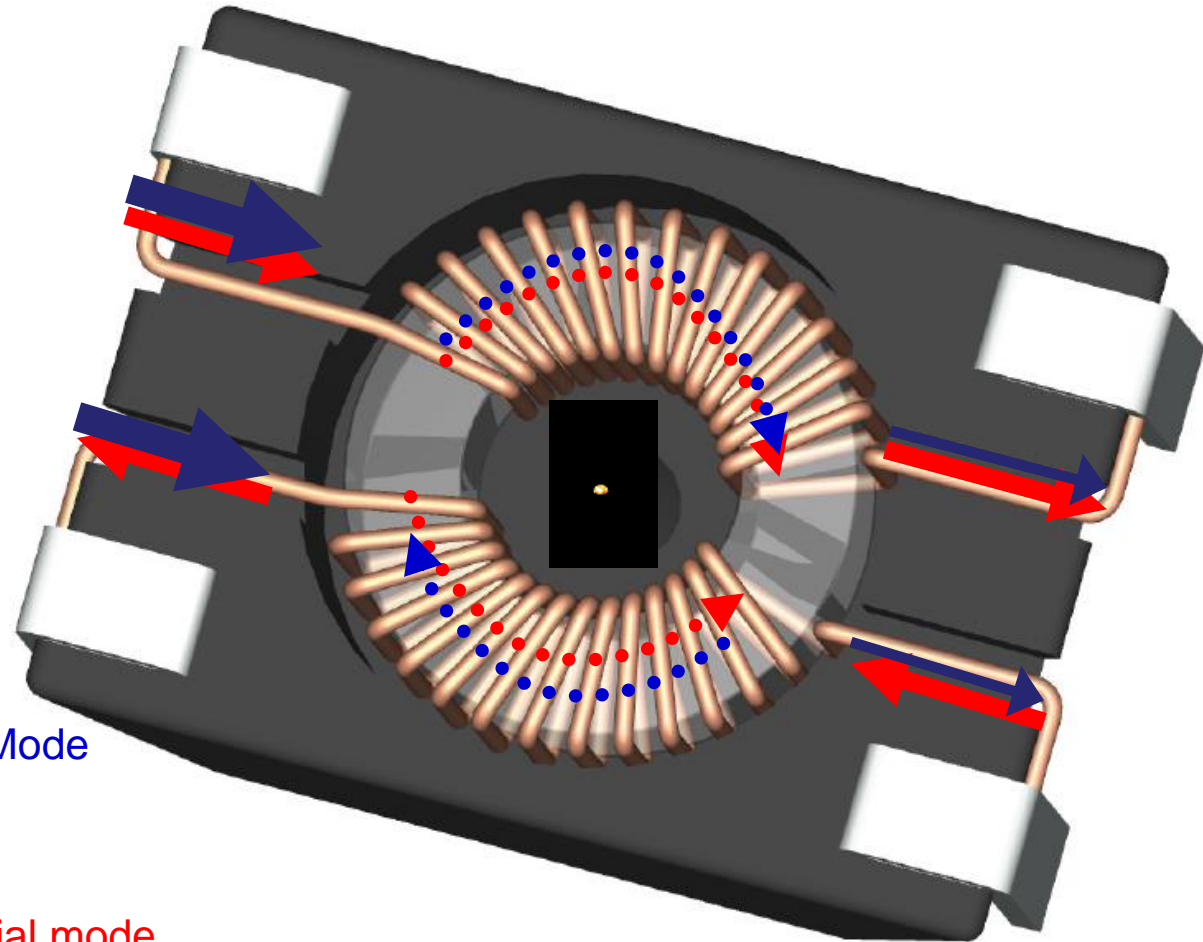
- From device to outside environment
- From outside environment to inside device

Intended Signal - **Differential mode**

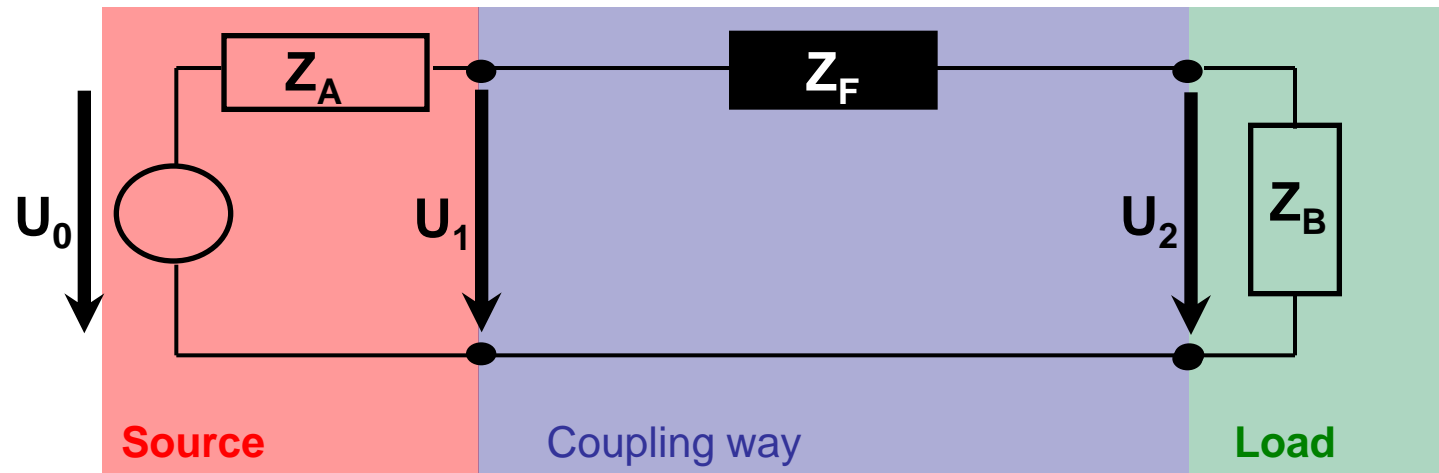
Interference Signal (noise) – **Common Mode**

Conclusion:

- “almost” no affect the signal - **Differential mode**
- high attenuation to the interference signal (noise) – **Common Mode**



Insertion loss – Mathematical Definition



- System attenuation

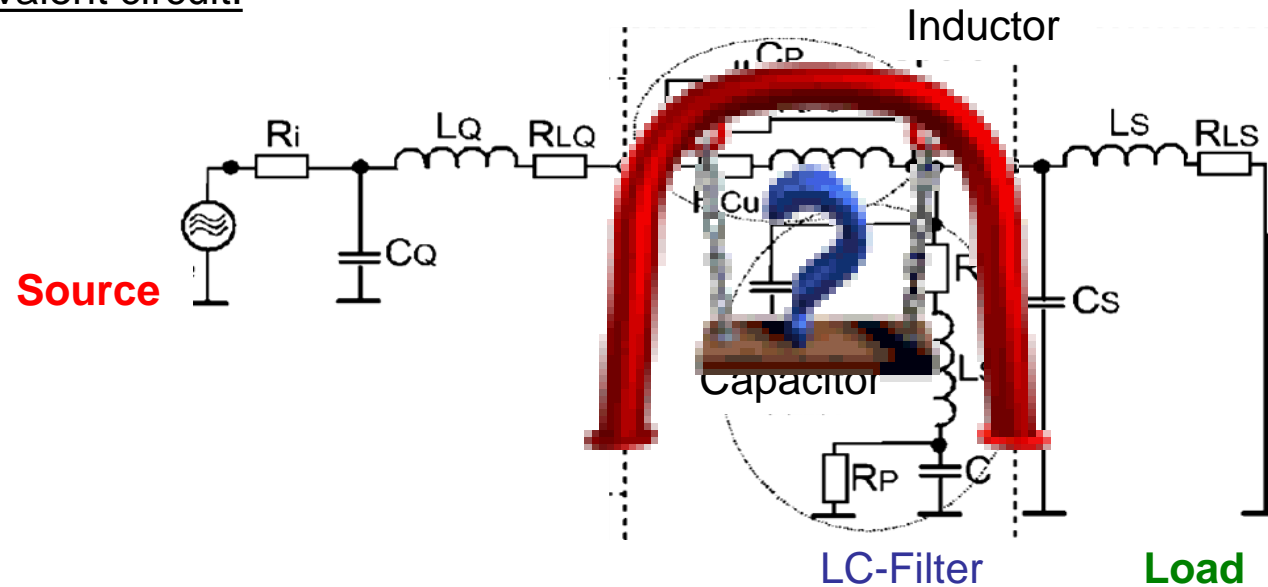
$$A = 20 \cdot \log \frac{Z_A + Z_F + Z_B}{Z_A + Z_B} \quad \text{in (dB)}$$

- Impedance

$$Z_F = \left[10^{\frac{A}{20}} \cdot (Z_A + Z_B) \right] - (Z_A + Z_B) \quad \text{in } (\Omega)$$

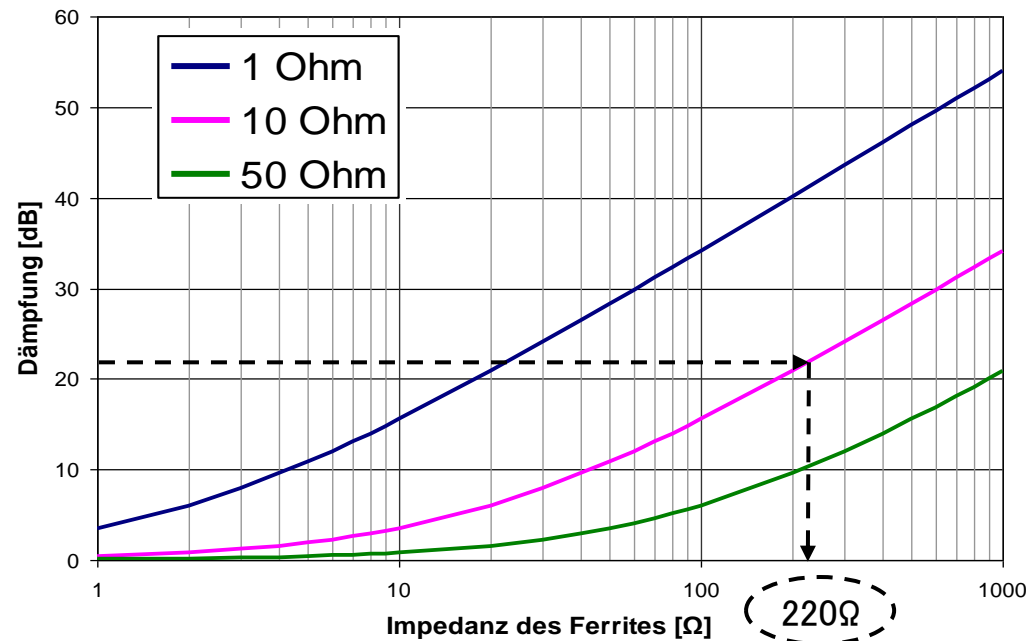
Insertion loss - Definition

Equivalent circuit:



- Practical values for source and load impedances:
 - Ground planes $<1 \dots 2 \Omega$
 - Vcc distribution $10 \dots 20 \Omega$
 - Video- /Clock- /Data line $50 \dots 90 \Omega$
 - long data lines $90 \dots >150 \Omega$

How to calculate the right chip bead ferrite?

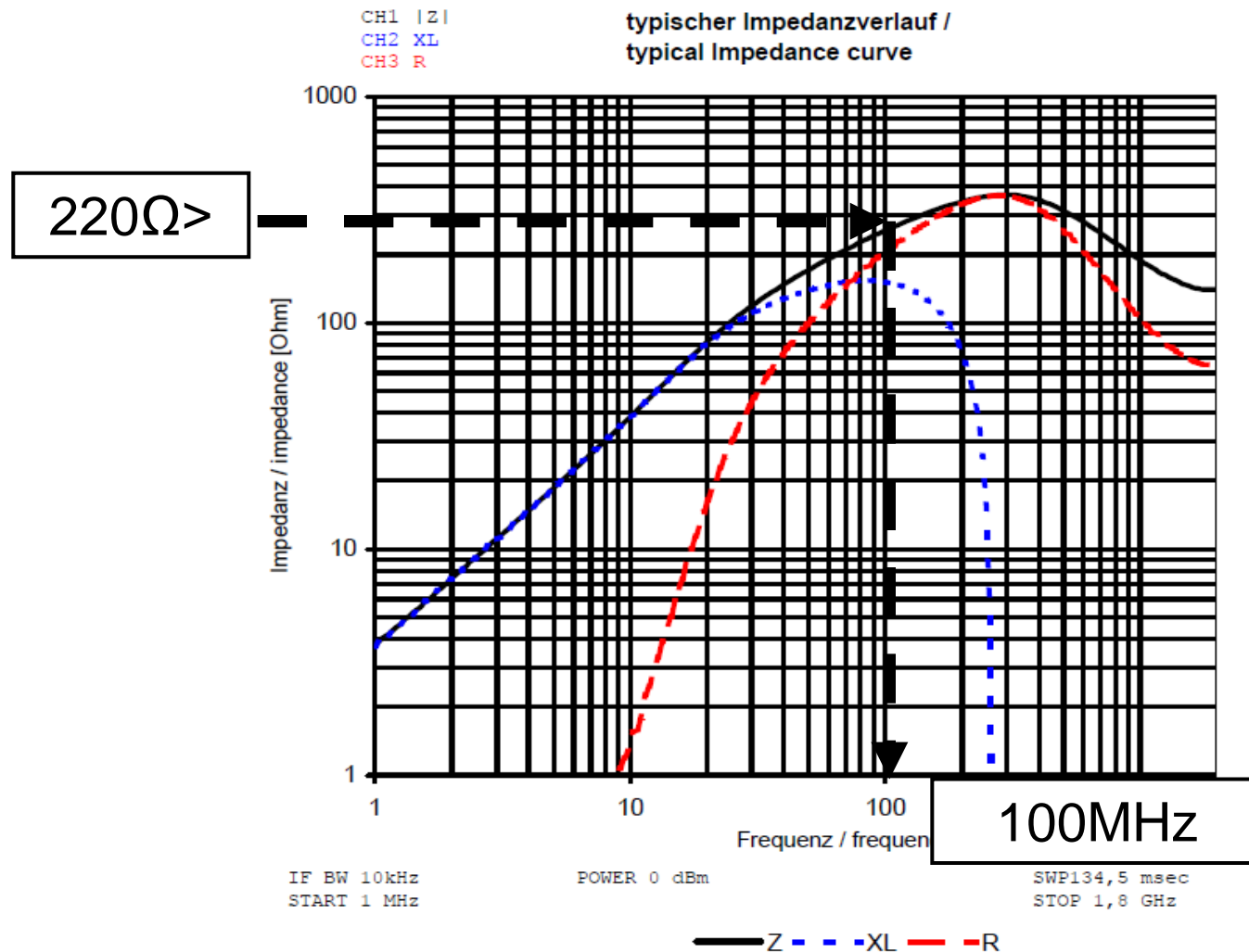


Example: power supply

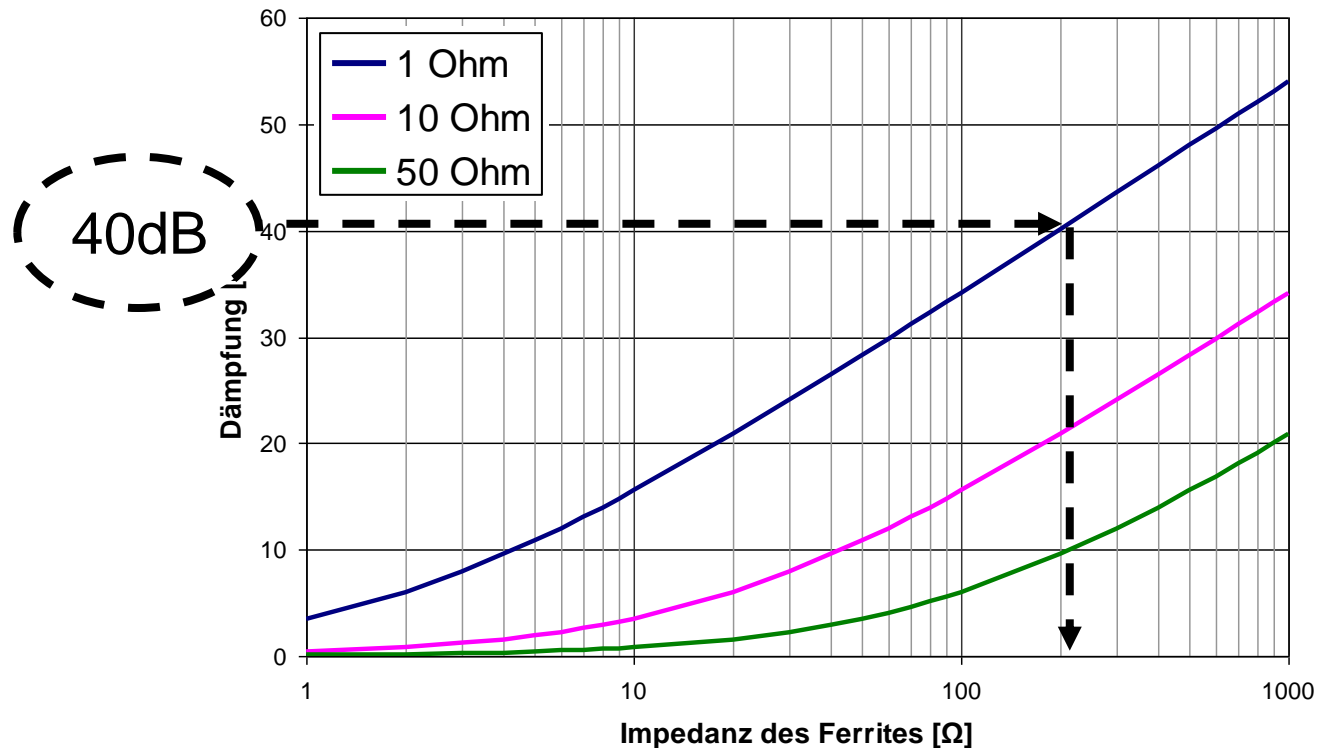
- (1) Required insertion loss of ferrite: **22dB @ 100 MHz**
- (2) System impedance for power supplies: **$Z < 10 \Omega$**
- (3) **$Z_{\text{ferrite}} = 220 \Omega$**
- (4) **742792022**



Use the WE catalog and use 749792022

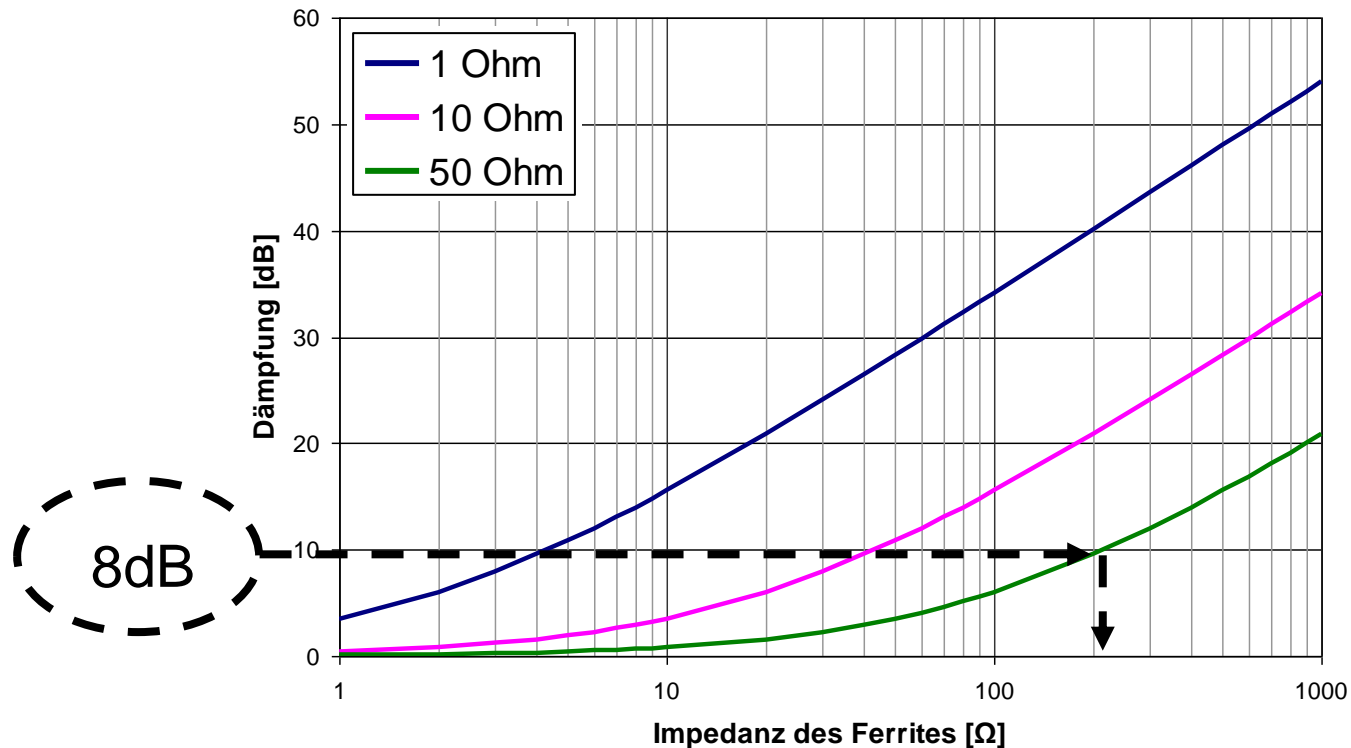


Possibility 1: too high attenuation



→ Could be because of wrong system impedance
→ reduce the impedance of ferrite

Possibility 2: too low attenuation



- Could be because of wrong system impedance
- increase the impedance of ferrite ($Z_F \sim 1000\Omega$)

Insertion loss – recommended filter topology

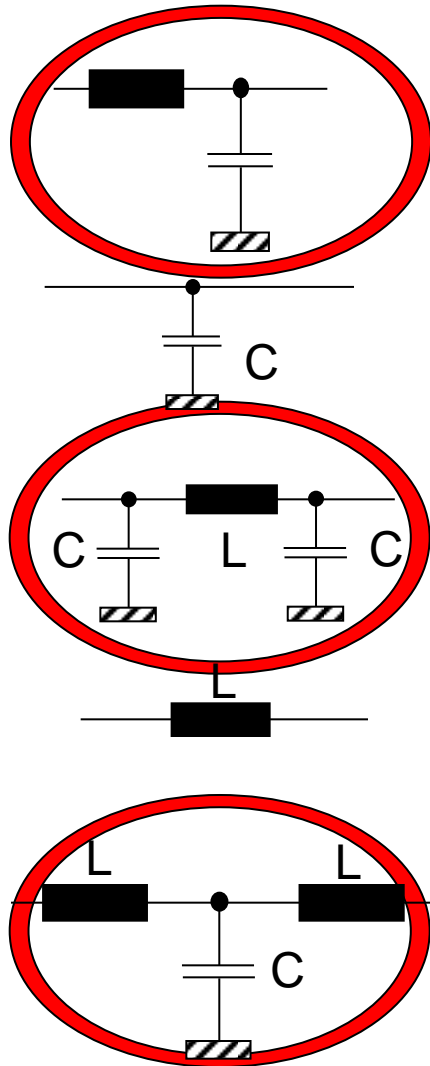
Source Impedance

low

high

high or
unknown

low

low or
unknown

Load Impedance

high

high

high or
unknown

low

low or
unknown

→ small C = higher SRF

Choose ferrite bead or inductors L which
 = build no resonance with C
 = broadband filter

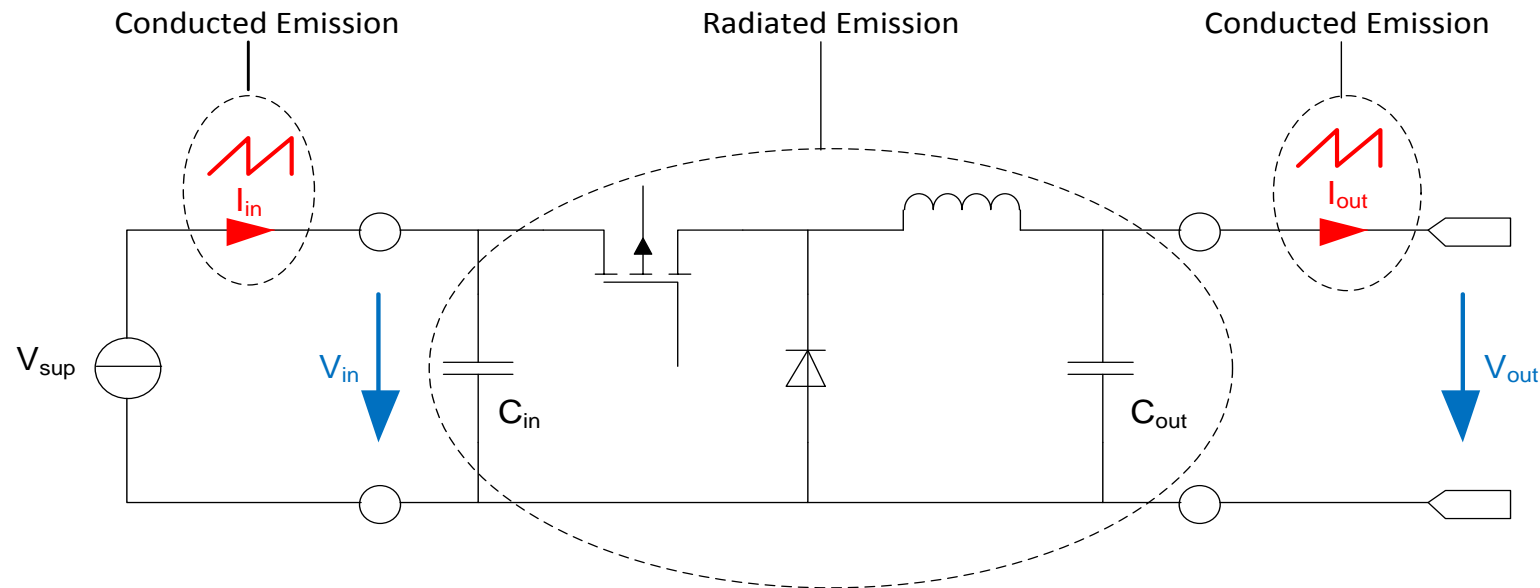
Pay attention to:

SRF of used components



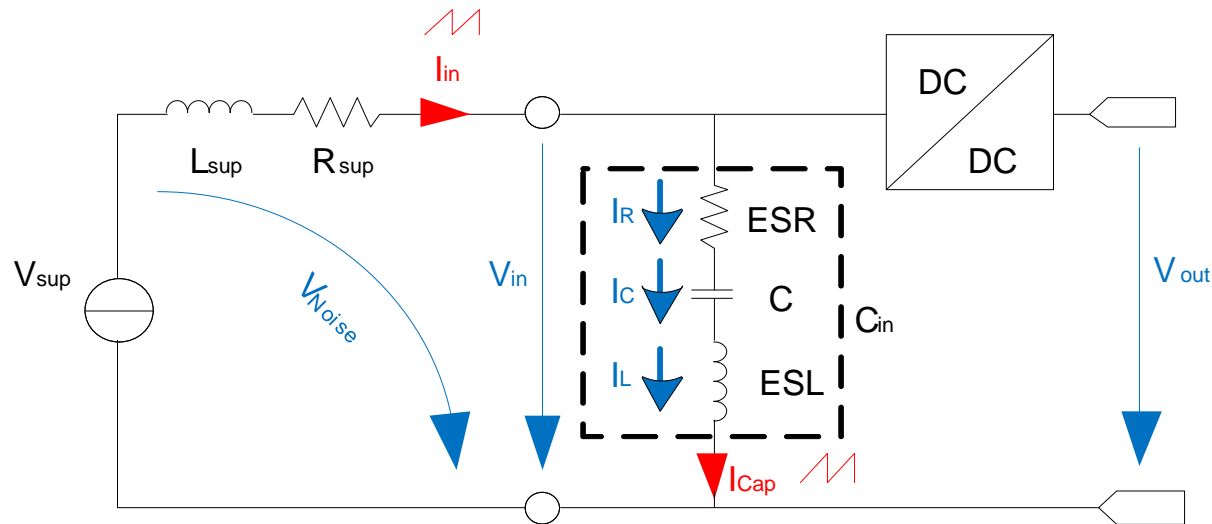
EMI NOISE SOURCES

Representative noise sources



- Input current caused by voltage ripple → „Conducted Emission“
- Power traces and choke radiate EMI → „Radiated Emission“
- Output current caused by voltage ripple → „Conducted Emission“
- Radiated emission will increase by using long input / output lines(cables)

Conducted noise at converter input

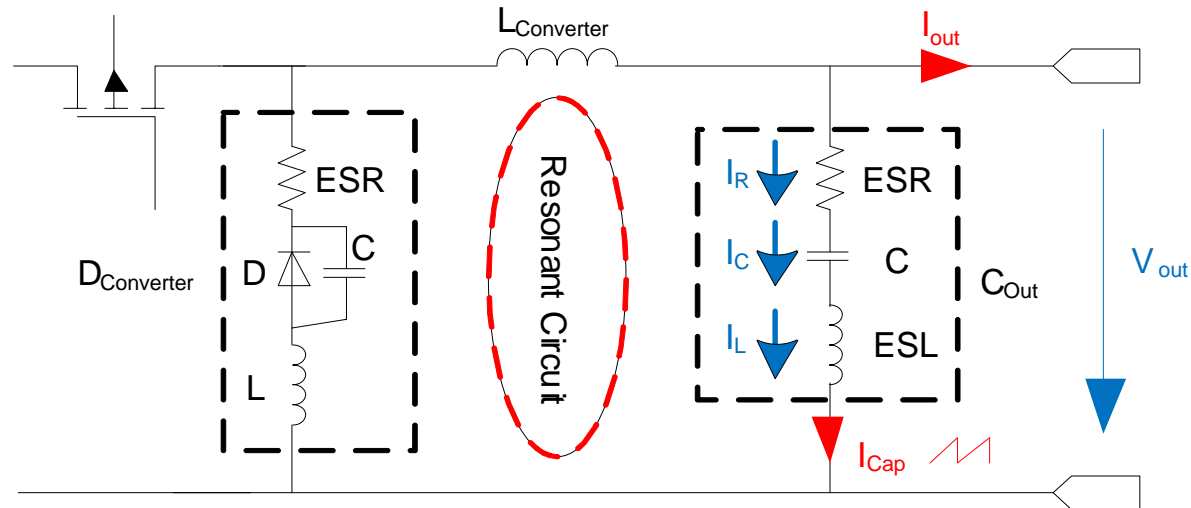


- **Conducted Emission is generated by voltage drop across R_{sup} and ESR_L**

$$V_{Noise} = R_{sup} * I_{in} + ESR * I_{Cap}$$
- **$V_{Noise} = R_{sup} * I_{in} + ESR * I_{Cap}$**
- **Resonance circuit is formed by L_{sup} , C_{in} and ESL_{Cin}**

$$f_0 = 1 / 2\pi \sqrt{(L_{sup} - ESL) * C_{in}}$$
- **Different harmonics due to fundamental frequency from $f_{DC/DC}$ and $f_{Resonance\ Circuit}$**

Conducted noise at converter output



- Conducted emission is generated by voltage drop at ESR_C

$$U_{Noise} = ESR_{Cout} * I_{Cout}$$

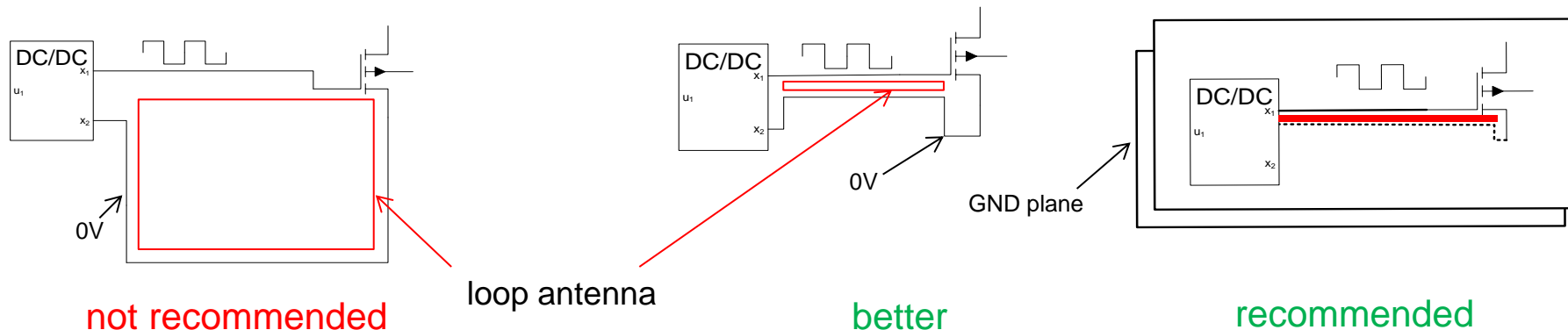
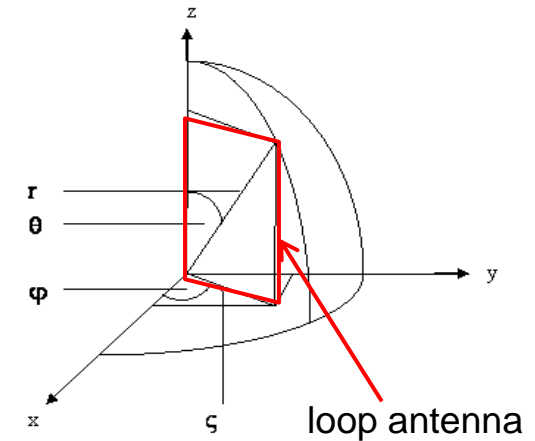
- Resonance circuit is formed by $C_{Dconverter}$, C_{Out} , $L_{Converter}$, and ESL_{Cout}

$$f_0 = \frac{1}{2\pi\sqrt{(ESL_{Cout}) * C_{Out}}}$$

- Different harmonics due to fundamental frequency from $f_{DC/DC}$ and $f_{Resonance\ Circuit}$

Radiation of PCB traces

- Power and signal loops have antenna characteristics
- Radiation can occur over the entire power and signal loops
- Field strength depends on spanned loop, peak value of alternating current, frequency, distance between noise source and noise receiver
- Design recommendations:
 - Keep power and signal traces as short as possible
 - Keep power and signal loops as small as possible
 - Route the trace over GND plane

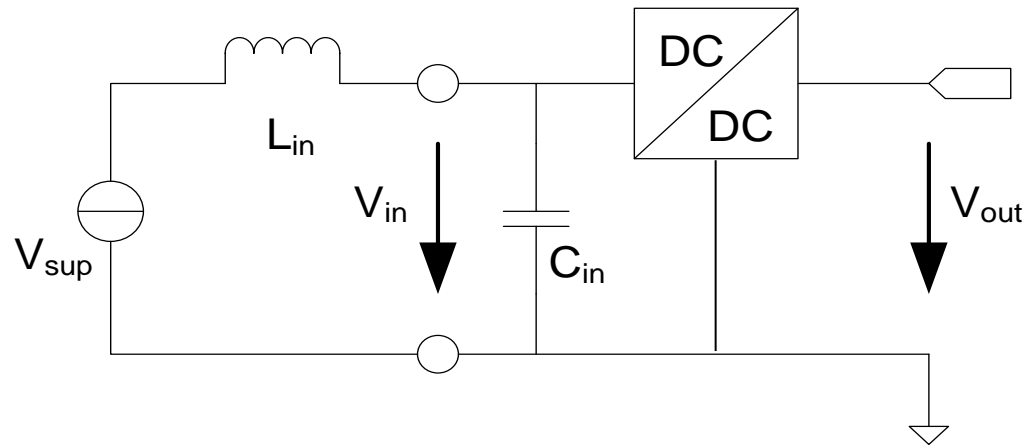




FILTER DESIGN

„L“ Input filter

(minimal recommend filter)



■ Simple L-Filter

- Input filter reduce current ripple on input line
- Input filter reduce differential mode noise on input line
- Input filter reduce radiated emission via input traces

Attention!!! This filter is not efficient to reduce common mode noise on input lines

Calculating input inductance

$$L_{filter} = \frac{ESR(DC)(1 - DC)}{f \left(\frac{I_{sup}}{I_{con}} - \frac{ESR}{R_f} \right)}$$

ESR = Effective series resistance of input capacitor

DC = Converter duty cycle

I_{con} = Peak-to-peak input ripple current

I_{sup} = Required peak-to-peak ripple current for buck converters $I_{con} \approx I_{out}$

R_f = “Damping” resistor (for lower Q)

➤ For better filter performance choose next higher standard inductance value

Example:

$$V_{out} = 5.0V$$

$$I_{out} = 4.0A$$

$$f = 2.5MHz$$

$$ESR = 0.08\Omega$$

$$DC = 0.5 (\%50)$$

$$I_{con} \approx I_{out}$$

$$I_{sup} = 0.1A$$

$$R_f = 0 (\rightarrow \infty)$$

$$L_{filter} = \frac{0.08\Omega \times 0.5(1 - 0.5)}{2.5MHz \left(\frac{0.1A}{4A} - 0 \right)} = 320nH$$

➤ Choose 1μH (closest standard value)
 e.g.: WE-LQ “744 045 001
 or: WE-PD2 “744 773 0”



Calculating rated current

- $$I_L = \frac{(V_{out})(I_{out})}{(V_{in})(E)}$$

V_{out} = Output voltage

I_{out} = Output current

V_{in} = Input voltage

E = Efficiency

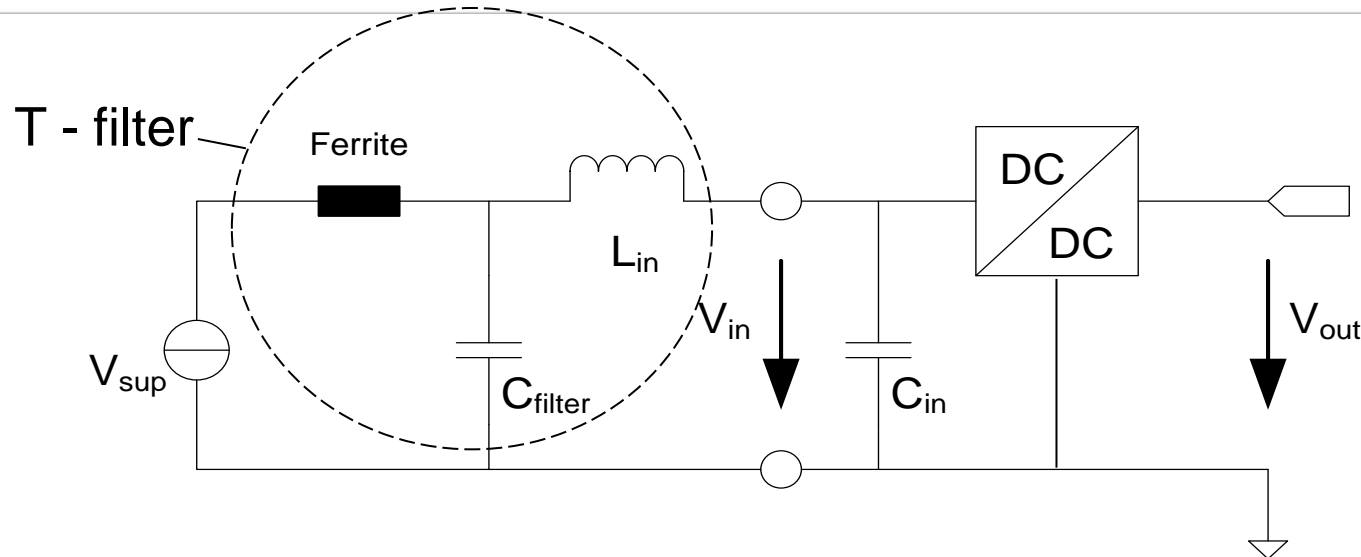
- To avoid overload considerations choose a choke with higher rating current
- To avoid losses in efficiency choose a choke with low DCR

For example:

- $$I_L = \frac{(5V)(4A)}{(20V)(0.8)} = 1.25A$$

Wideband input filter

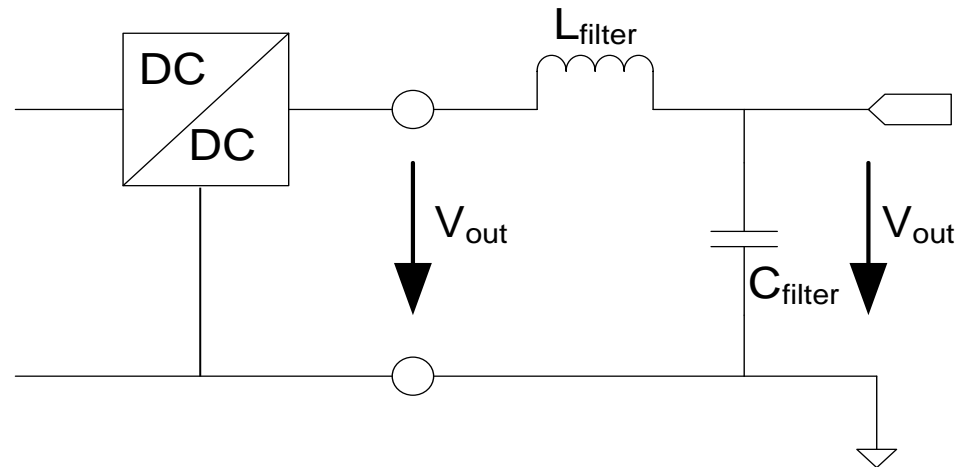
(recommended filter solution)



- **T-filter recommend for wideband filtering**
 - L_{in} for low frequency filtering (DC/DC converter switching frequency)
 - Ferrite for high frequency filtering
 - C_{filter} shorting ACnoise to GND ($220pF < C_{filter} < 1nF$, low ESR)

Attention!!! This filter is not efficient to reduce common mode noise on input lines

„L / C“ output filter (minimal recommended filter)



■ Simple L/C Filter

- Output filter reduce voltage ripple on output traces (Conducted Emission)
- Output filter reduce radiated emission via output traces (Radiated Emission)
- No optimal solution for radio power devices

Attention!!! This filter is not efficient to reduce common mode noise on output lines

Calculating „L / C“ output inductor

$$f_0 = \frac{1}{2\pi\sqrt{LC}} \quad \Leftrightarrow \quad L_{filter} = \frac{1}{(2\pi\frac{1}{10}f_0)^2 C_{filter}}$$

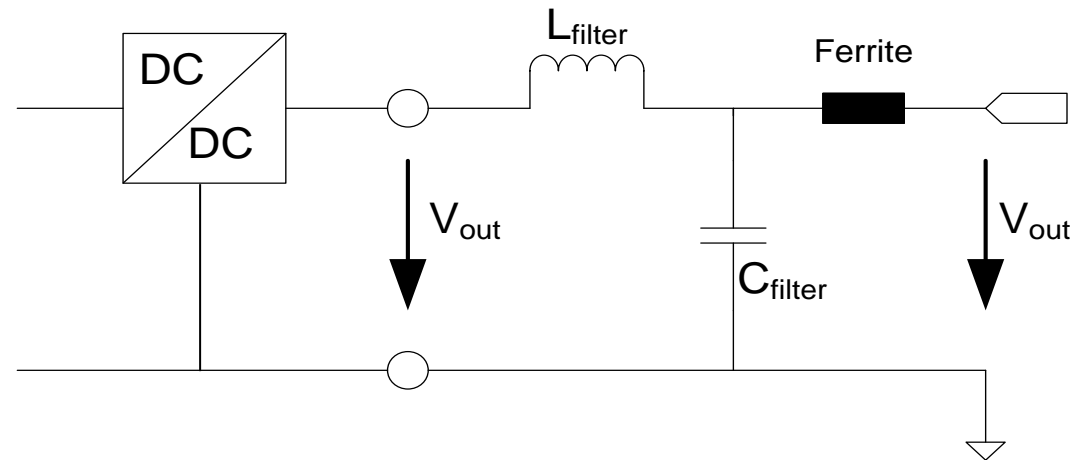
- **Example:**
 - (1) Choose capacitor e.g. 1µF, Electrolytic, low ESR
 - (2) Determine switching frequency of DC / DC converter
 - (3) Calculate inductor
 - (4) Choose next larger inductance value

$$L_{filter} = \frac{1}{(2\pi\frac{1}{10}f_0)^2 C_{filter}} \quad \Leftrightarrow \quad L_{filter} = \frac{1}{(2x\pi\frac{1}{10}1.6MHz)^2 X 1\mu F} = 989.5nH$$

$$L = 1.0\mu H$$



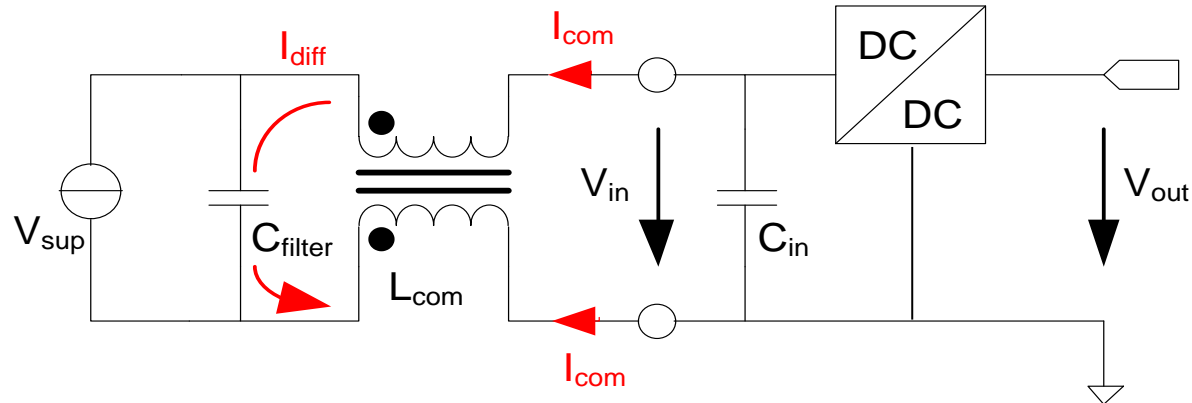
„T“ - output filter (recommended filter solution)



- T-filter recommend for wide bandwidth filtering
 - L_{filter} for low frequency filtering (DC/DC converter switching frequency)
 - Ferrite for high frequency filtering
 - This kind of output filter is for powering radio devices high recommended

Attention!!! This filter is not efficient to reduce common mode noise on output line

Decoupling common mode noise



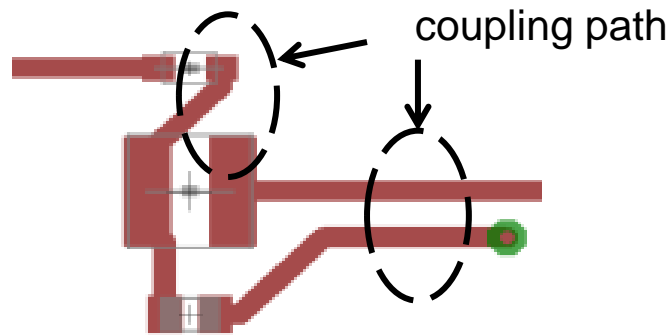
- For common mode rejection use common mode chokes
- For supplying over long distance common mode chokes are recommended
- Additional capacitor reduce differential mode noise
 - Small value for ceramic capacitor is recommended
 - Capacitor and common mode choke act as a LC - filter for differential mode noise
- Can be used for input and output lines



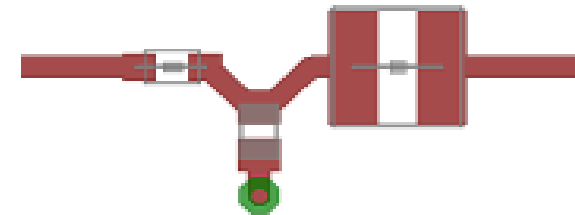
PCB - LAYOUT RECOMMENDATIONS

PCB-Layout recommendations

T-filter



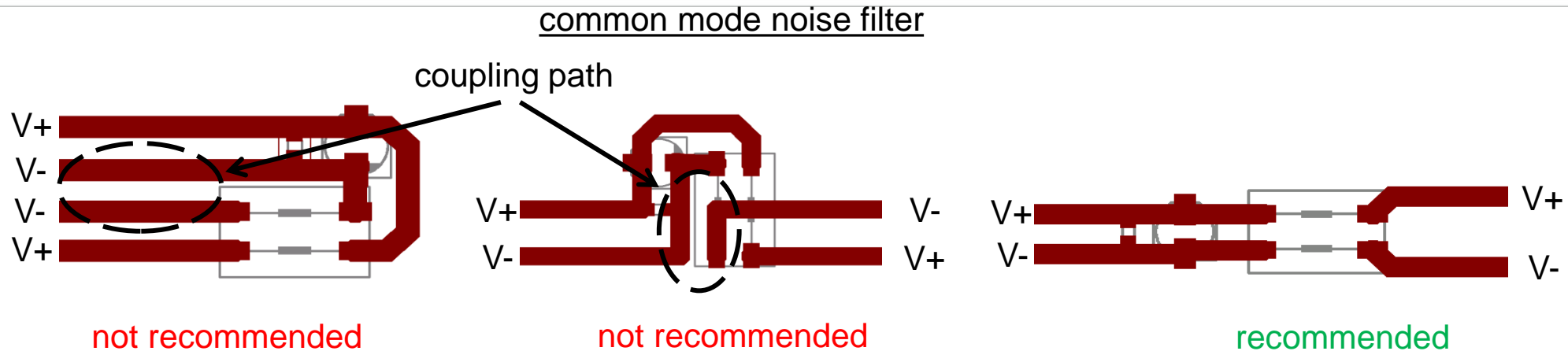
not recommended



recommended

- Keep PCB traces as short as possible
- Avoid indirect trace routing
- Avoid any kind couplings → “capacitive”, “inductive”
- AC-current should flow across capacitor
- Short way for AC-current direct to GND (place double vias to GND)

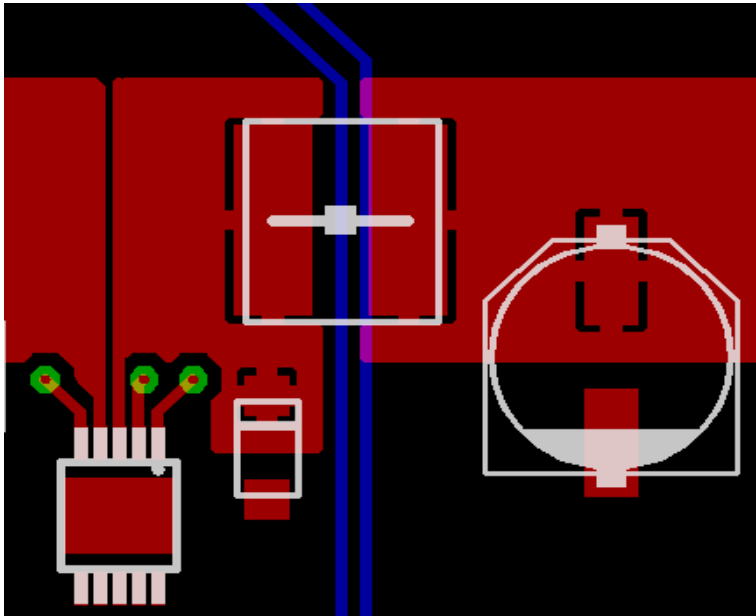
PCB-Layout recommendations



- Avoid indirect routing of power traces
- Avoid any kind of couplings → “capacitive”, “inductive” ... etc ...
- AC-current should flow across common mode choke
- Route power traces on component layer
- Do not use vias

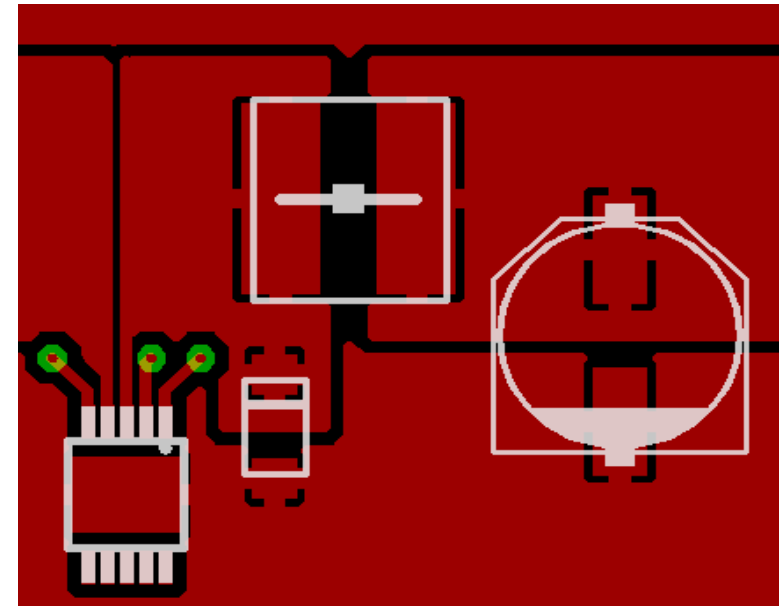
PCB-Layout recommendations

DC/DC buck converter



not recommended

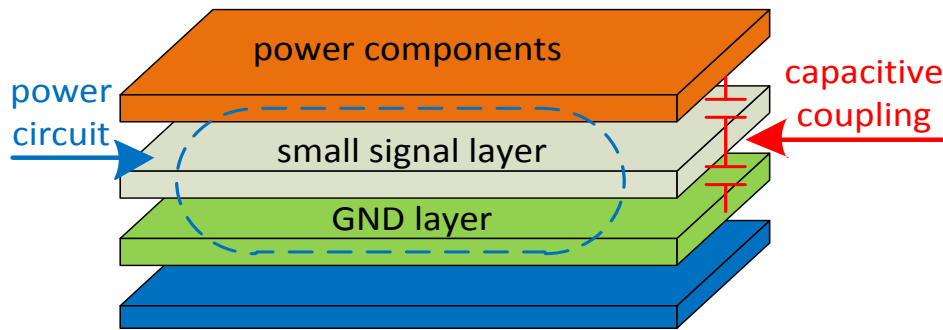
- **Avoid GND planes under inductor (between inductor pads)**
- **Don't route any kind of signals (analog, clock) under the inductor**
- **Fill out unused space on PCB with GND (flood)**



recommended

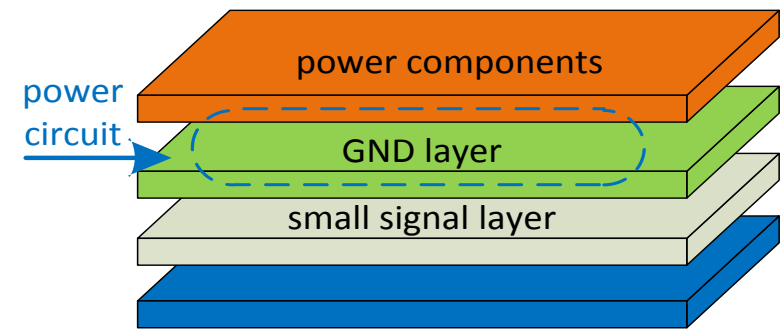
PCB-Layout recommendations

Layer 1: power components
Layer 2: small signal
Layer 3: pure GND layer
Layer 4: small signal / controller components



not recommended

Layer 1: power components
Layer 2: pure GND layer
Layer 3: small signal
Layer 4: small signal / controller components



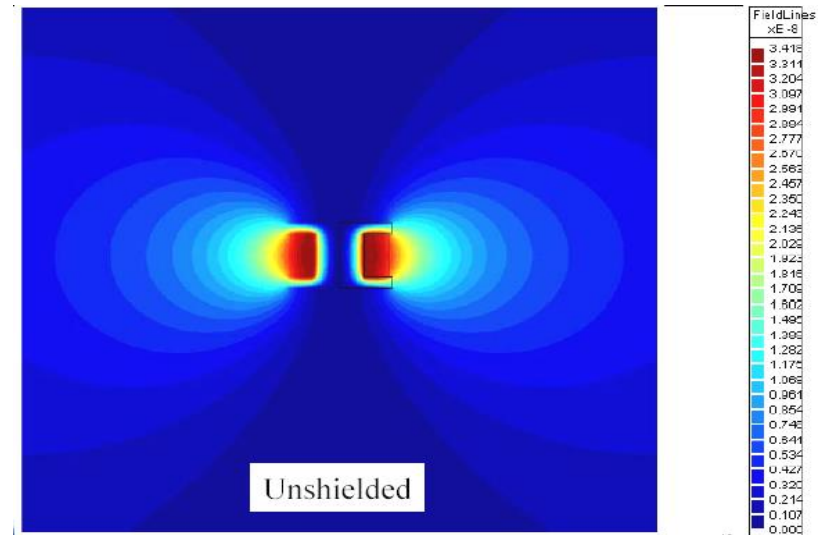
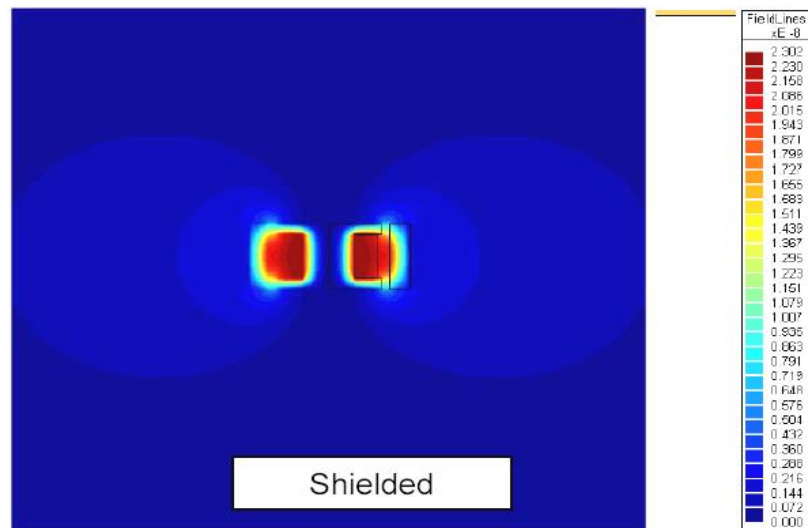
recommended

- PCB-design of at least 4 layers is recommended
- Place a solid ground plane below the power component layer
- Minimize loops for power components
- Keep power traces as short as possible
- Establish good GND connections using low impedance vias



SHIELD VS. UNSHIELD

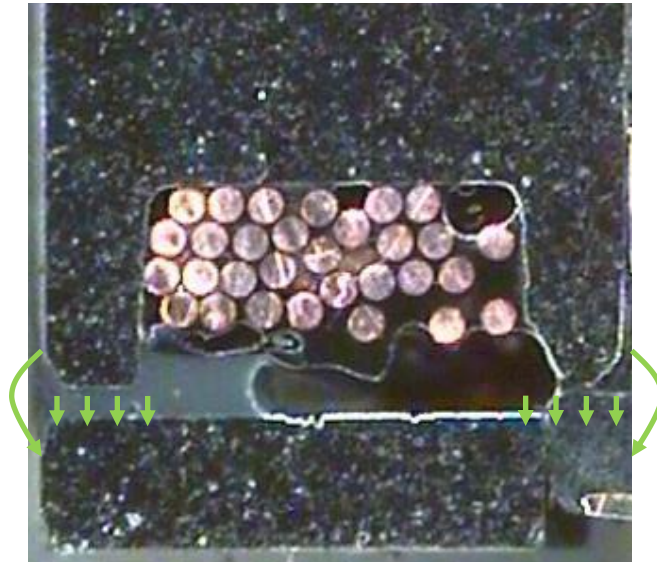
Magnetic field leakage



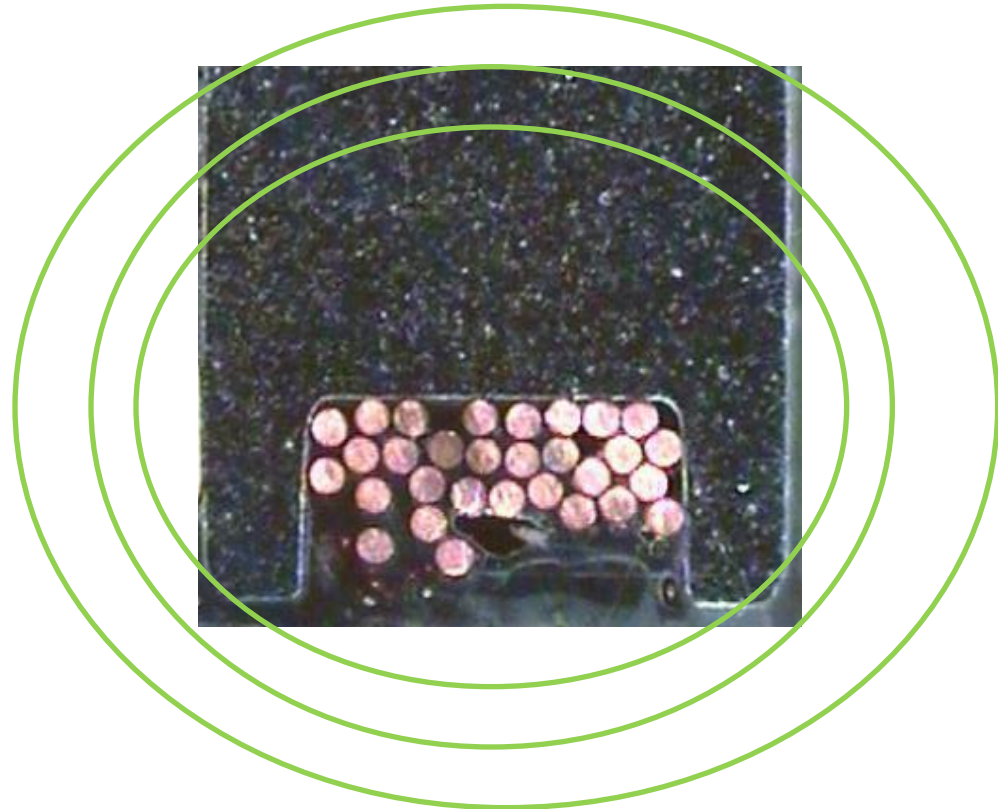
Magnetic Field – Shielded vs. Unshielded

- Magnetic field

shielded



unshielded



Radiation by inductor

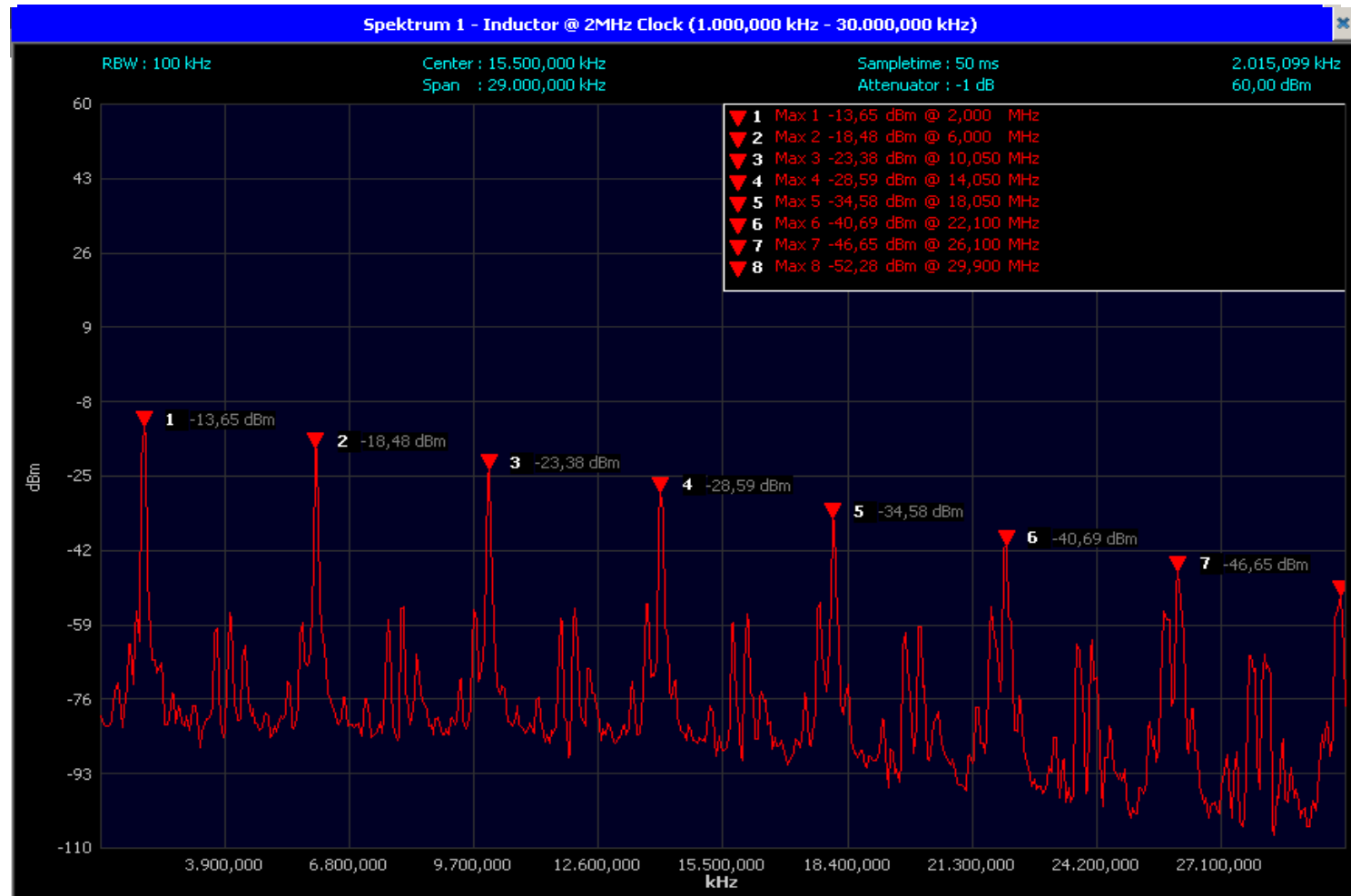
WE - PD2 unshielded
 10 μ H, 2MHz Clock, 1A



WE – PD shielded
 10 μ H, 2MHz Clock, 1A

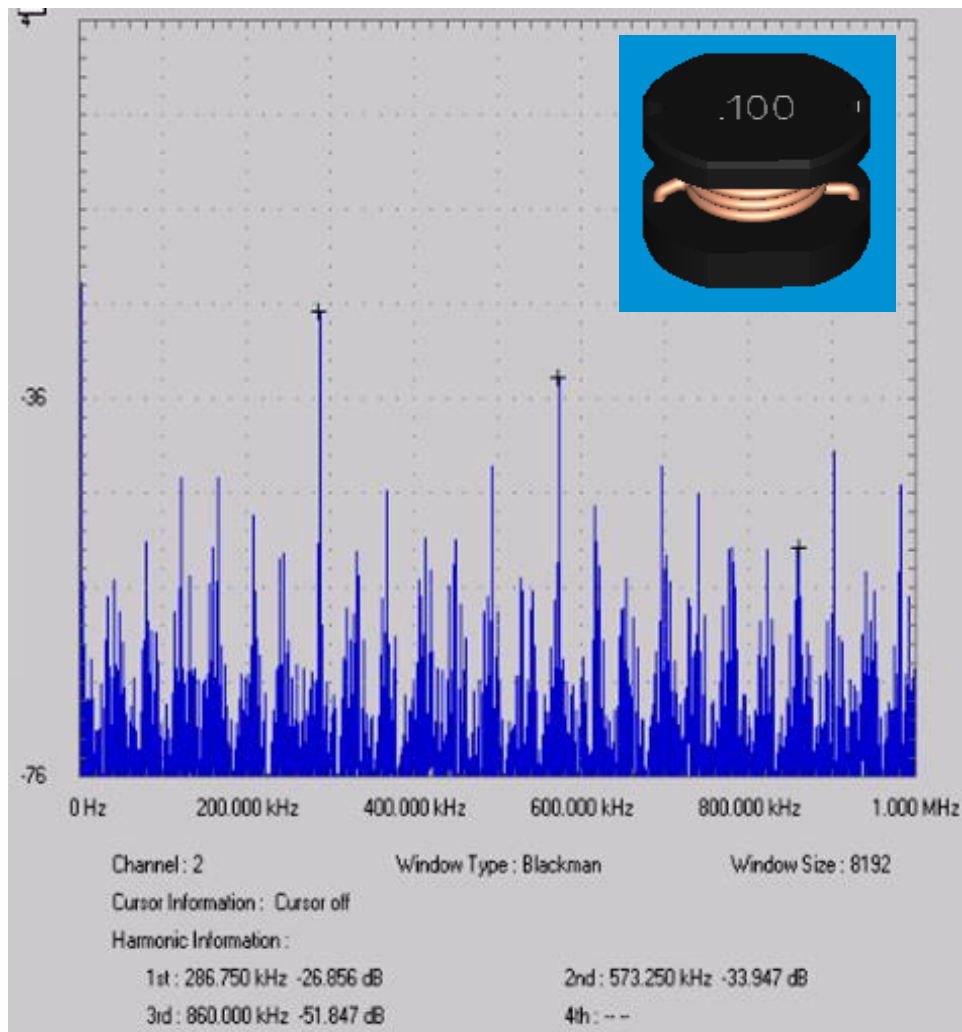


19dBm difference

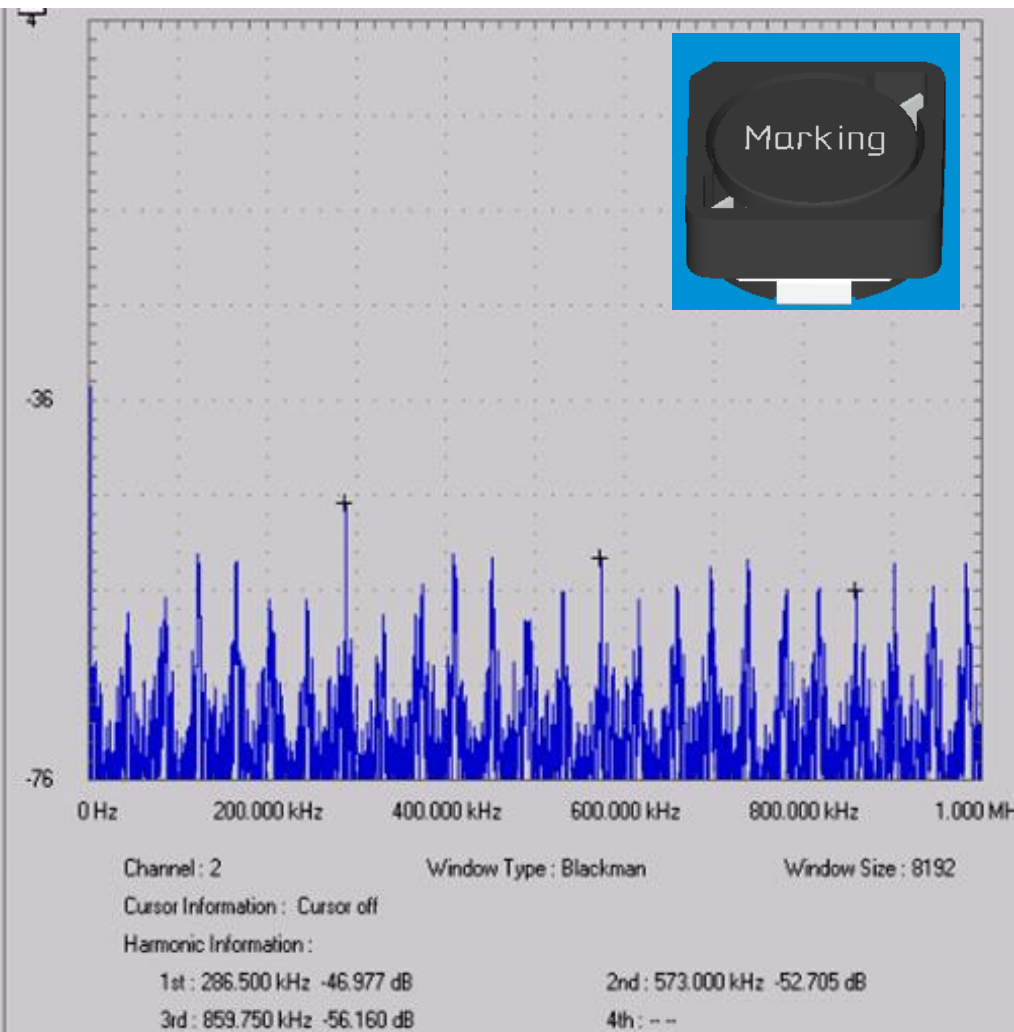


Magnetic leakage shielded vs. unshielded

unshielded

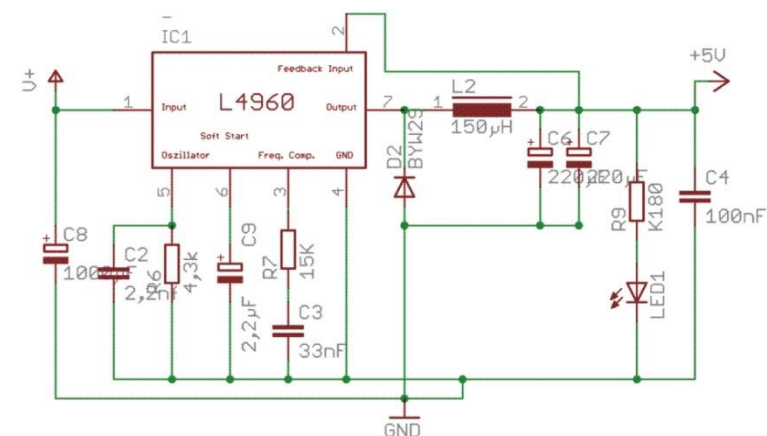
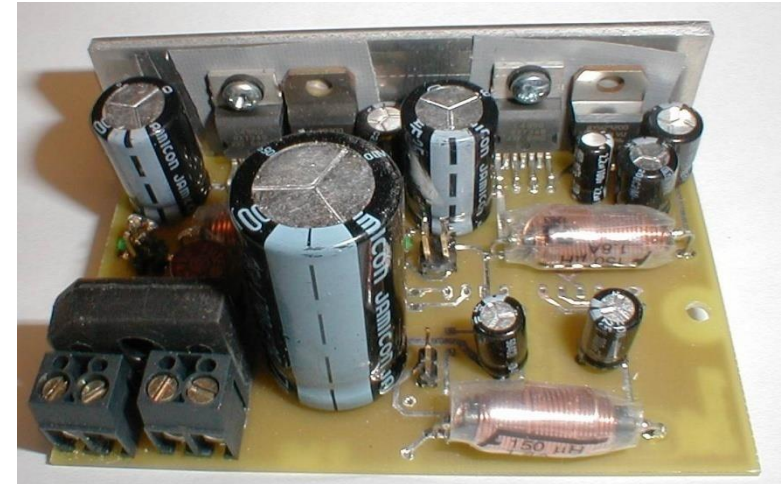
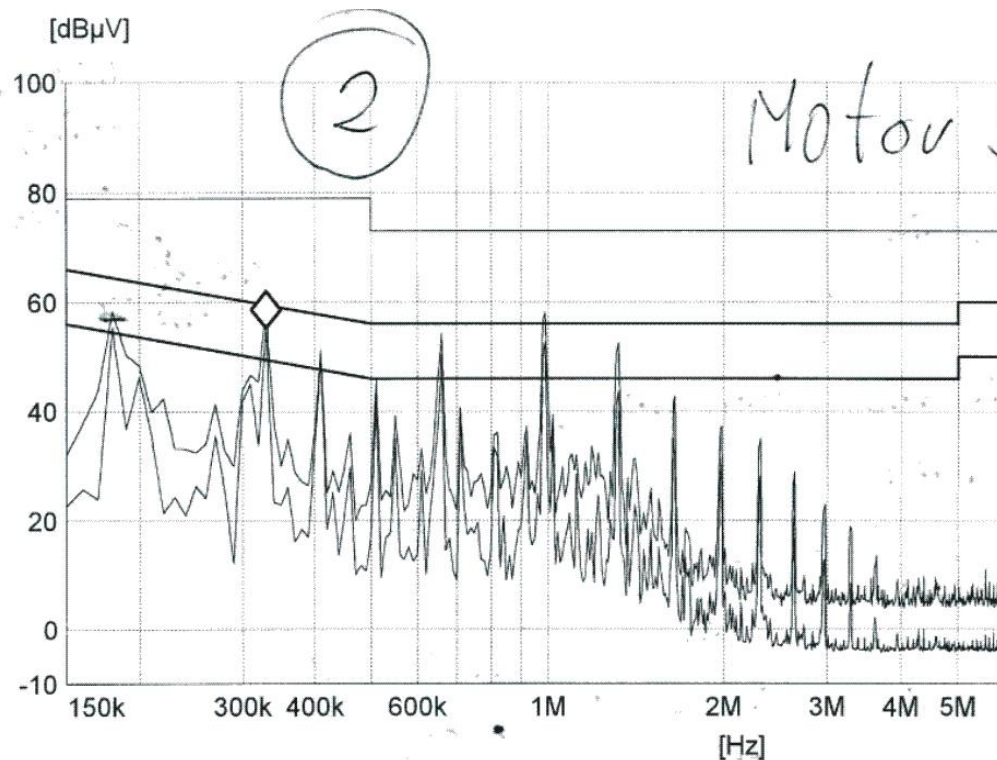


shielded



Magnetic Fields – Conducted Emission Measurement

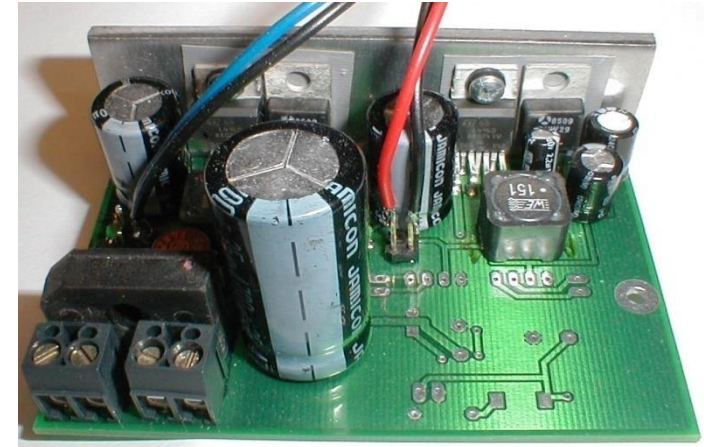
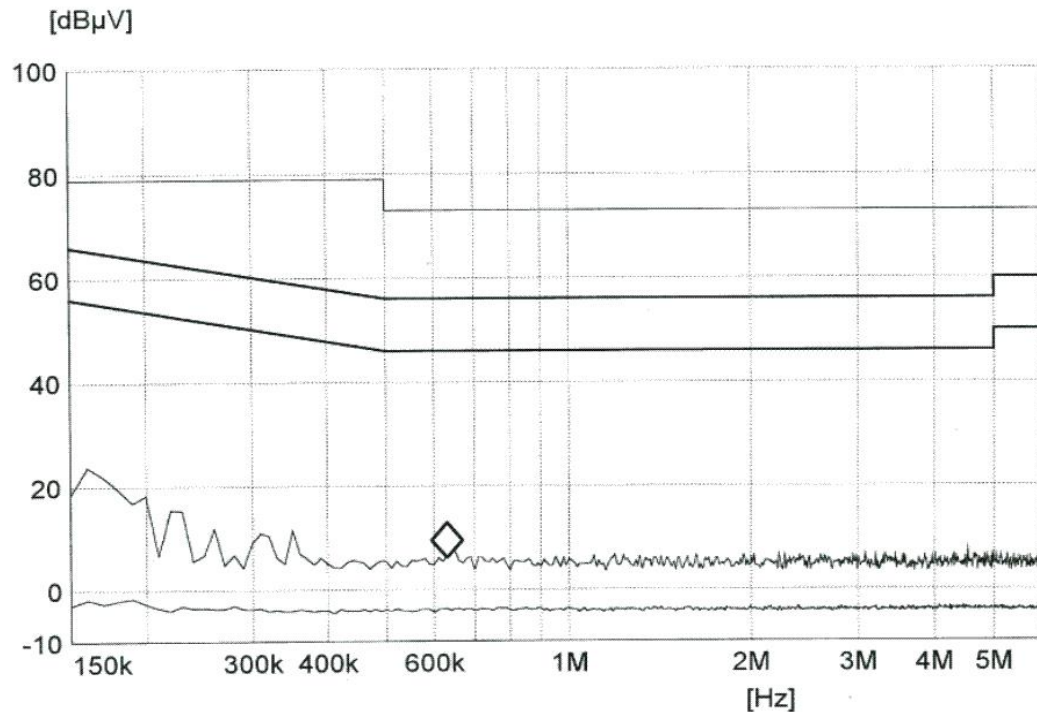
Power supply V 1.0



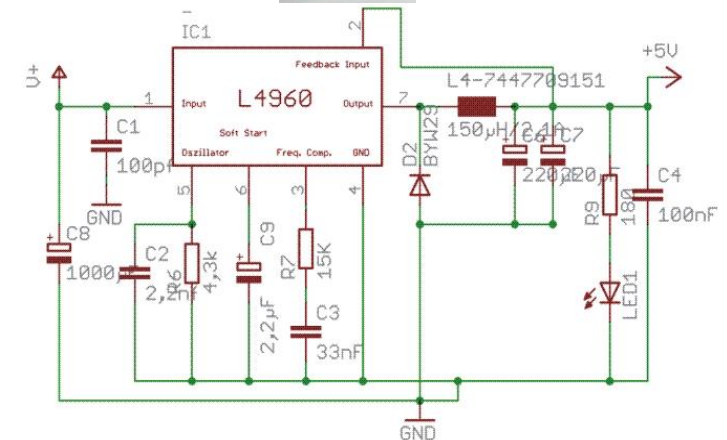
Buck Converter ST L4960/2.5A/fs 85-115KHz

Magnetic Fields – Conducted Emission Measurement

Power supply V 1.1



PCB



Schematic

Magnetic Fields – Be Aware!

- Select the right parts for your application.
- Do not always look on cost.

Very easy solution with a dramatic result!



Choke before



or

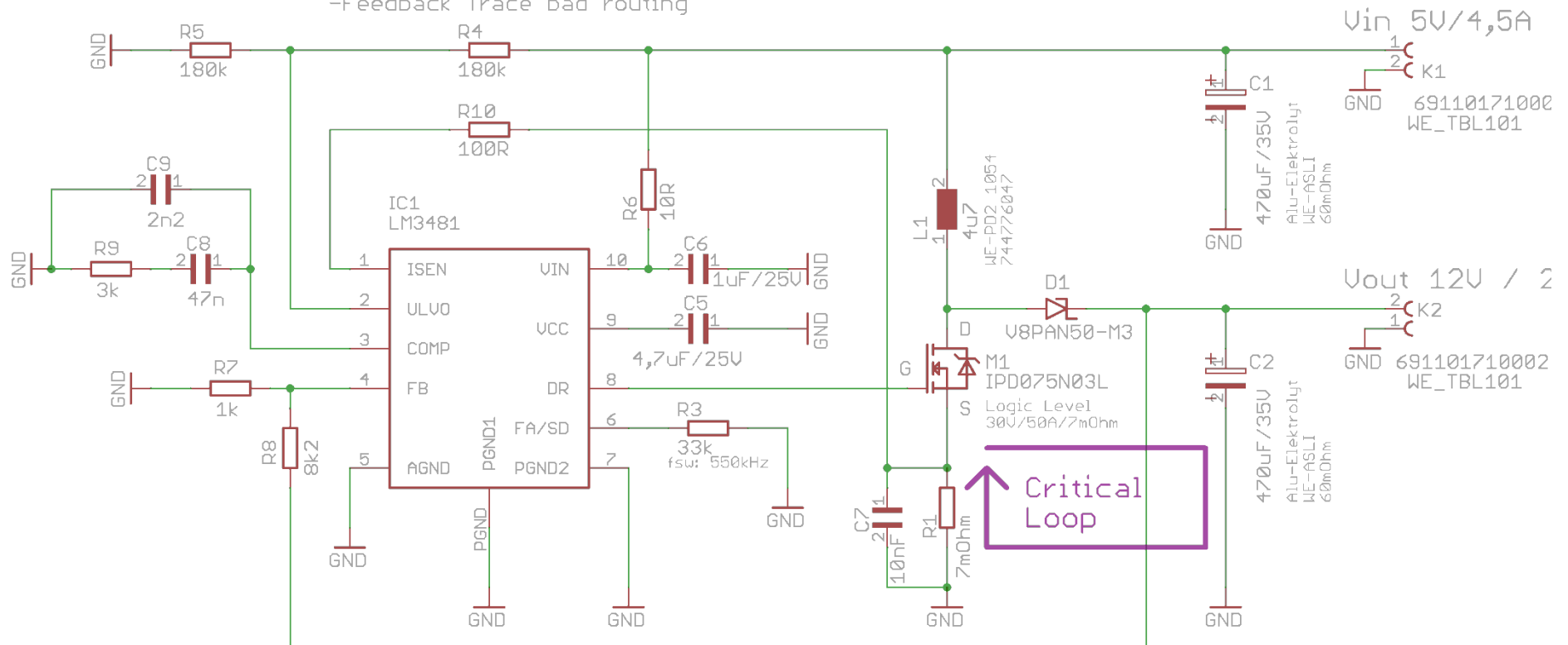


Choke after

Boost converter Bad Example

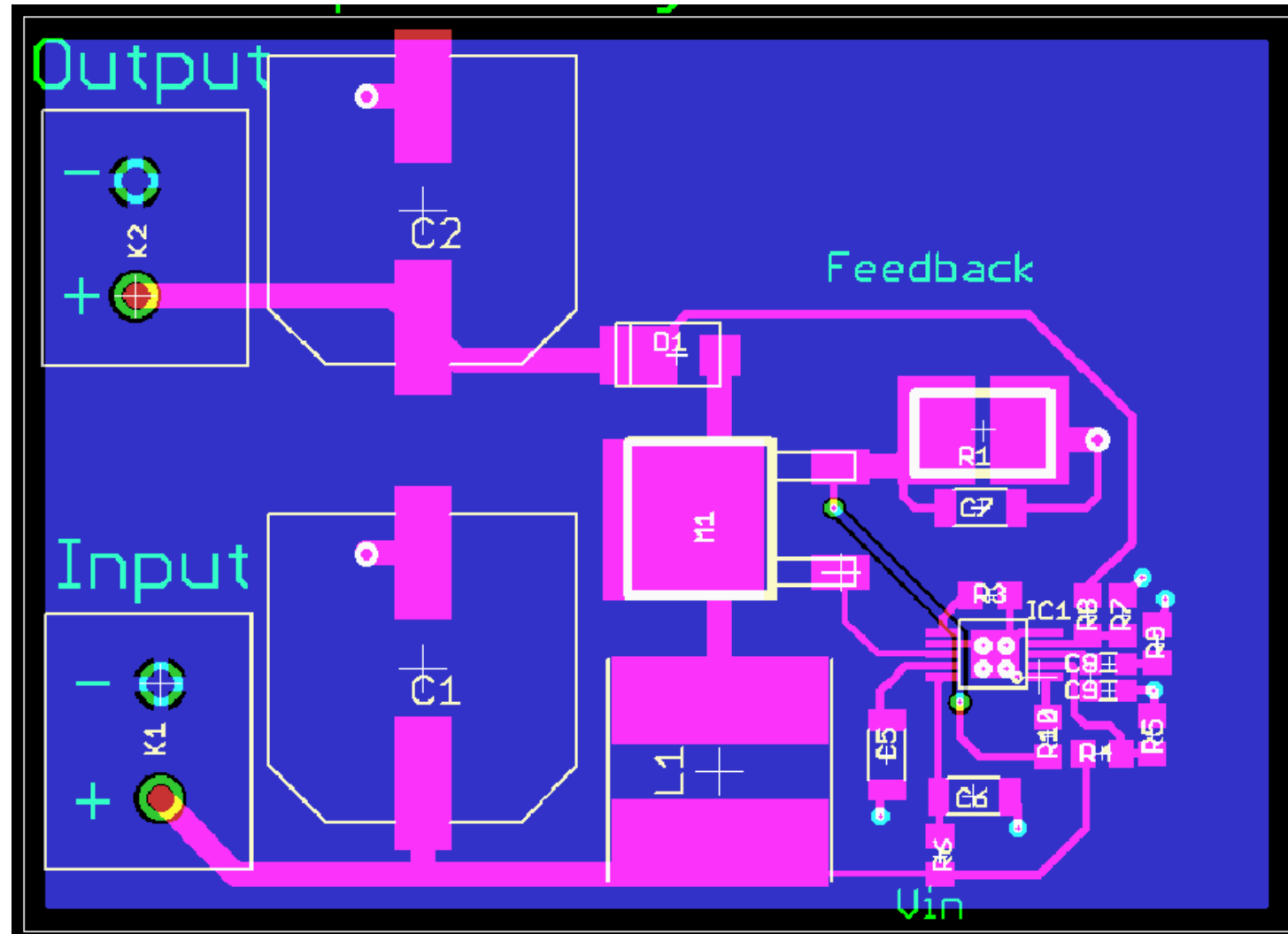
Bad Example:

- Standard Alu-Electrolyte Cap Input/Output >ESR
- Power Inductor is not shielded
- Bad Layout with large Current Loops
- Power and Analog GND not separated
- No MOSFET Gate Resistor
- No Input&Output Filter
- Feedback Trace bad routing

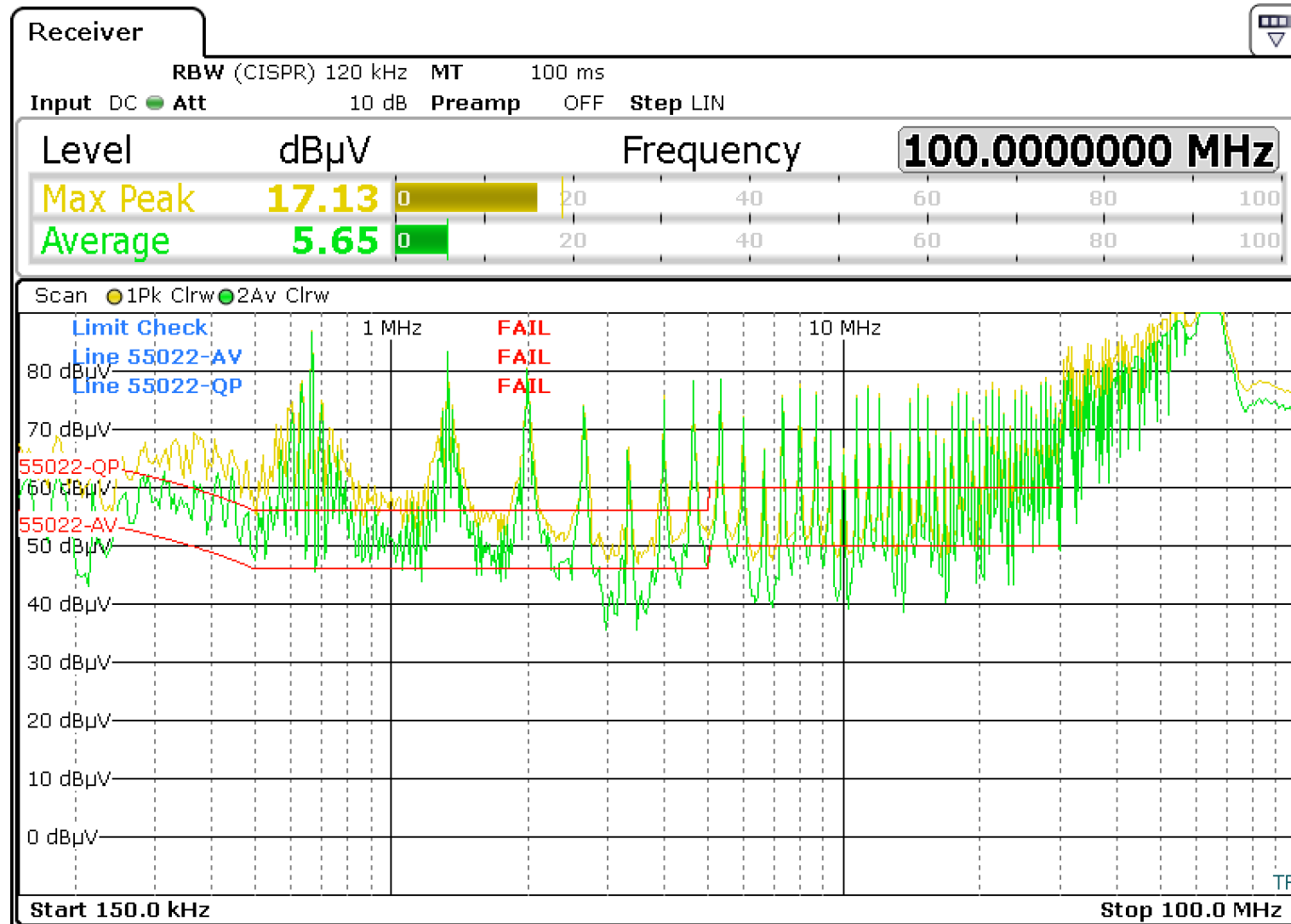


A Boost Converter is critical at the Output!!

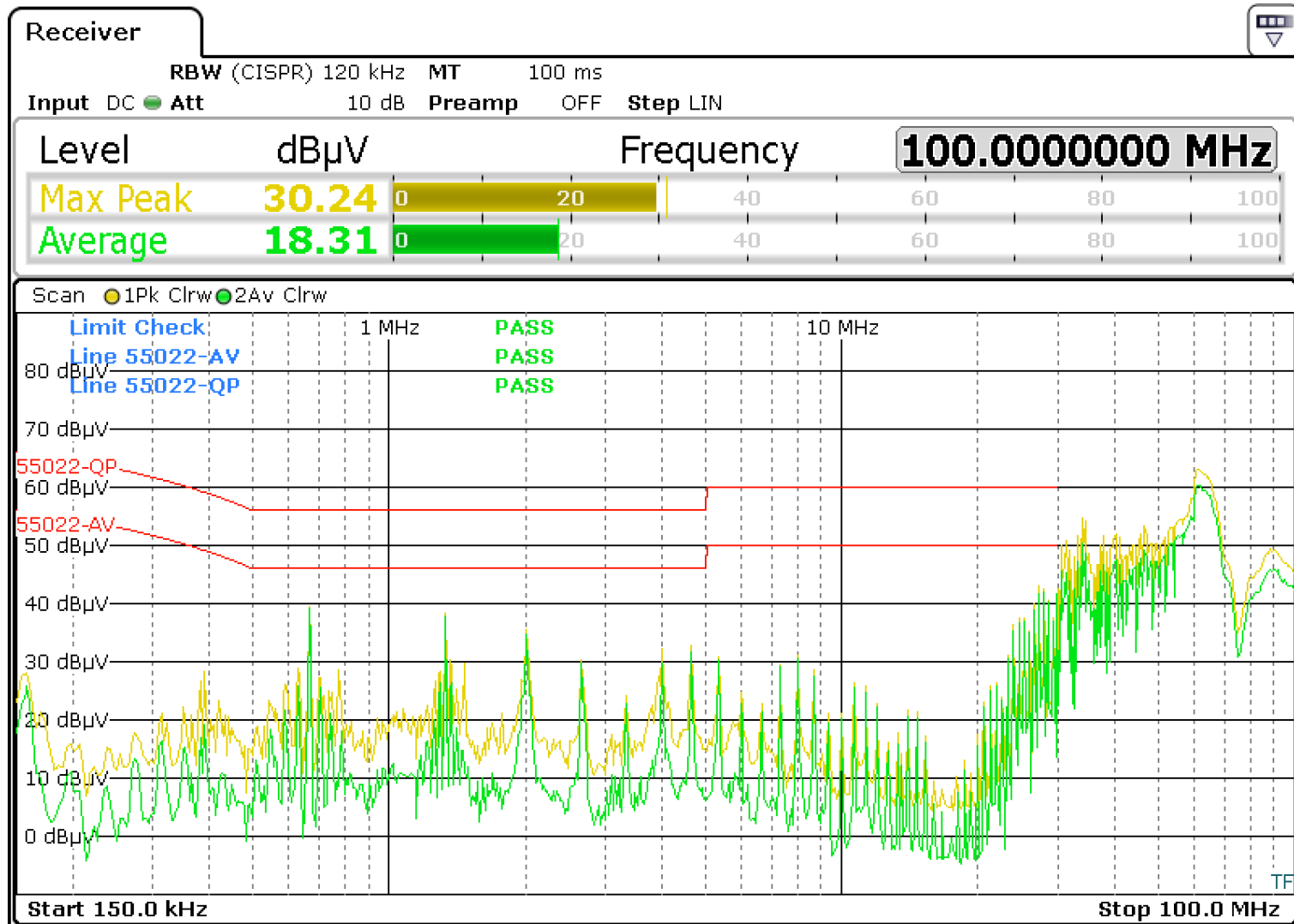
Boost converter Bad Example



Boost converter Bad Example no filter

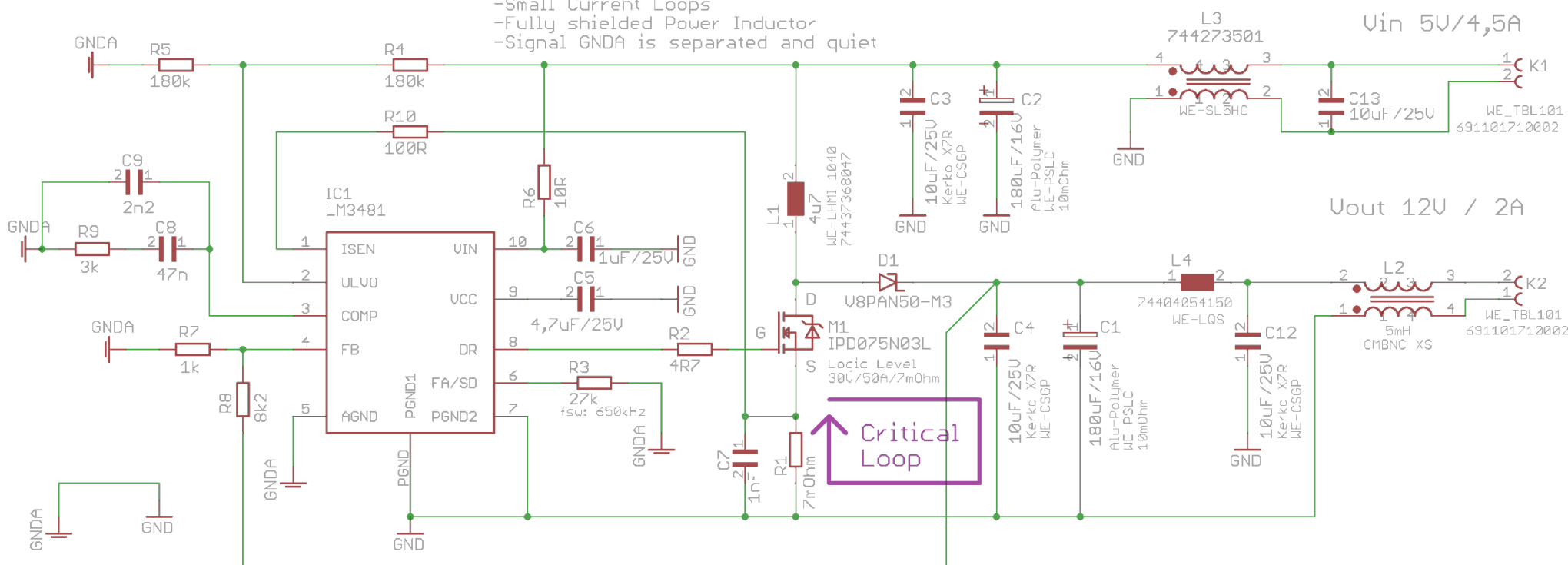


Boost converter Bad Example with filter



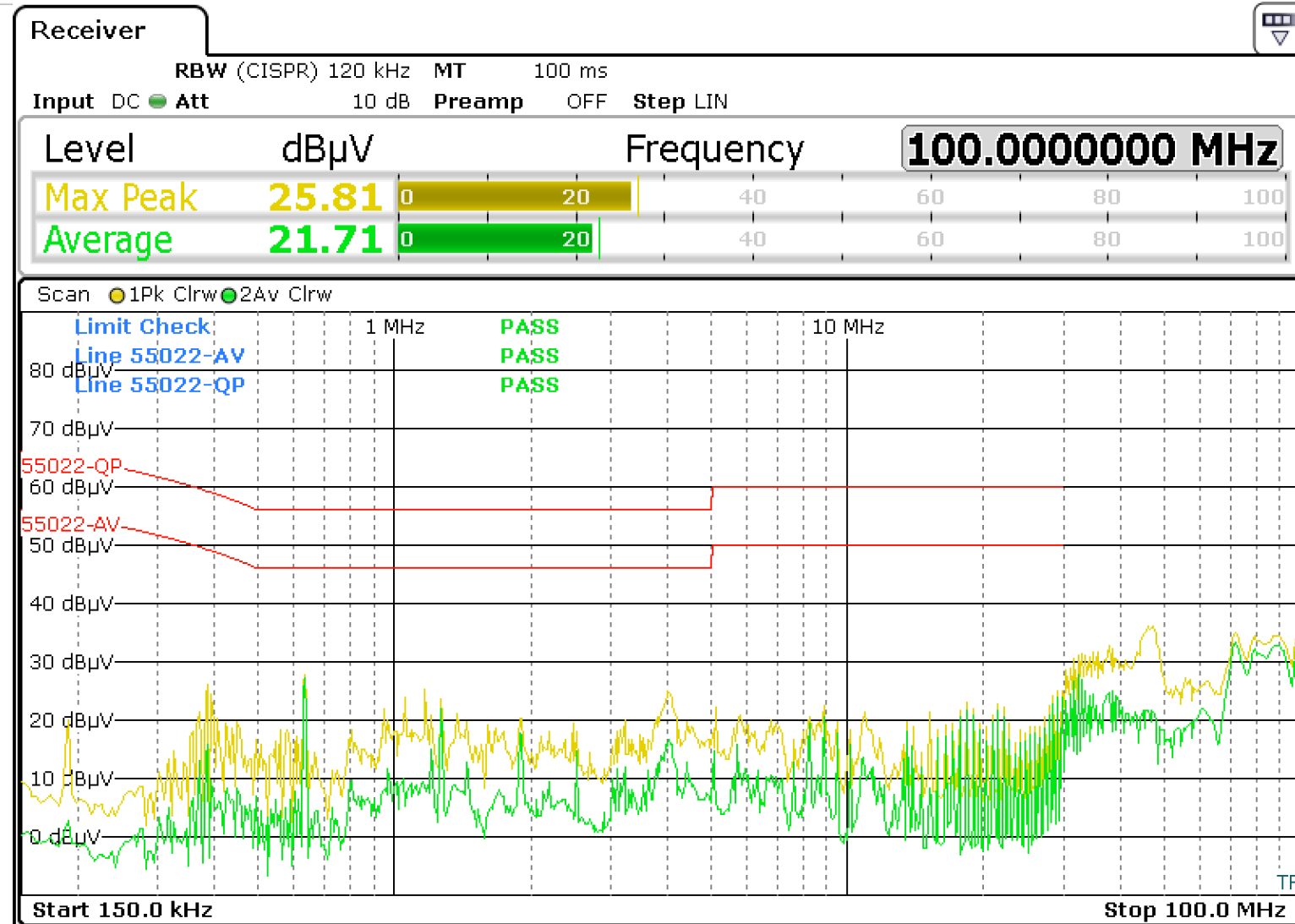
Boost converter Good Example

Good Example:
 -Low ESR Input Caps
 -Low ESR Output Caps
 -Input&Output Filter
 -Low Impedance Layout
 -Small Current Loops
 -Fully shielded Power Inductor
 -Signal GND is separated and quiet



A Boost Converter is critical at the Output!!

Boost converter Good Example

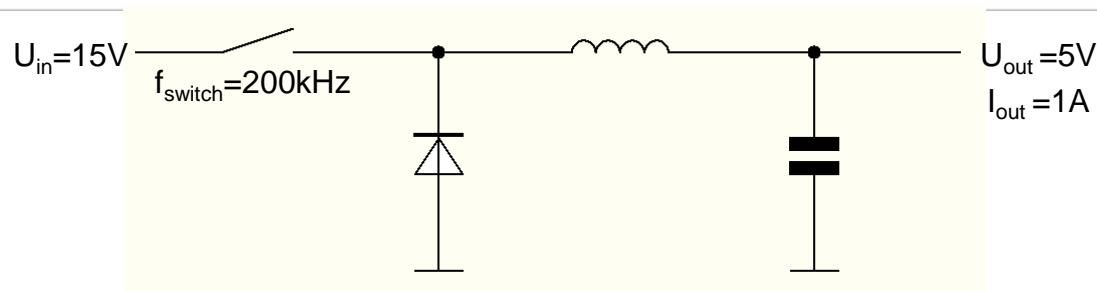




STORAGE INDUCTOR SELECTION

Inductor selection

Example: Step down converter



$$\textcircled{1} DC = \frac{U_{out}}{U_{in}} = \frac{5V}{15V} \cong 0,33$$

$$\textcircled{2} I_{rated} \cong I_{out} = 1A$$

$$\textcircled{3} I_{ripple} \cong 20\% \dots 40\% \cdot I_{out} = 0,2 \dots 0,4A \quad (\text{practical values})$$

$$\textcircled{4} L = \frac{DC \cdot (U_{in} - U_{out})}{f_{switch} \cdot r \cdot I_{out}} = \frac{0,33 \cdot (15V - 5V)s}{200E3 \cdot 0,2 \cdot 1A}$$

$$L = 83 \dots 33 \mu H$$

\uparrow $r = 0,2$ \uparrow $r = 0,5$

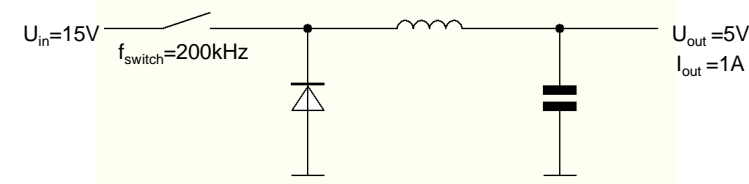
→ choose average value 56μH to begin optimization

Inductor selection

Example: Step down converter

$$I_{out} = 1A \xrightarrow{r=0,5} I_{Lmax} \cong 1,5A \rightarrow I_{sat} \geq 1,5A$$

$$L = 56\mu H \quad \text{at} \quad r = 0,3$$



WE-PD Type 1260 744 771 156 $I_{rated}=2,01A$ $I_{sat}=2,35A$



RECOMMENDATION:

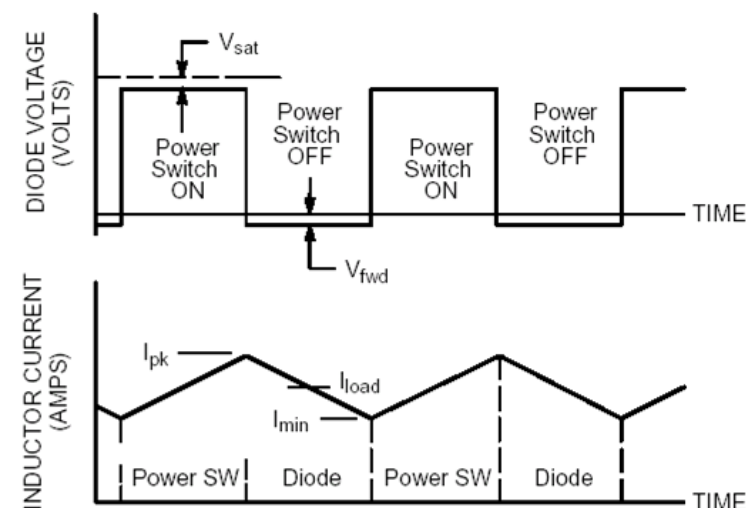
- test different inductor values
 - consider the tolerance of L- values
 - effect on the design (e.g. Ripple current; RDC)

choose additional:

744 771 133 + 744 771 168

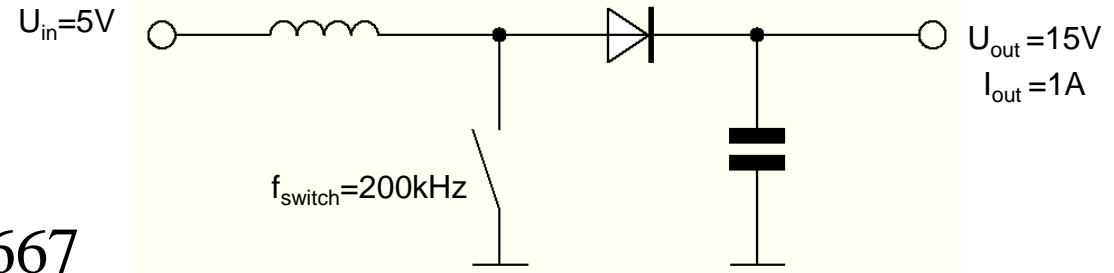
$$L = 33\mu H$$

$$L = 68\mu H$$



Inductor selection

Example: Step up converter



$$\textcircled{1} \quad DC = 1 - \frac{U_{in}}{U_{out}} = 1 - \frac{5V}{15V} \cong 0,667$$

$$\textcircled{2} \quad I_{rated} \cong \frac{I_{out}}{1 - DC} = 3A$$

$$\textcircled{3} \quad I_{ripple} \cong 20\% \dots 40\% \cdot I_{out} = 0,6 \dots 1,2A \quad (\text{practical values})$$

$$\textcircled{4} \quad L = \frac{(U_{out} - U_{in}) \cdot (1 - DC)^2}{f_{switch} \cdot r \cdot I_{out}} = \frac{(15V - 5V) \cdot (1 - 0,667)^2 s}{200E3 \cdot 0,2 \cdot 1A}$$

$$L = 10 \dots 56 \mu H$$

\uparrow $r = 0,5$ \uparrow $r = 0,1$

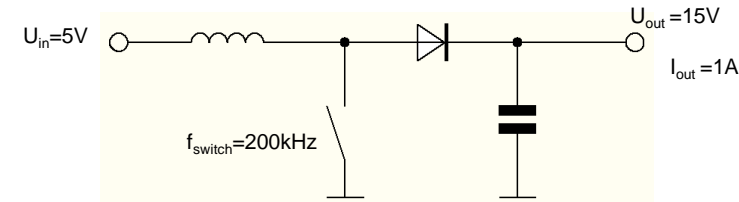
→ choose average value 33μH to begin optimization

Inductor selection

Example: Step up converter

$$I_{out} = 1A \xrightarrow{r=0,5} I_{Lmax} \cong 3,75A \rightarrow I_{sat} \geq 3,5A$$

$$L = 33\mu H \quad \text{at} \quad r = 0,2$$



WE-PD Type 1280 744 770 133 $I_{rated}=3,20A$ $I_{sat}=3,60A$



RECOMMENDATION:

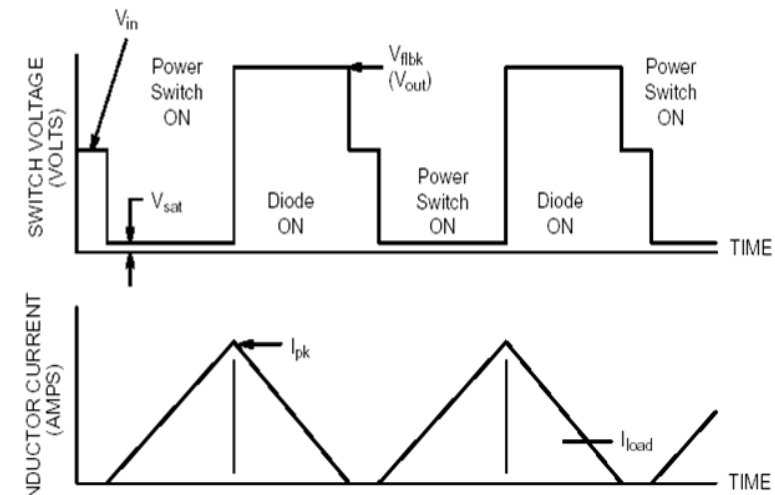
- test different inductor values
 - consider the tolerance of L- values
 - effect on the design (e.g. Ripple current; RDC)

choose additional:

744 770 9470 + 744 770 122

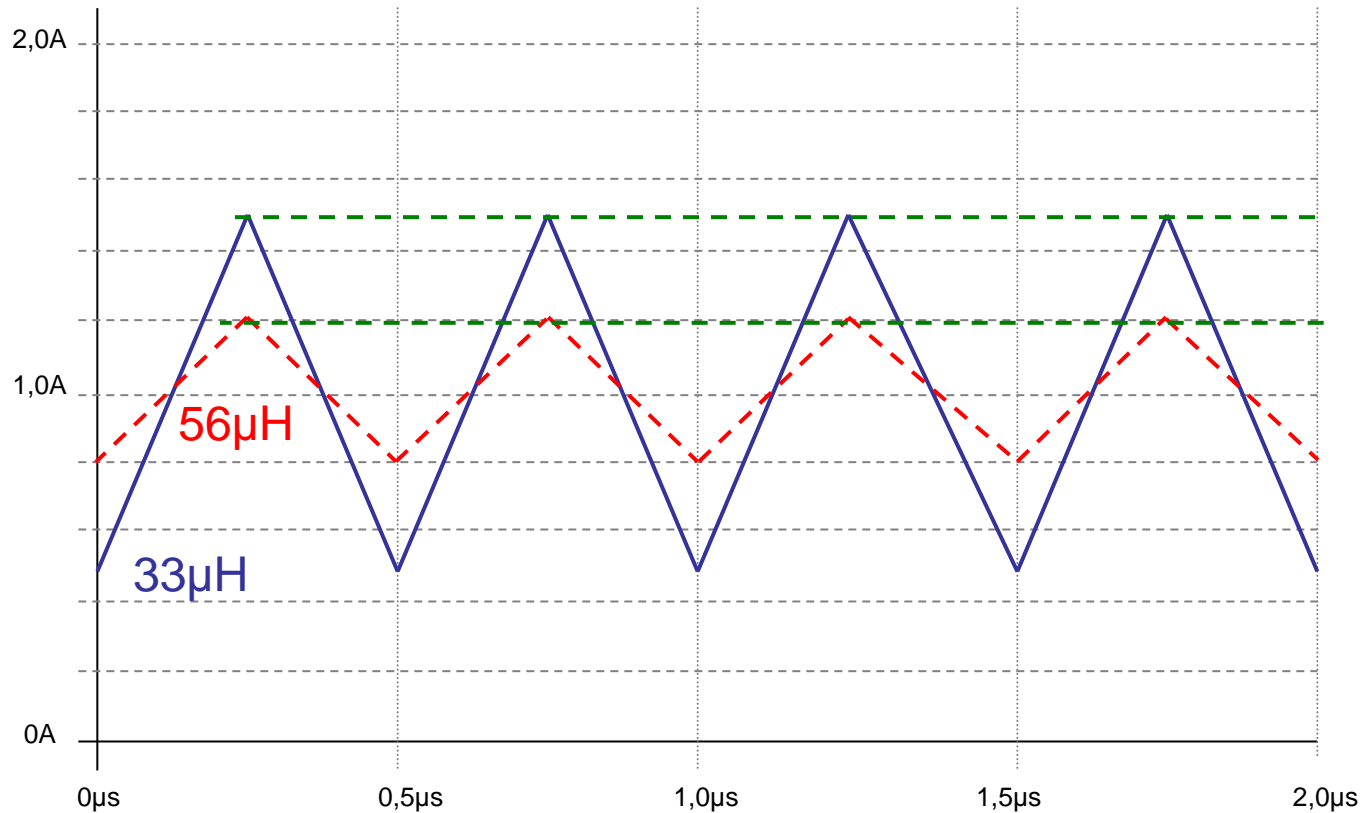
$$L = 47\mu H$$

$$L = 22\mu H$$



Inductor – ripple current

Comparing different inductor values



ripple range
20-50%

$$\Delta I_{peak} = 0,3A$$

higher ripple current \rightsquigarrow higher losses (AC)

Inductor - Rated current

- current load for power inductor can be calculated by

→ software

→ calculation step-by-step

→ use following approach as a simplified calculation

→ BUCK

$$I_{RMS_{inductor}} \approx I_{out_{application}}$$

→ BOOST

$$I_{RMS_{inductor}} \approx \frac{U_{out}}{U_{in}} \cdot I_{out_{application}}$$

Inductor – Saturation current

Buck -Regulator:

- I_{peak} Inductor

$$I_{L\text{max}} = I_{\text{out}} \cdot \left(1 + \frac{I_{\text{rip}}}{2}\right)$$

Boost-Regulator:

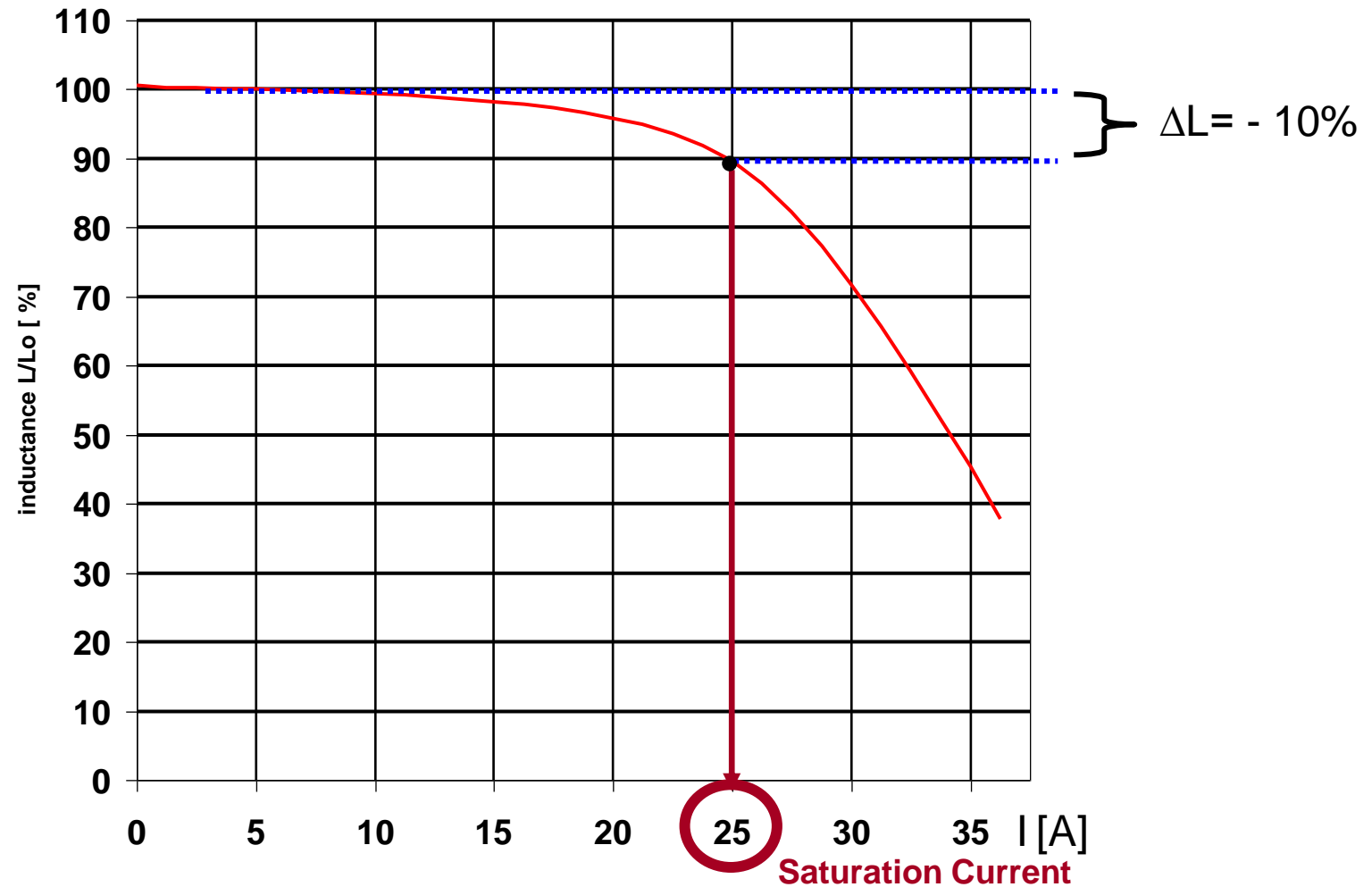
- I_{peak} Inductor

$$I_{L\text{max}} = \frac{I_{\text{out}}}{1 - DC} \cdot \left(1 + \frac{I_{\text{rip}}}{2}\right)$$

Inductor should be not saturated

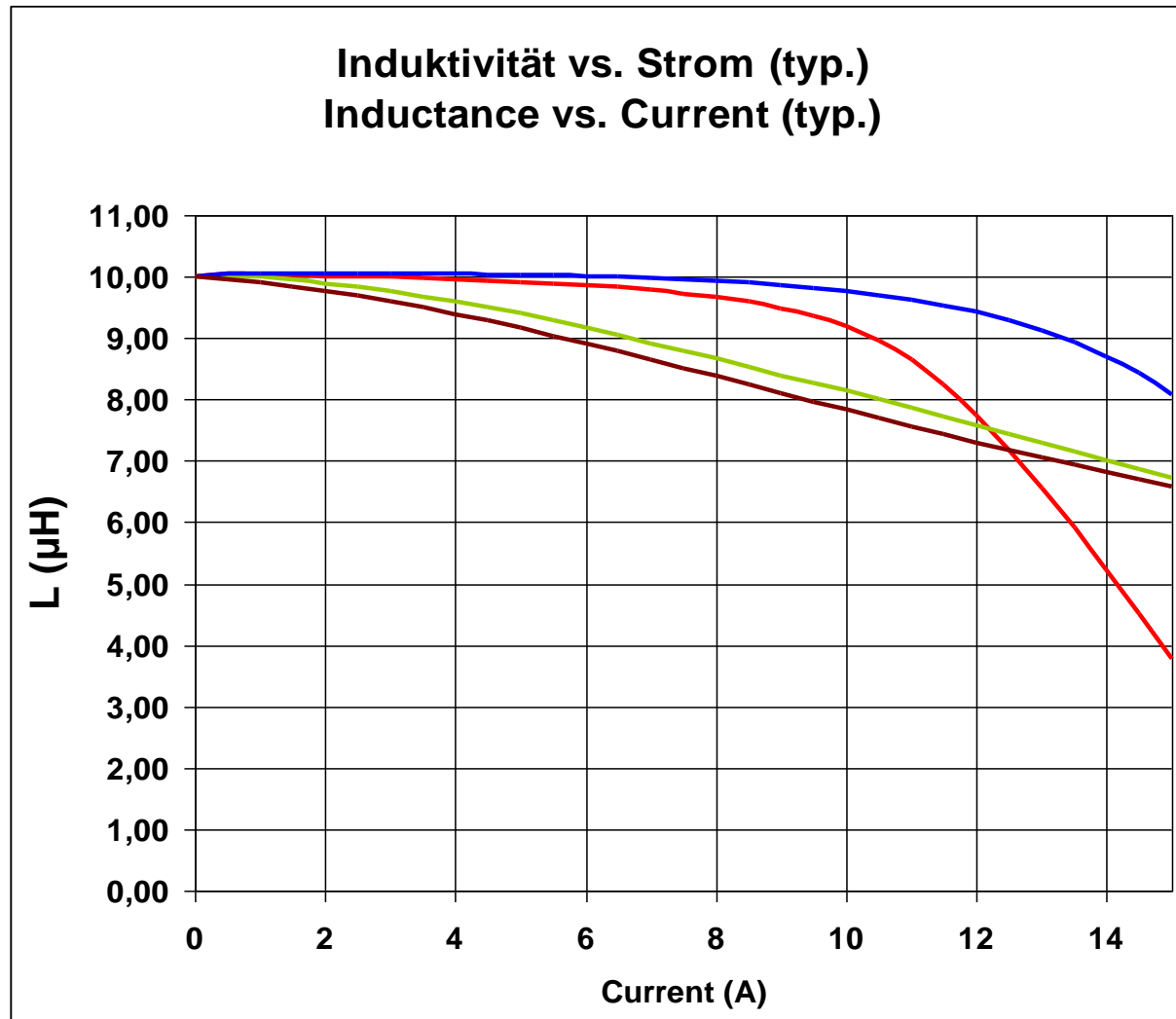
Definition of saturation currents

Definition
Würth Elektronik:
e.g. WE-PD

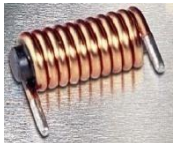


- the saturation current always refers to a certain inductance drop and is individually

What is saturation current?



WE-SI



WE-SD



WE-HCI



WE-PD

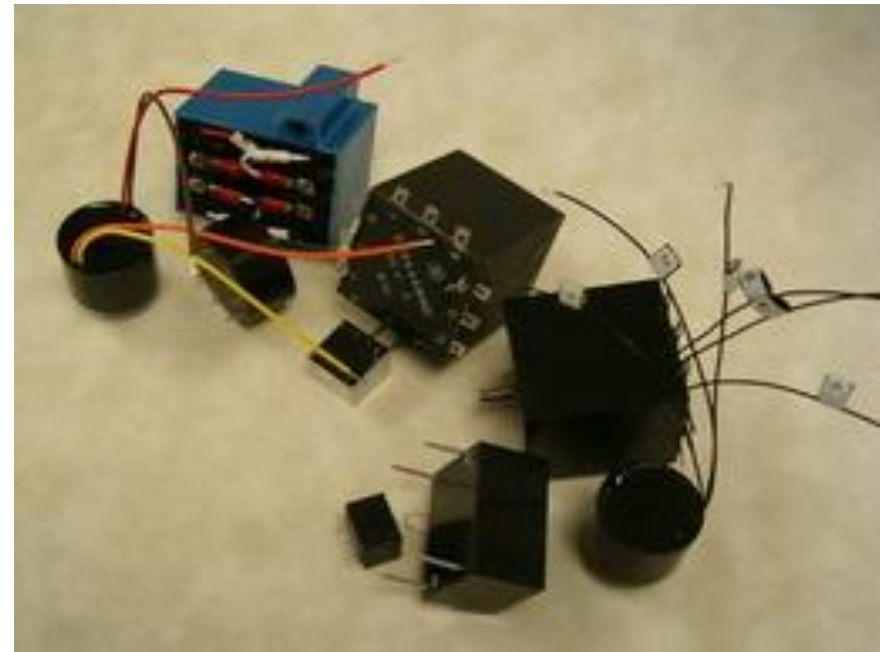


LIVE EMC DEMONSTRATION



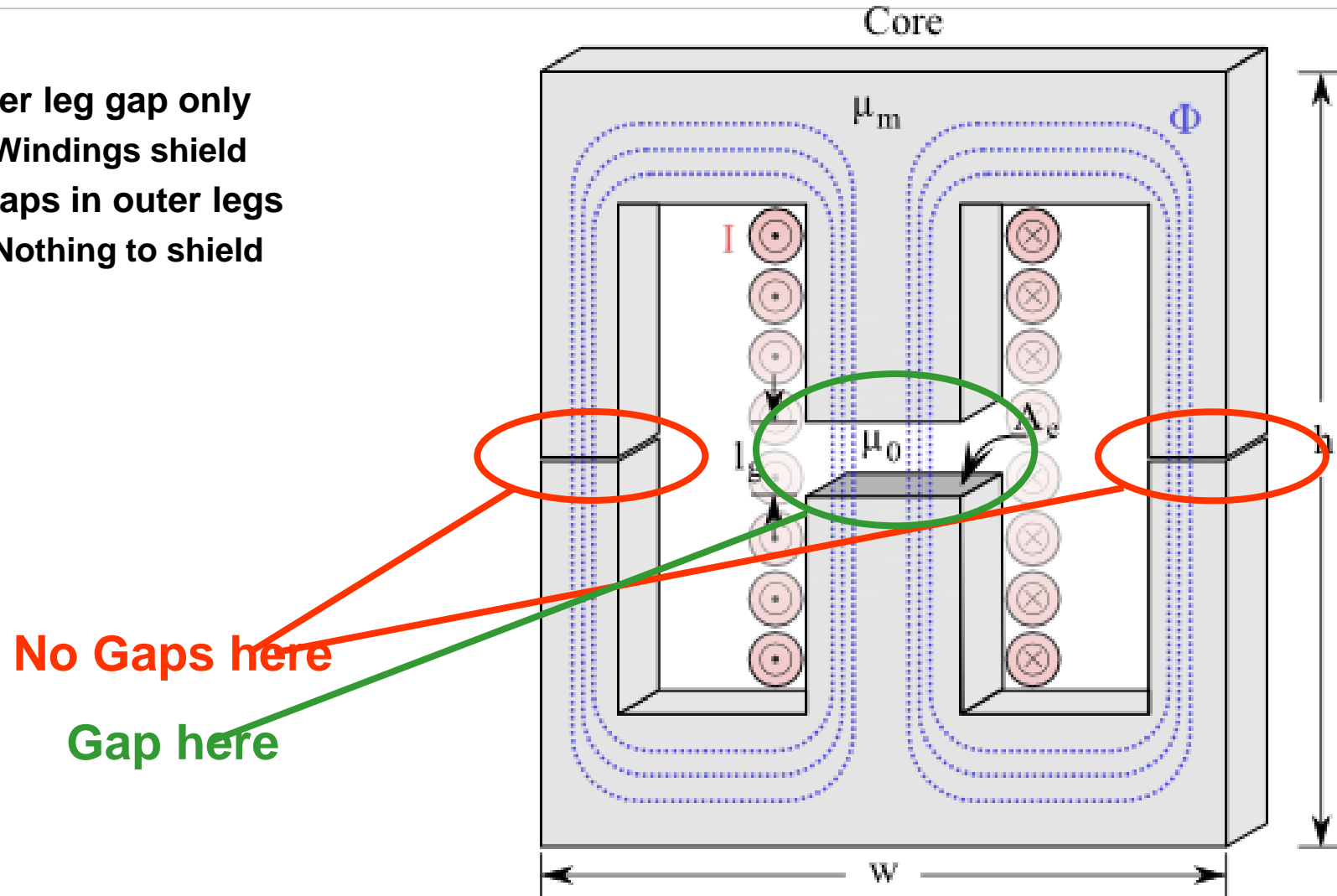
AC/DC CONVERTER EMI

Transformers for EMC – What to choose?



Transformers for EMC – No external gaps

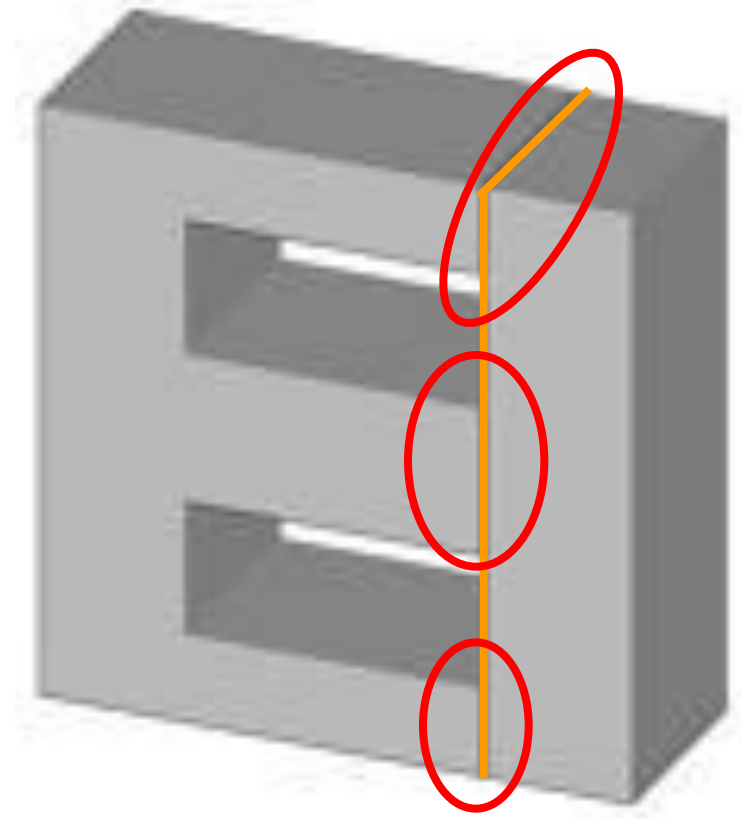
- Center leg gap only
 - Windings shield
- No gaps in outer legs
 - Nothing to shield



Transformers for EMC – No EI core

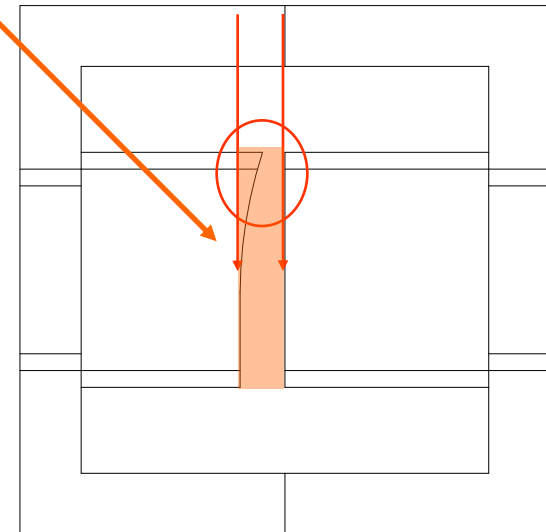
- EI core style
- Mylar or tape used for gap
- Three unshielded gaps

Not a good solution!



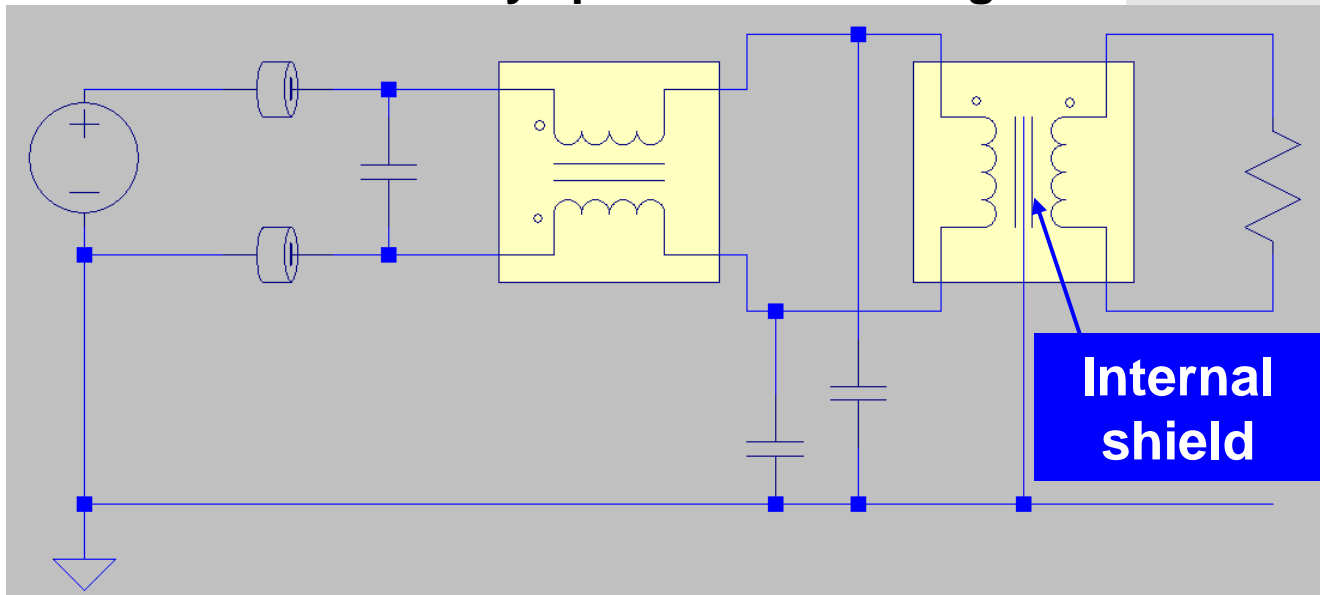
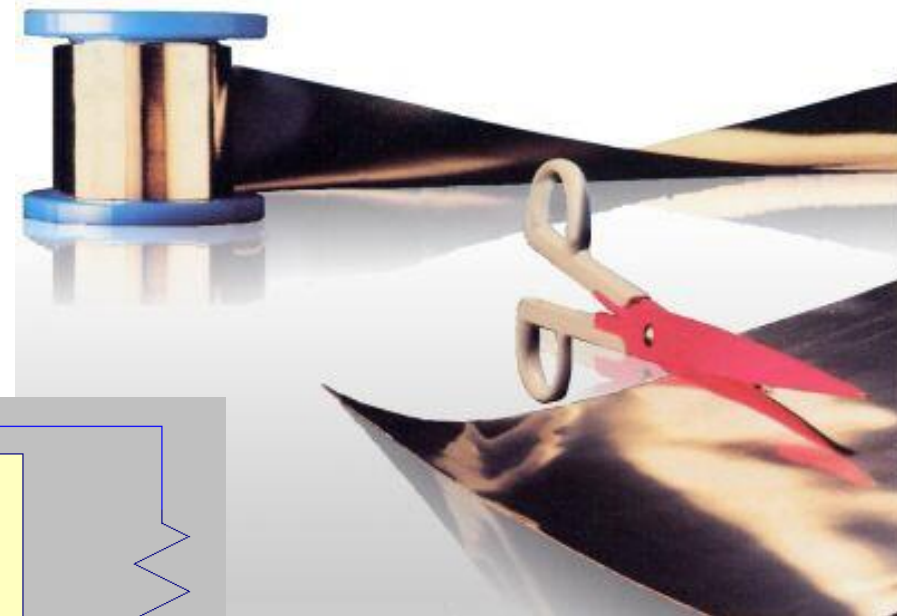
Transformers for EMC – Gap

- **Gap must be perpendicular to flux lines**
 - Here only one side is gapped
- **Uneven gaps are inefficient. => Why?**
 - Core saturates at minimum gap.
 - Requires a larger gap
- **Also larger gap – More potential EMI**



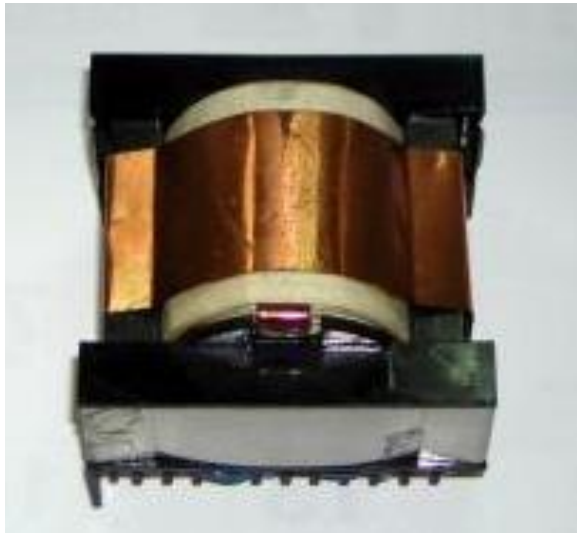
Transformers for EMC – Internal shields

- Shield both conducted and radiated noise
- **Copper foil or wound magnet wire?**
- Copper foil shields – Expensive, => **Why?**
 - Must build shield
 - Must be covered with tape
 - Winding machine stopped to apply
- All shields take away space from winding area

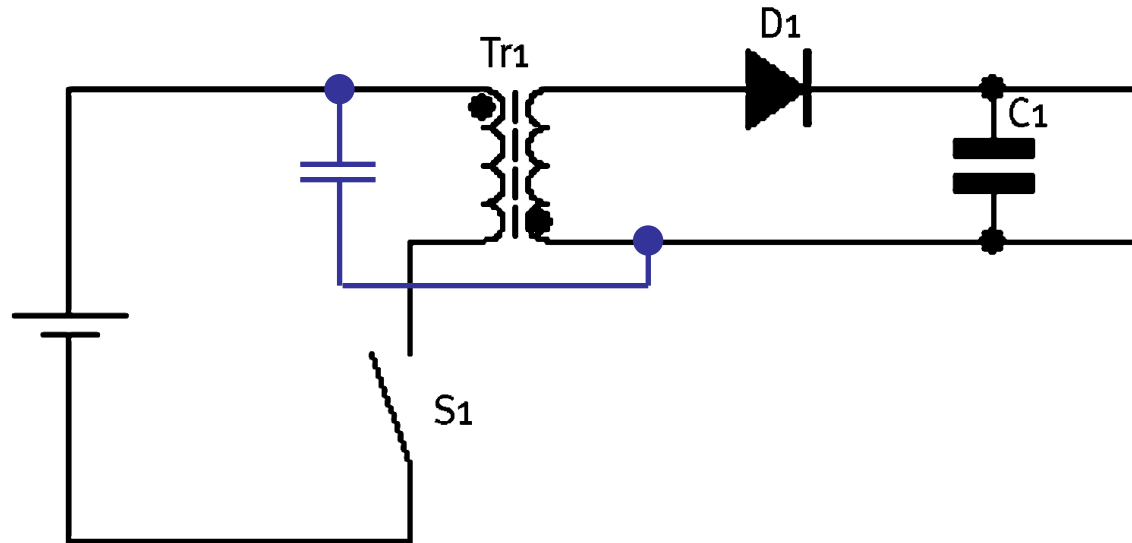


Transformers for EMC – External shields

- How do external shields differ from internal shields?
- Shield radiate noise only!
- As expensive as internal shields



Transformers for EMC – Y-Cap termination



- Noise couples through the transformer via C_{ww}
 - Noise seeks path to primary circuit
 - Without path, noise may become conducted emissions
- Y-Cap across transformer reduces noise
 - Tune the capacitor for optimum loss vs. noise reduction
 - Capacitor usually in the 470pF to 4.7nF range
 - Y-Caps to transformer terminals not on switch nor on diode
 - Close to transformer as possible

What Can We Do?

Decrease C_{ww} ?

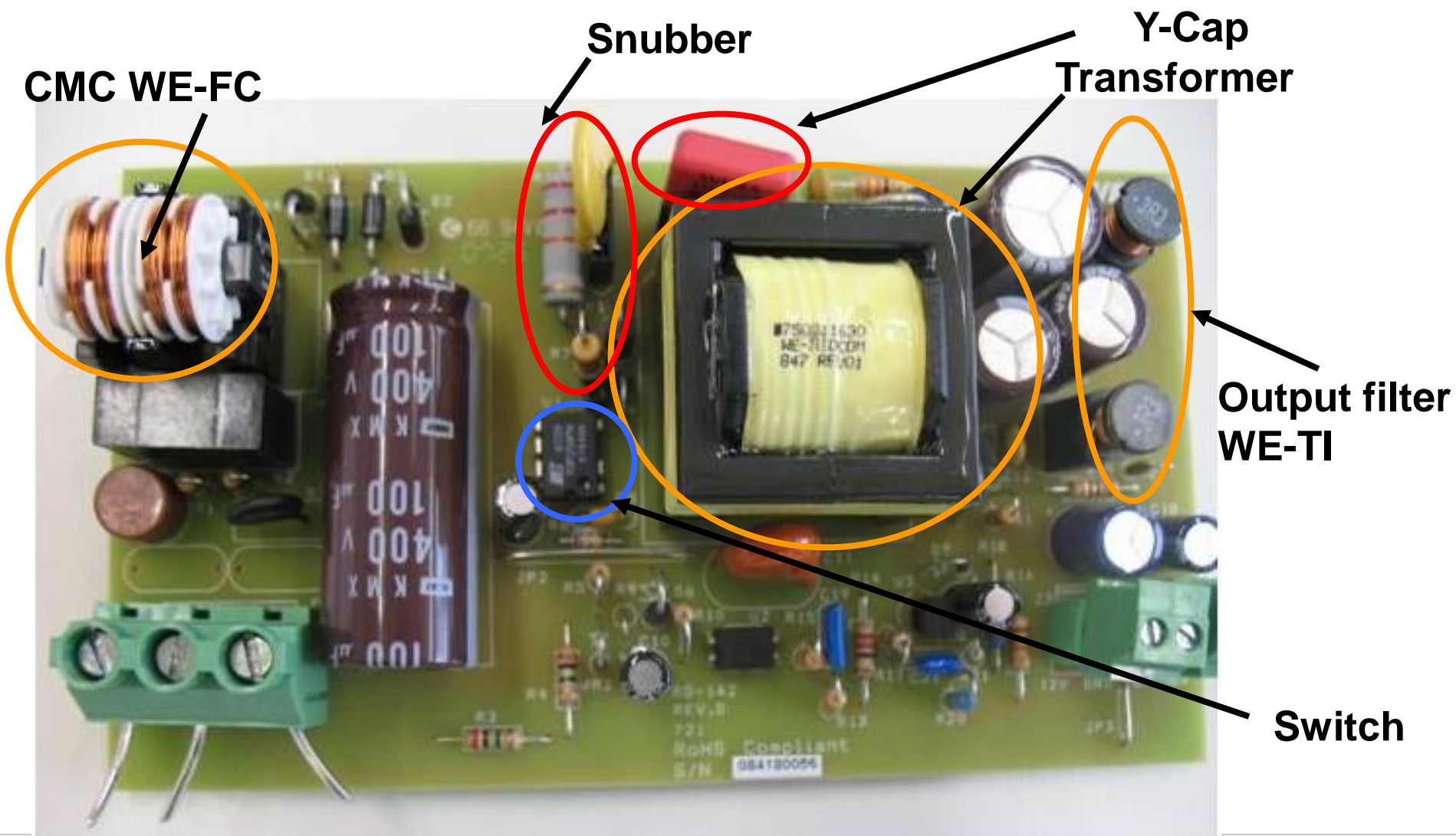
What Else Can We Do?

Transformers for EMC – Reducing C_{ww}

- High C_{ww} causes conducted emissions
- May reduce C_{ww} , but what happens?
- Leakage inductance increases
- L_{LKG} can be controlled by Snubber but efficiency and cost suffer
- Balance between C_{ww} and L_{leakg}

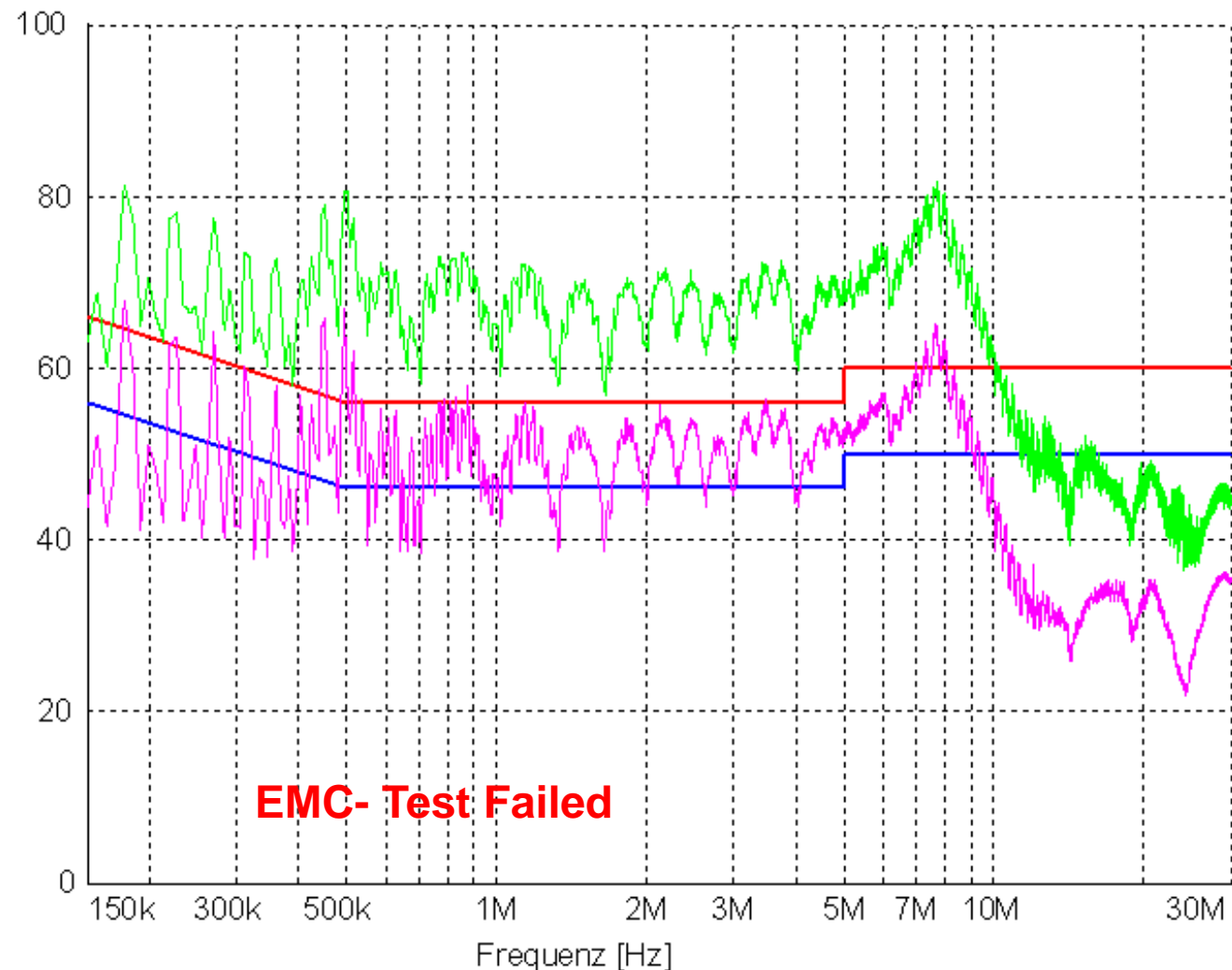


Transformers for EMC – Power Supply



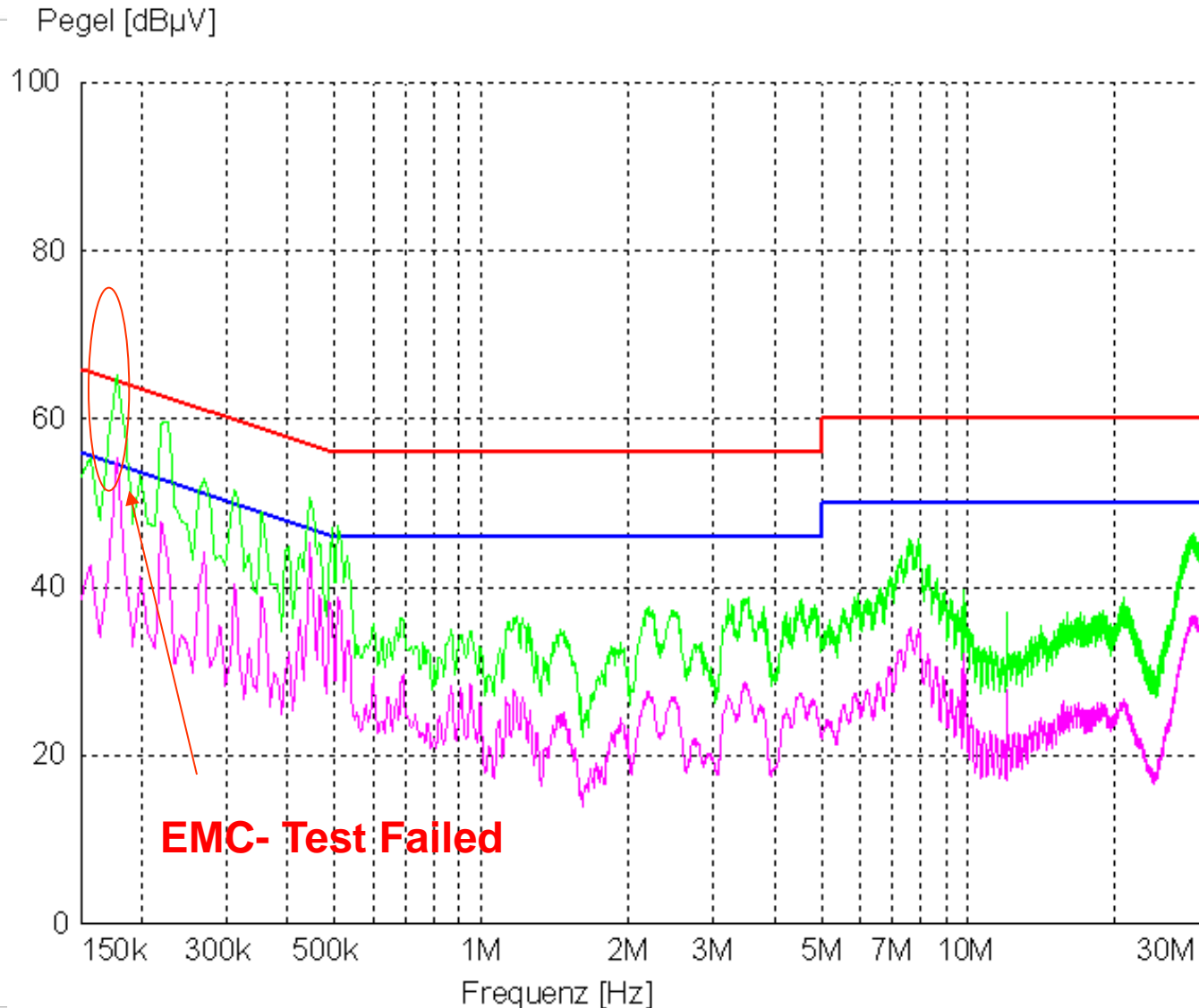
Transformers for EMC – Example 1

Pegel [dB μ V]



- Without common mode choke
 - With adjusted Snubber
 - Without adjusted Y-Cap
- QPeak
- Avg.
- Peak
- Avg.

Transformers for EMC – Example 2



- With common mode choke
- With adjusted Snubber
- Without adjusted Y-Cap

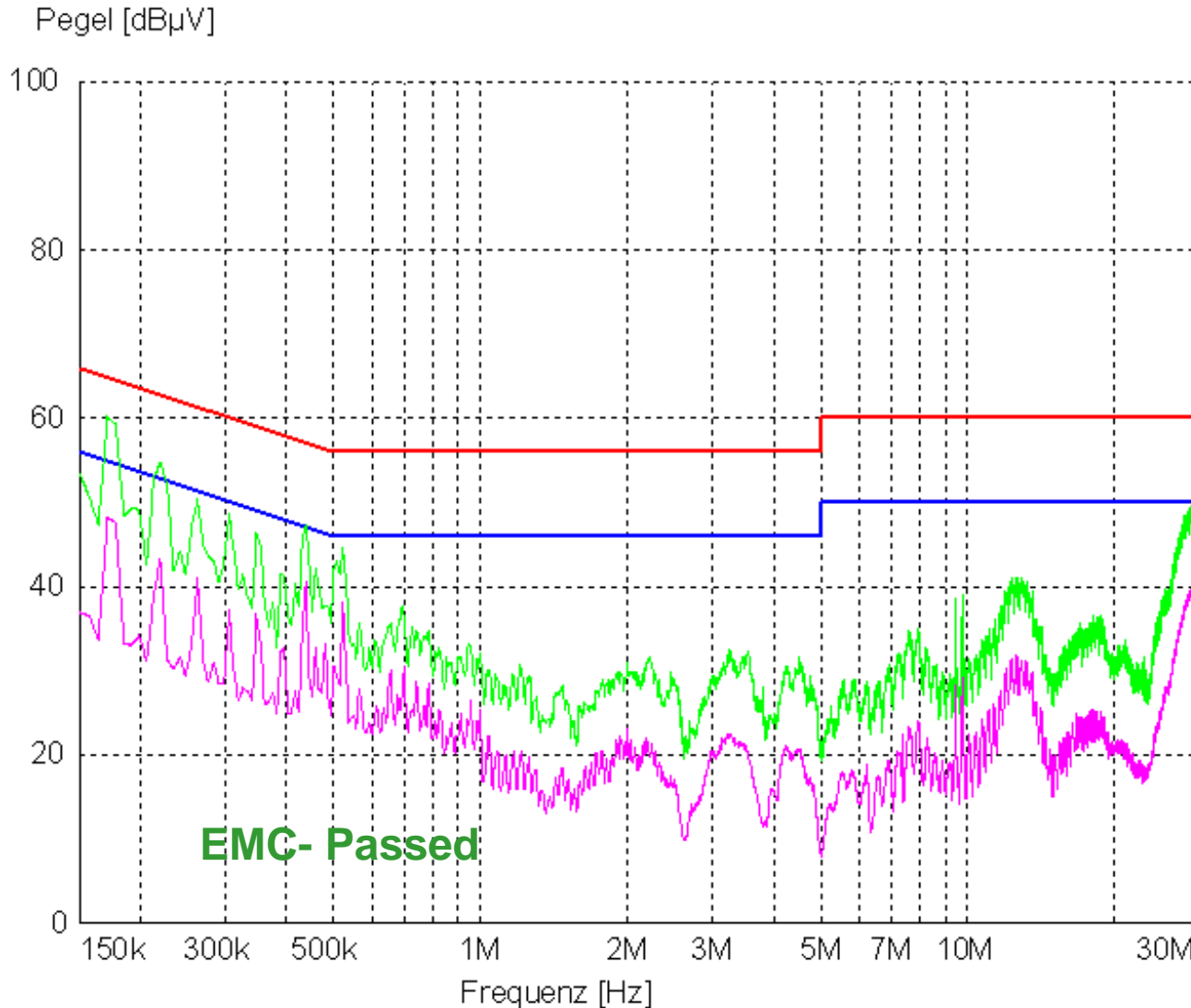
QPeak

Avg.

Peak

Avg.

Transformers for EMC – Example 3



- With common mode choke
- With adjusted Snubber
- With adjusted Y-Cap

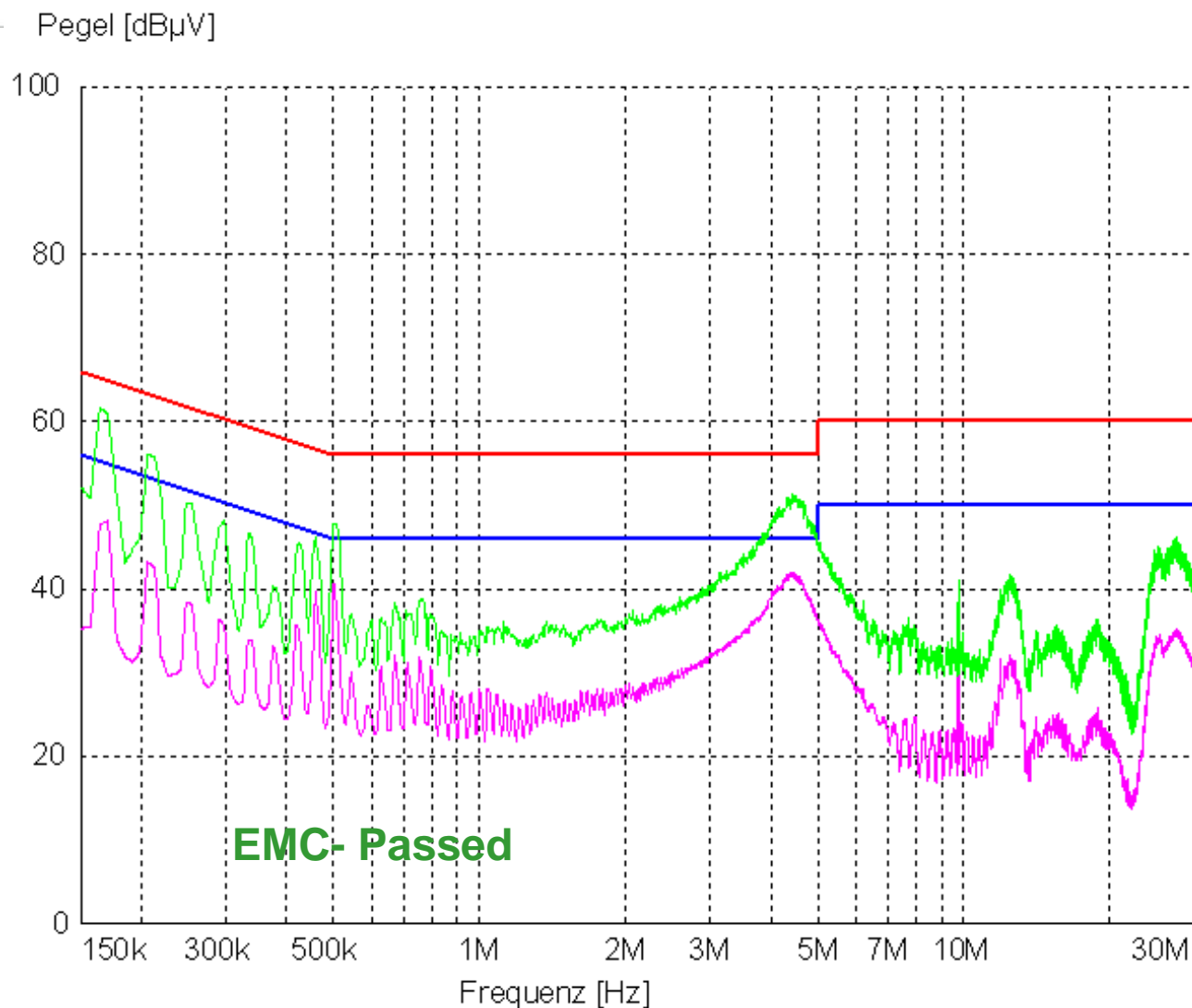
QPeak

Avg.

Peak

Avg.

Transformers for EMC – Example 4



- With common mode choke
- Without adjusted Snubber
- With adjusted Y-Cap

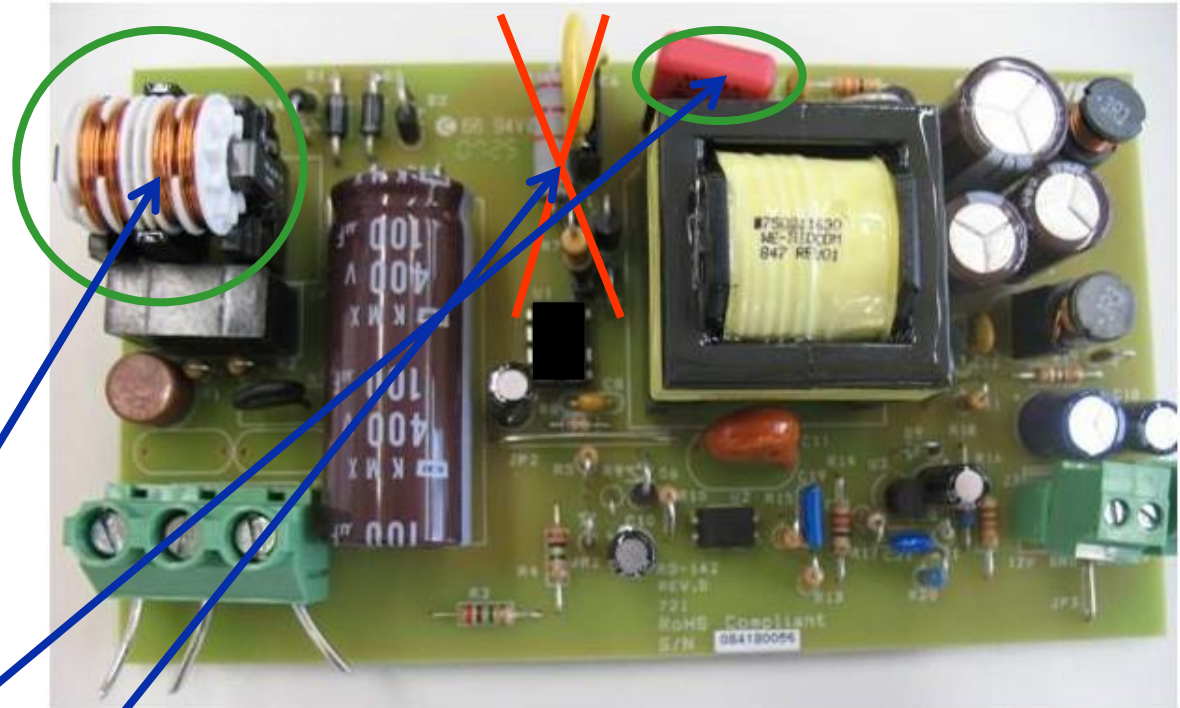
QPeak

Avg.

Peak

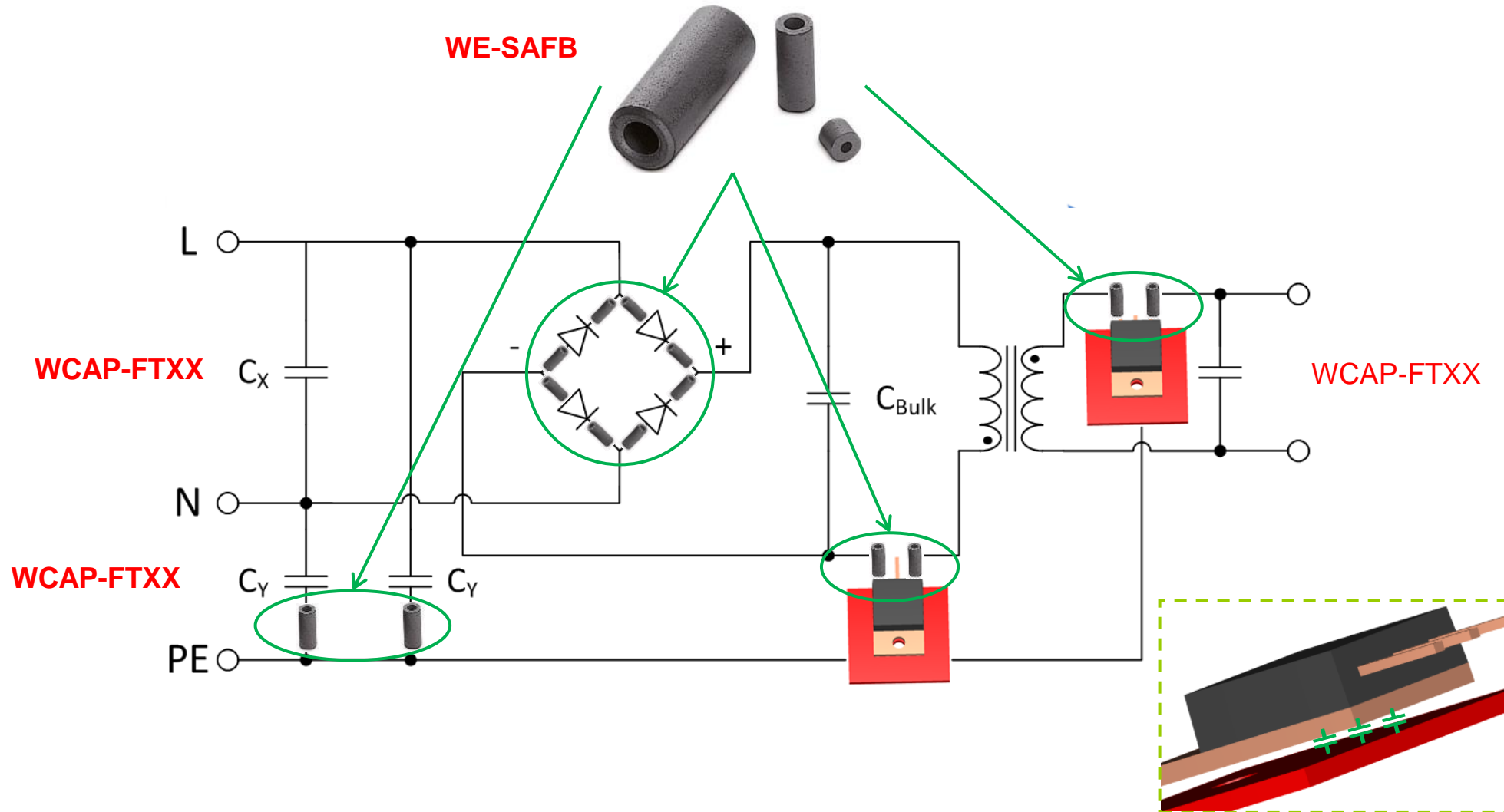
Avg.

Transformer for EMC – Conclusion for this power supply



- Necessary to pass EMI:
 - Common Mode Choke (CMC)
 - Y-Cap
- Not necessary to pass EMI
 - Optimized Snubber

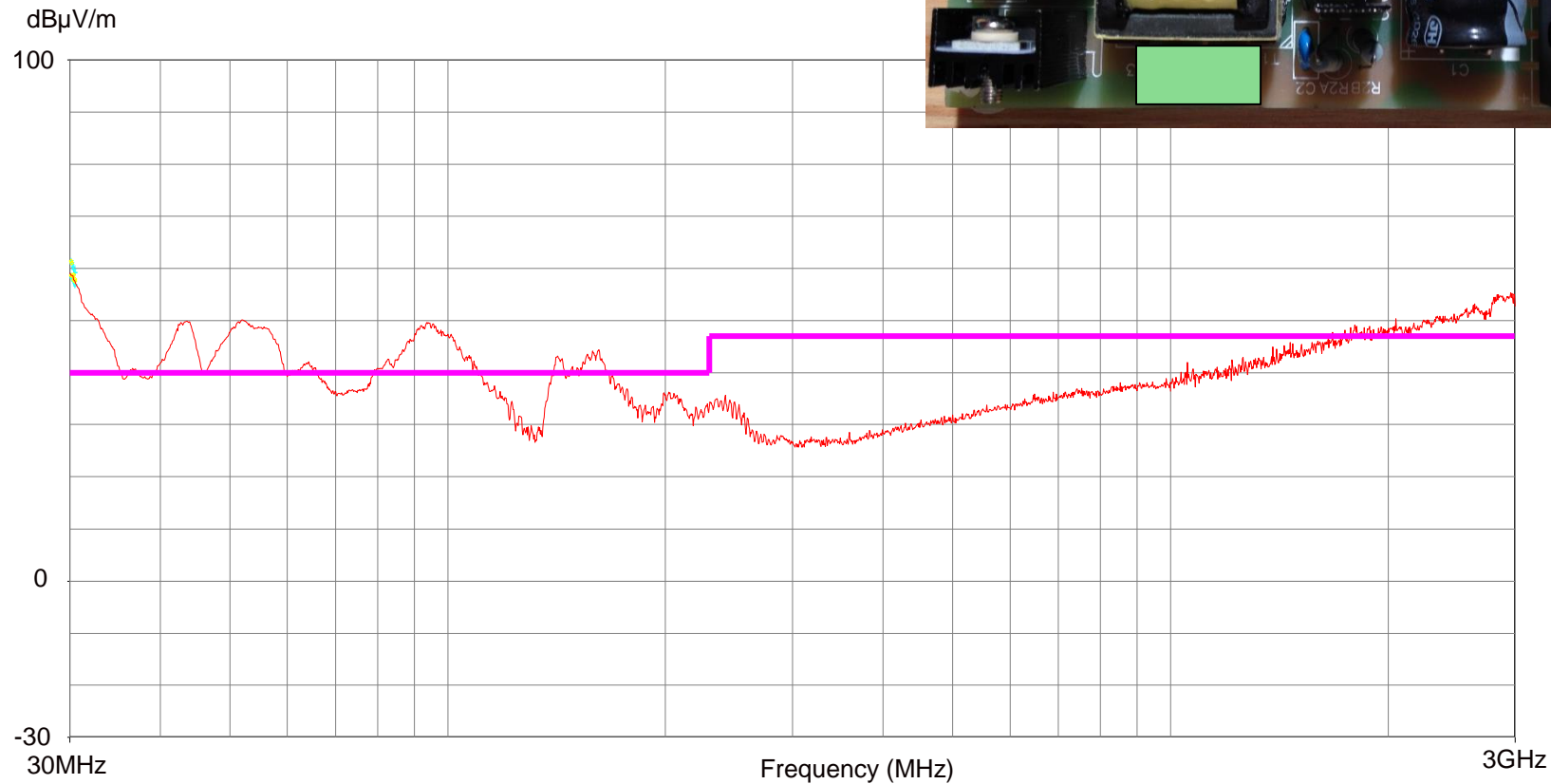
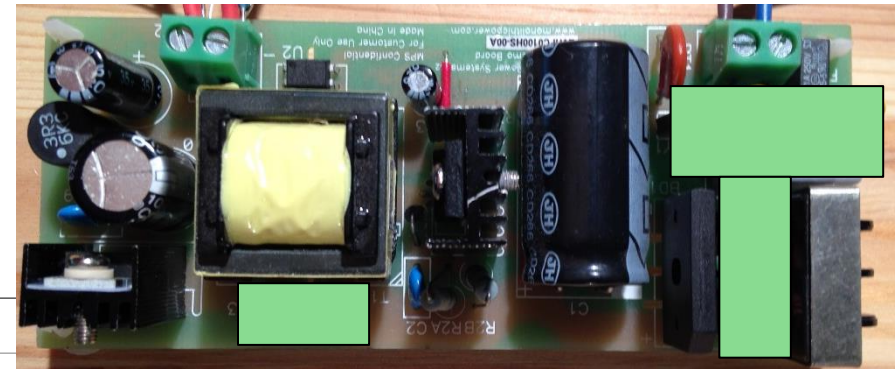
Common Mode Noise Suppression



Radiated Emissions made by AC-DC Converter

No Filter- no Y -Cap

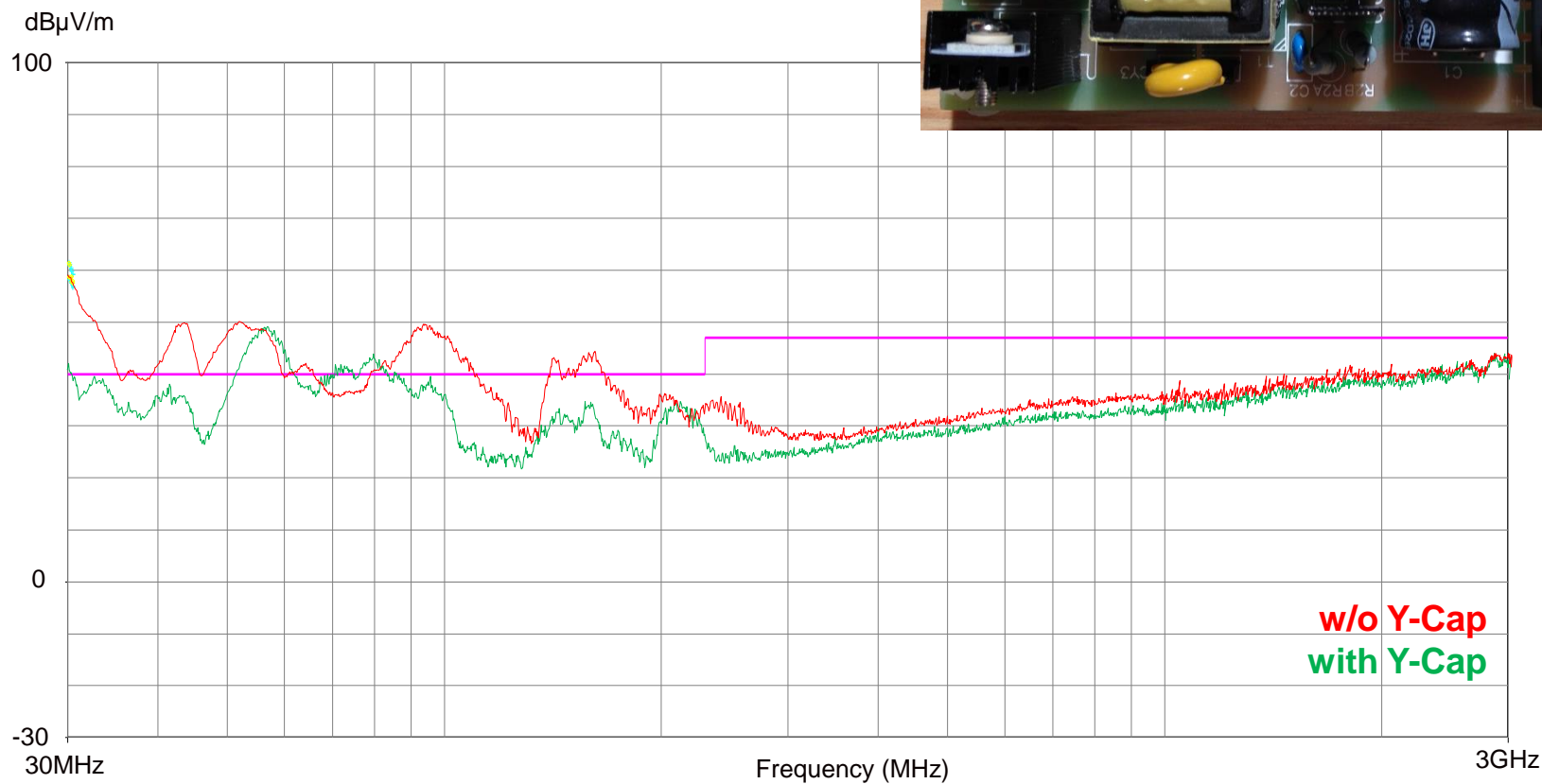
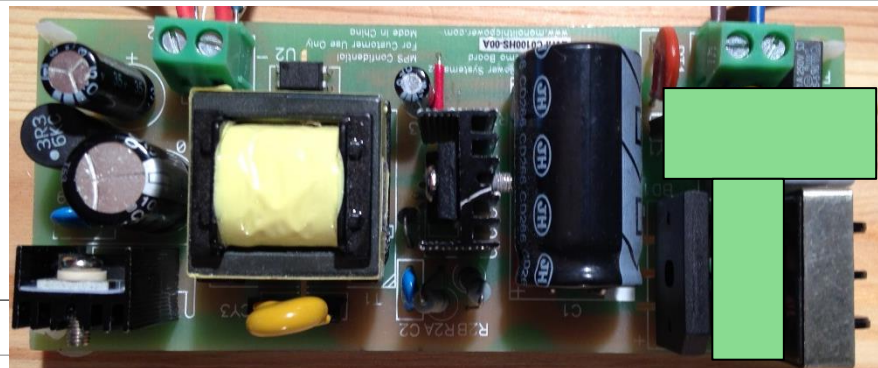
- **U_{in}:** 230Vac
 - **I_{out}:** 1,5A
 - **Polarization:**
 - **Norm:**
- U_{out}:** 24Vdc
f_{sw}: 100kHz
Horizontal
EN55022A



Radiated Emissions made by AC-DC Converter

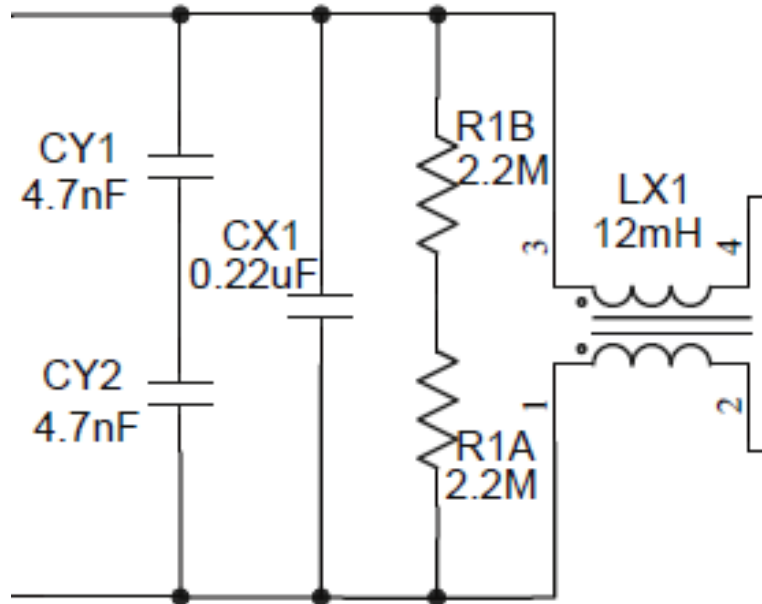
No Filter- using Y -Cap

- U_{in} : 230Vac
 - I_{out} : 1,5A
 - Polarization:
 - Norm:
- U_{out} : 24Vdc
 - fsw: 100kHz
 - Horizontal
 - EN55022A



Radiated Emissions made by AC-DC Converter

Use Input Filter & Y -Cap



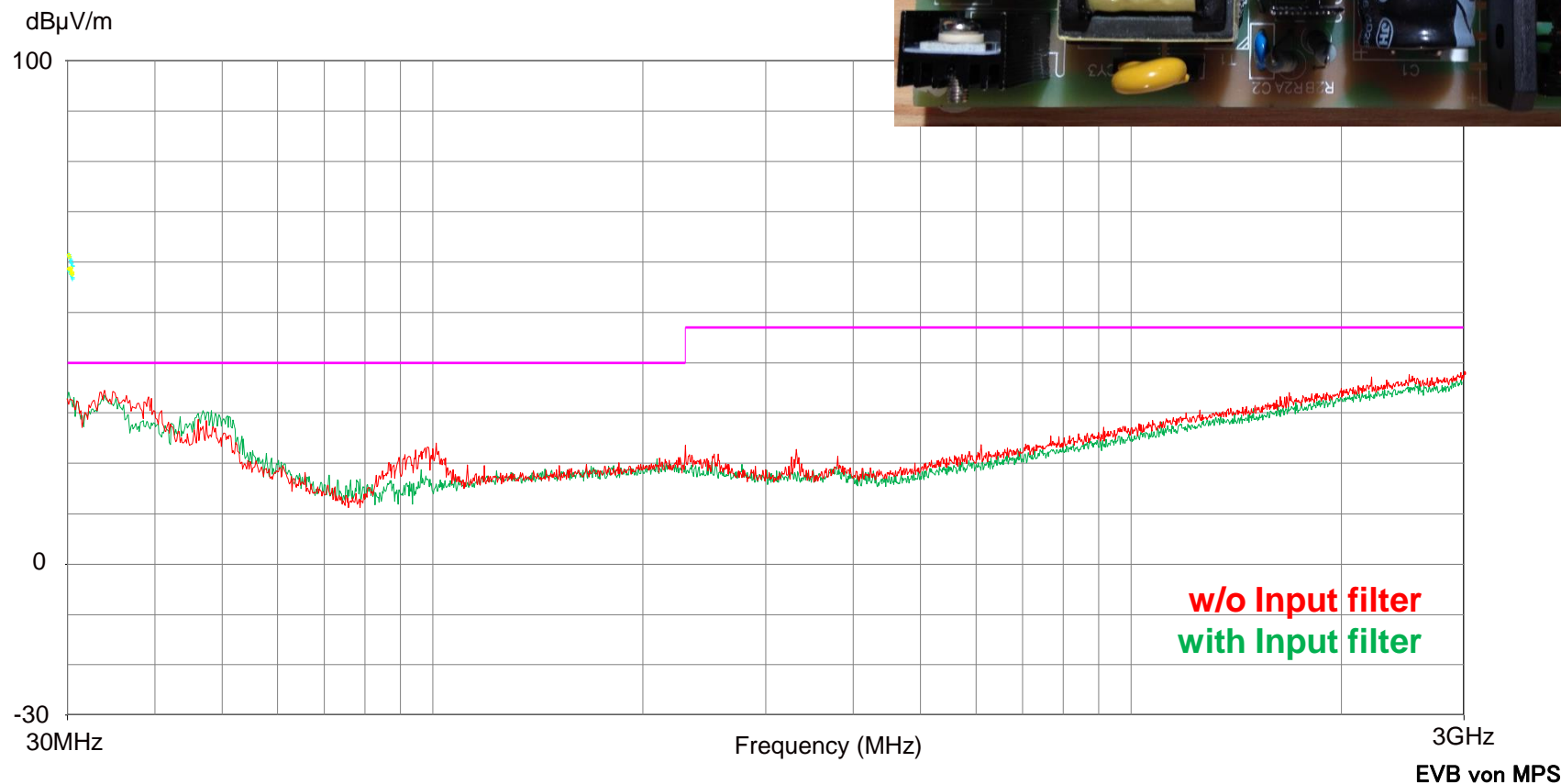
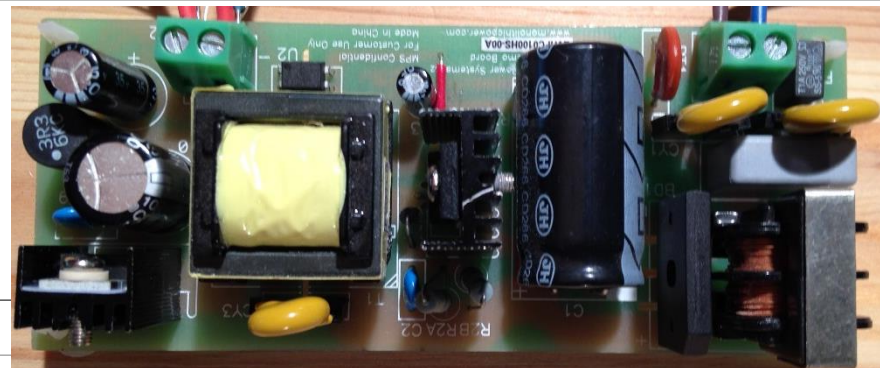
Input Filter



Y- Cap

Radiated Emissions made by AC-DC Converter With Input Filter & Y -Cap

- U_{in} : 230Vac
 - I_{out} : 1,5A
 - Polarization:
 - Norm:
- U_{out} : 24Vdc
 - fsw: 100kHz
 - Horizontal
 - EN55022A

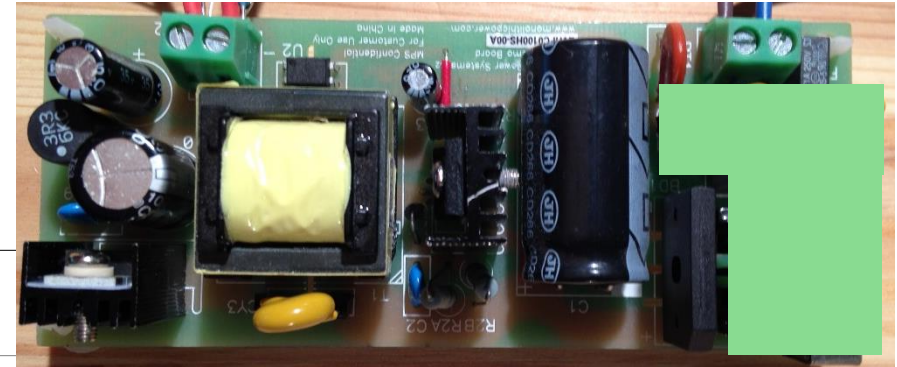
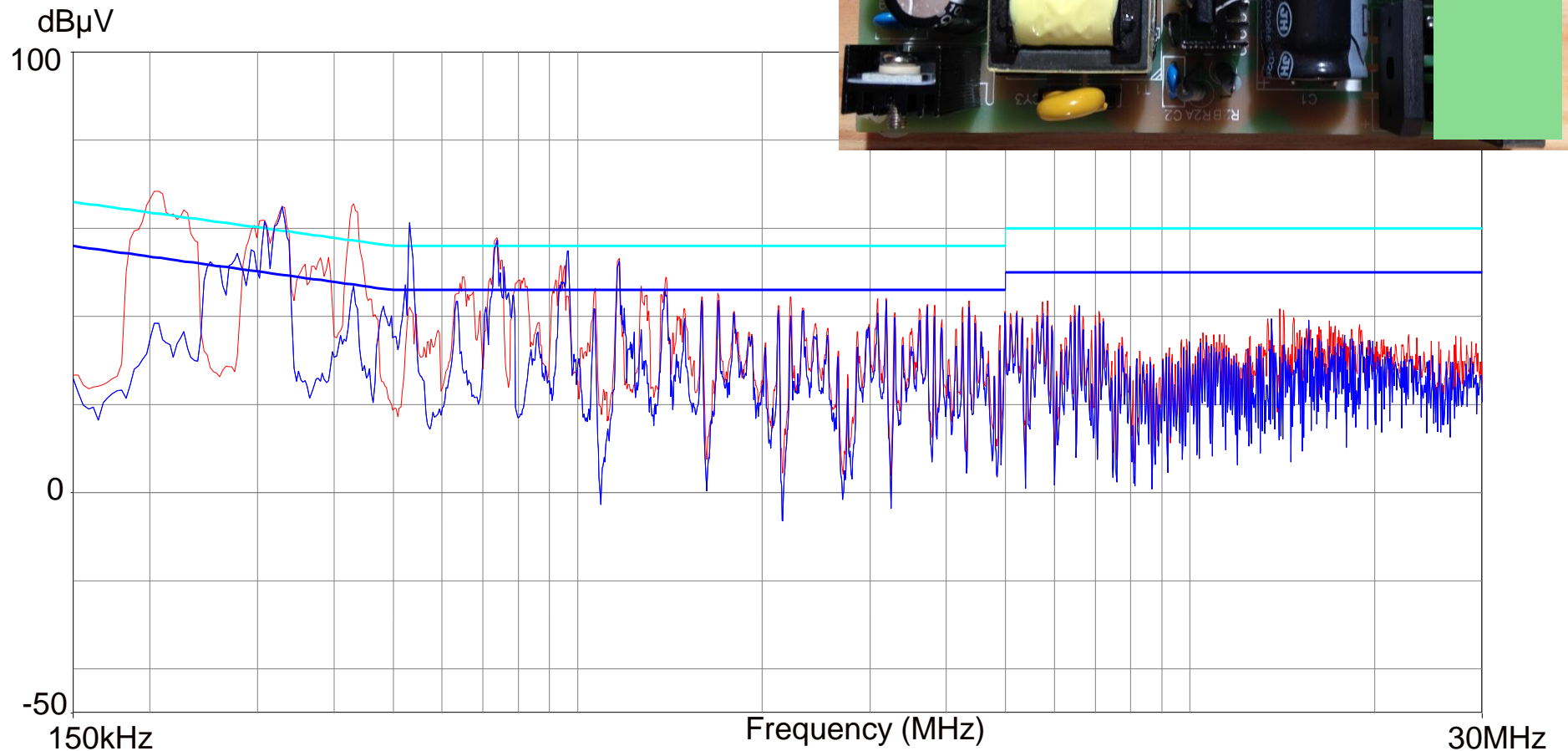




INPUT FILTER

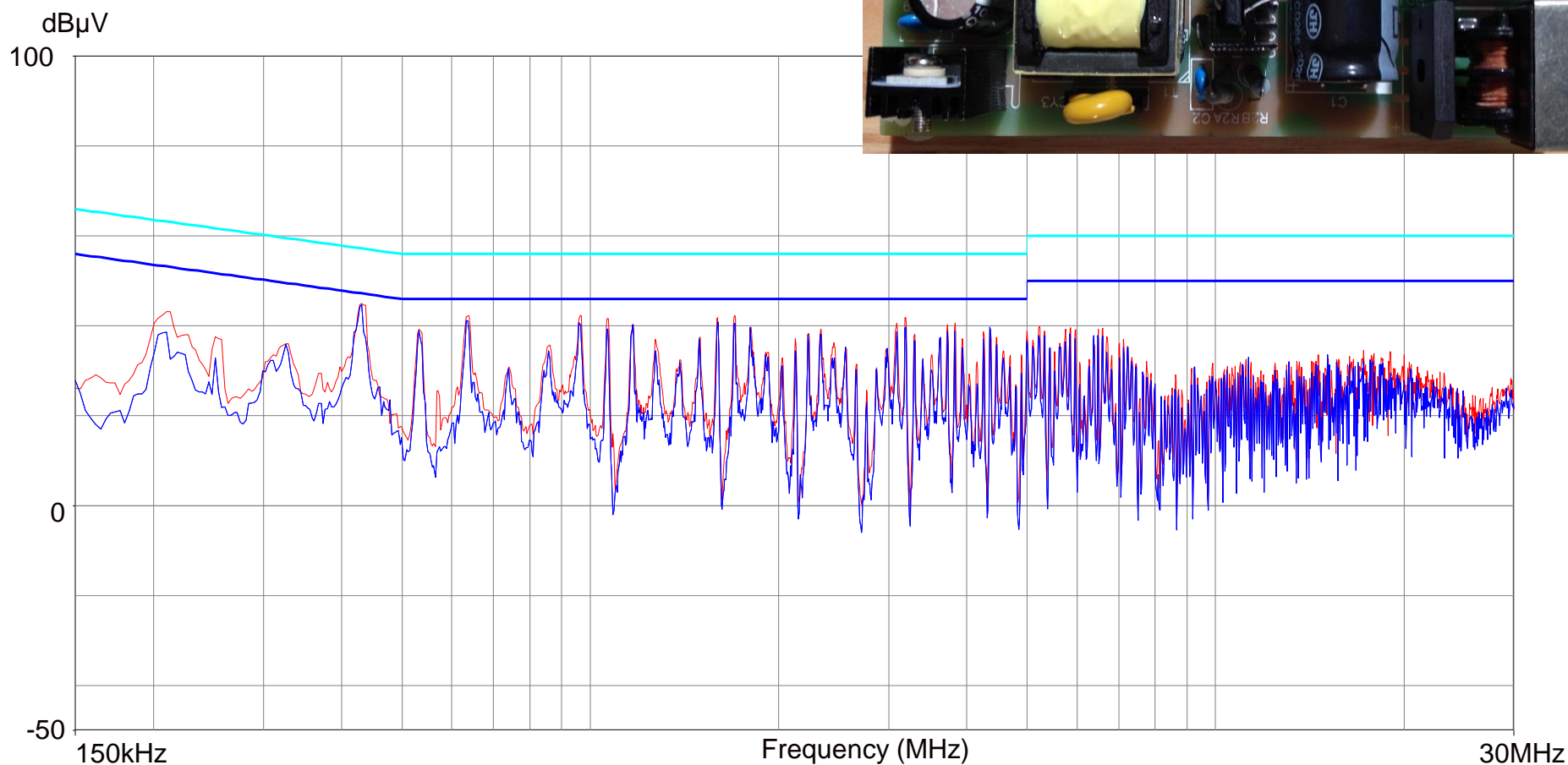
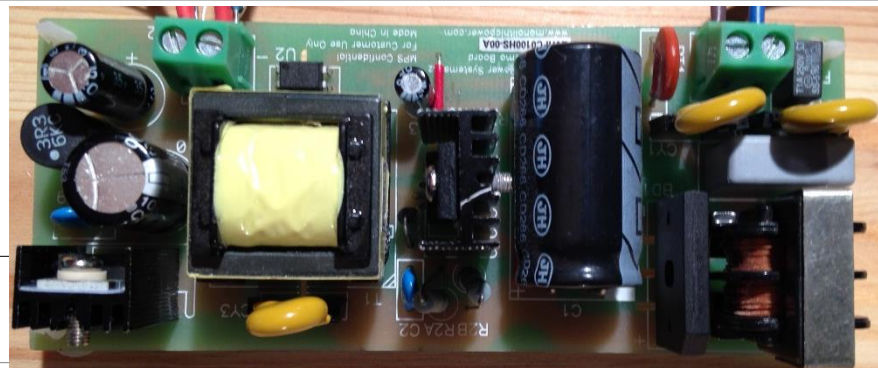
Conducted Emissions made by AC-DC Converter without Input Filter with Y-Cap

- **U_{in}:** 230VAC,
- **I_{out}:** 1,5A,
- **Measured:** L to PE
- **Norm:** EN55022A



Conducted Emissions made by AC-DC Converter with Input Filter & Y-Cap

- U_{in} : 230VAC,
 - I_{out} : 1,5A,
 - Leitung: L nach PE
 - Norm: EN55022A
- U_{out} : 24VDC
 f_{CLK} : 100kHz

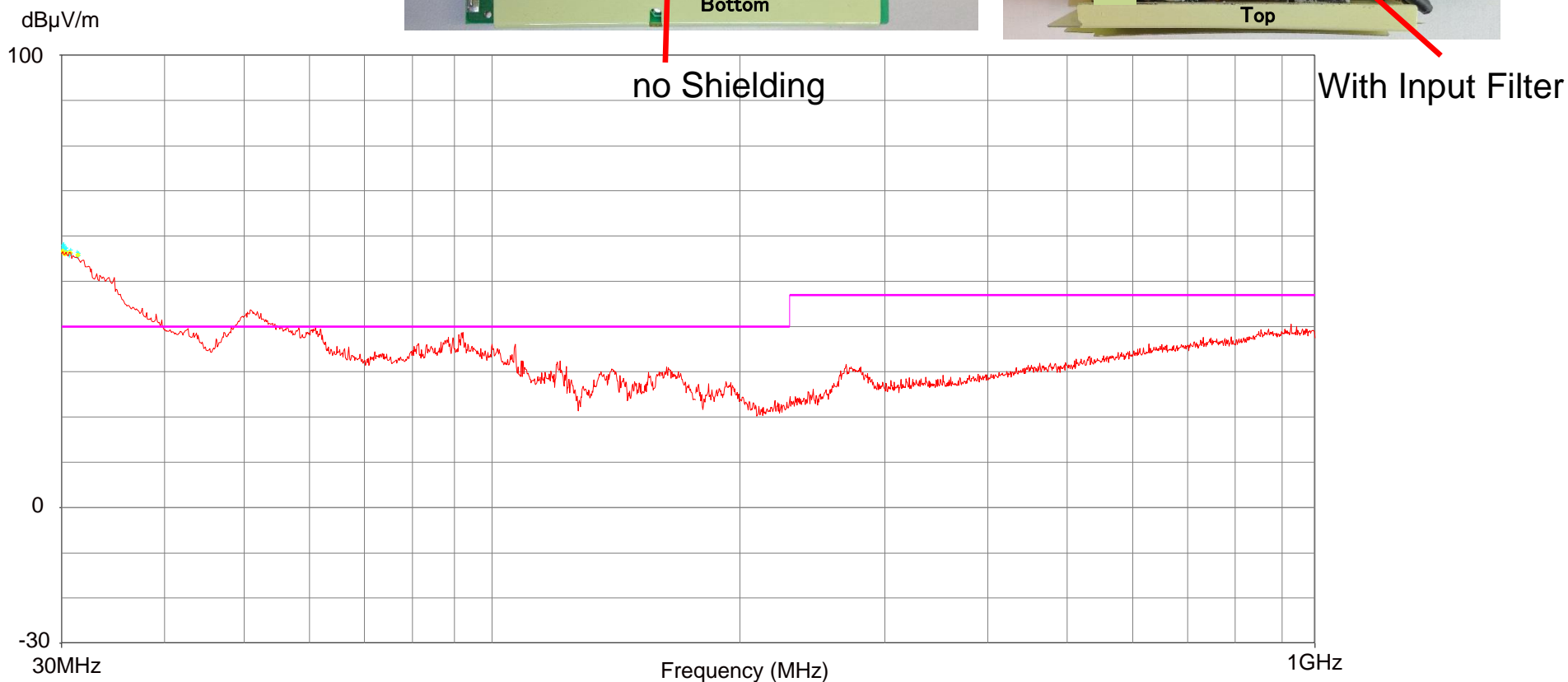
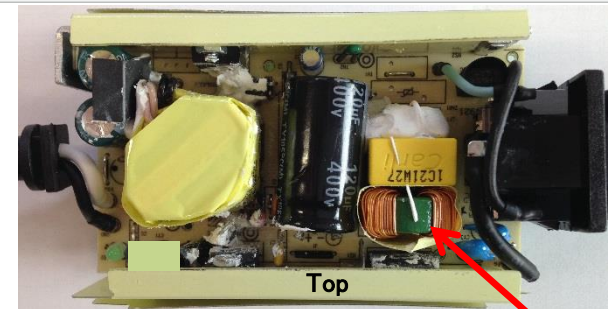
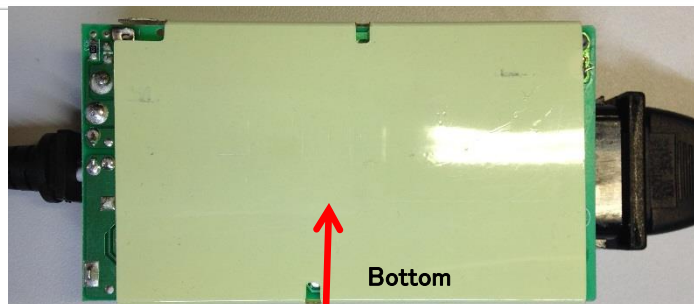




OTHER EMC SITUATION FOR AC/DC CONVERTER

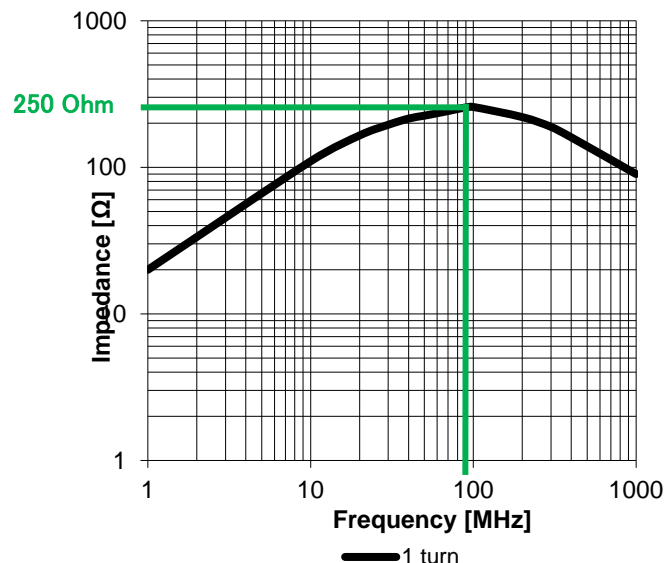
Radiated Emissions made by AC-DC Converter without Ferrite bead and without Y -Cap

- U_{in} : 230VAC
- U_{out} : 12VDC
- I_{out} : 4,16A
- f_{sw} : 90kHz

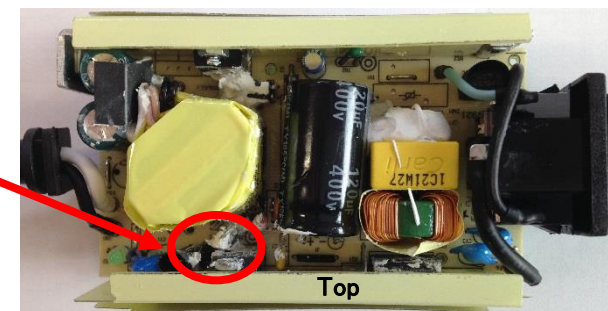
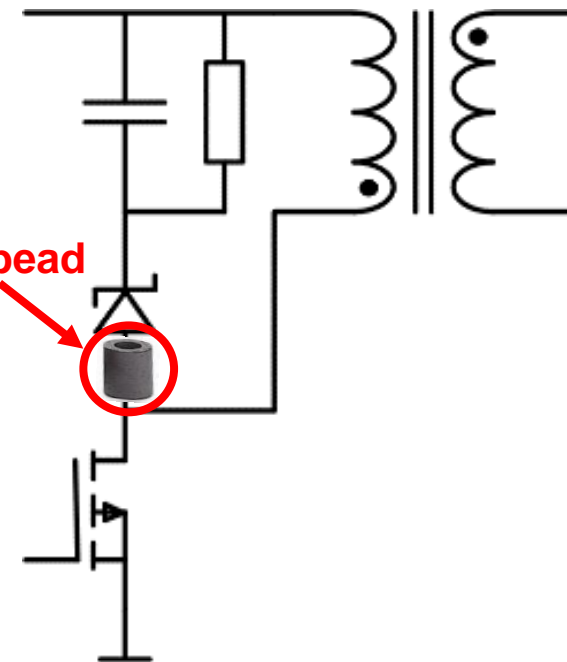


Radiated Emissions made by AC-DC Converter with Ferrite bead and without Y -Cap

- Ferrite bead selection:
 - Check noise frequency
 - NiZn Ferrite bead
 - use WE-SAFB 4x2 mm, 250Ohm @ 90 MHz



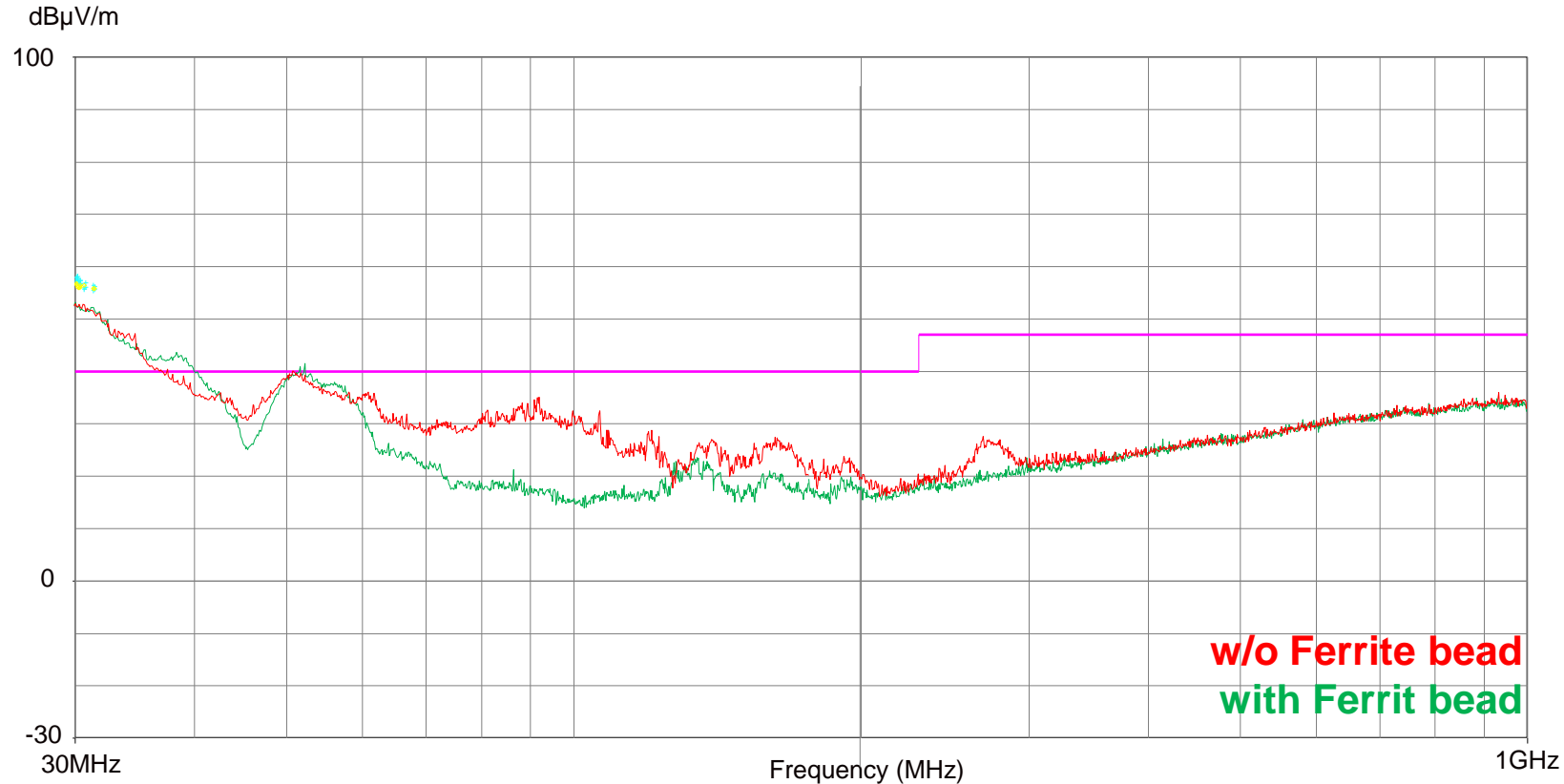
Ni-Zn Ferrite bead



Ni-Zn Ferrite bead

Radiated Emissions made by AC-DC Converter with Ferrite bead and without Y -Cap

- **Uin:** 230VAC,
 - **Iout:** 4,16A,
 - **Polarization:**
 - **Norm:**
- Uout:** 12VDC
 - fsw:** 90kHz
 - Horizontal**
 - EN55022A**

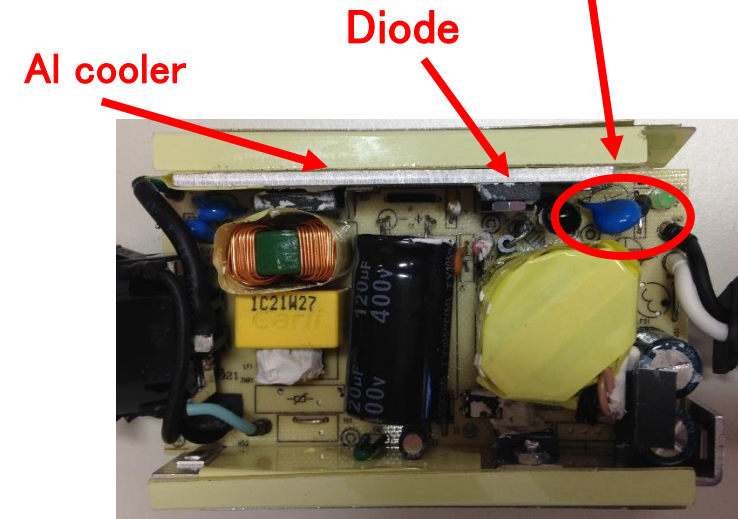
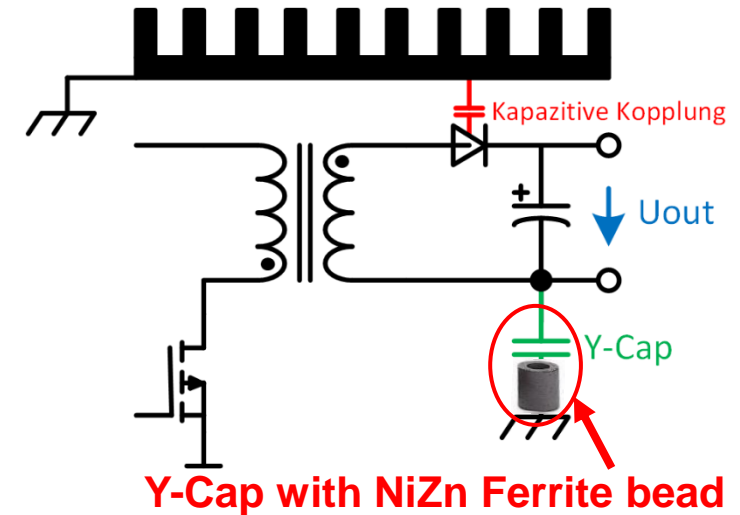


Radiated Emissions made by AC-DC Converter with Ferrite bead and with Y -Cap



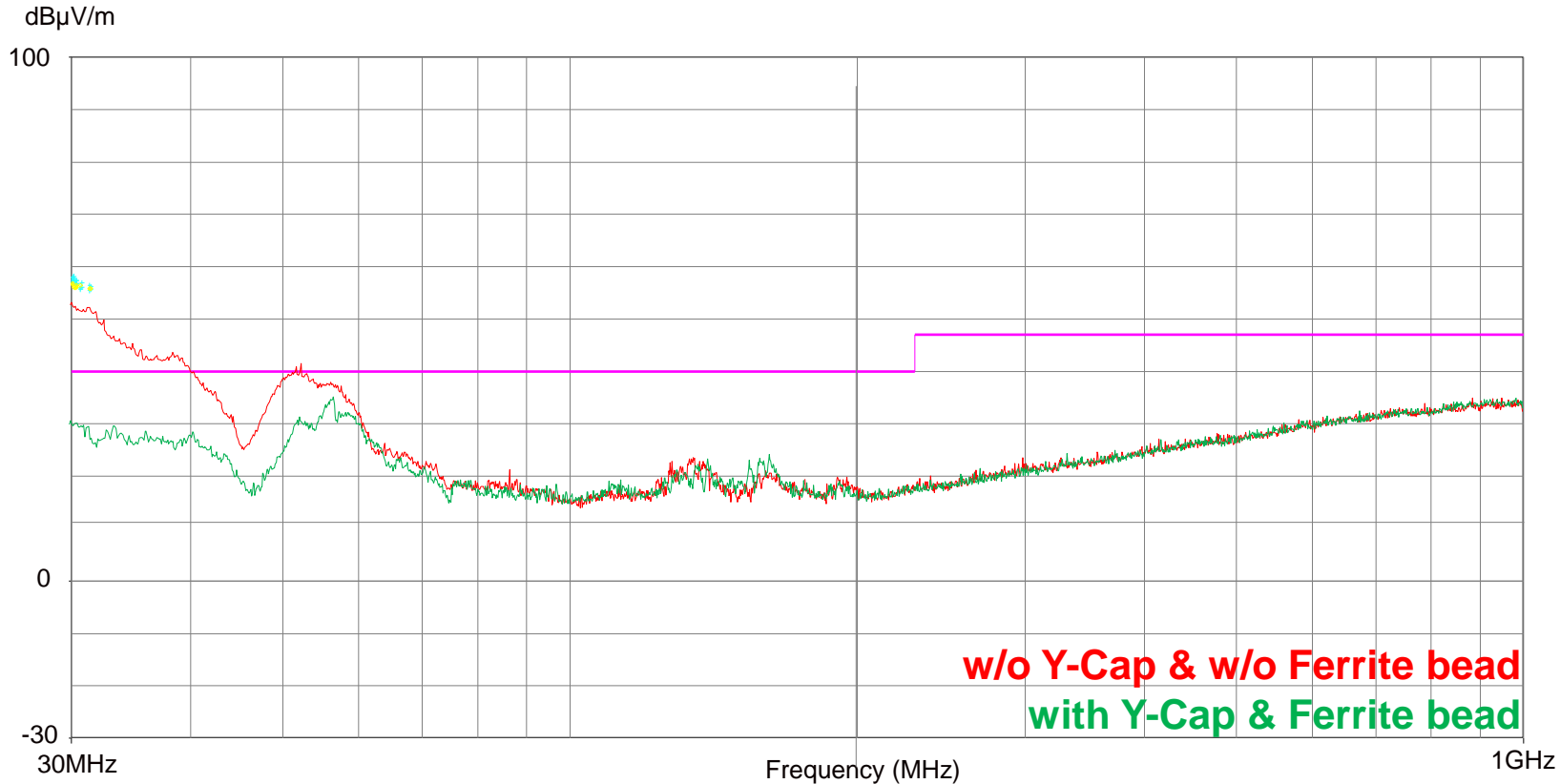
- Selection of Y Cap
 - High freq. type
 - High Voltage
 - Low ESR
 - Small package
 - Example: WCAP-CSSA 1nF

- Selection of Ferrite bead:
 - NiZn Ferrite bead
 - Small size bead
 - Example: WE-SAFB 4x2 mm, 250Ohm @ 90 MHz



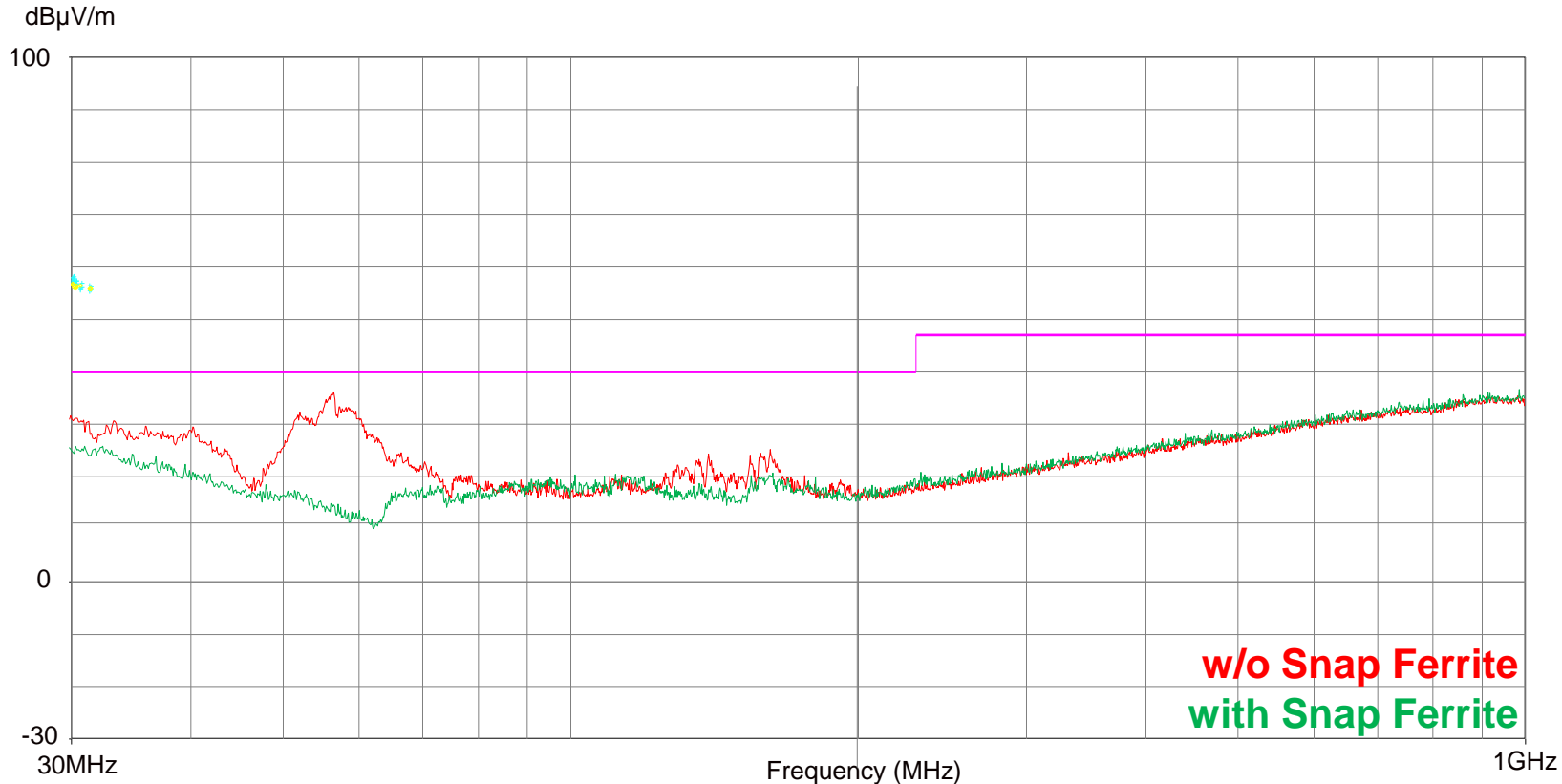
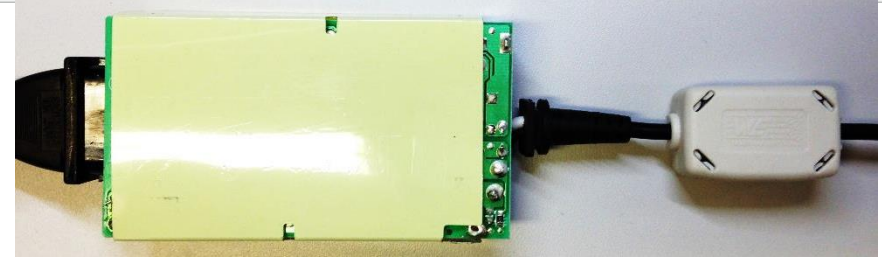
Radiated Emissions made by AC-DC Converter with Ferrite bead and with Y -Cap

- Uin: 230VAC,
 - Iout: 4,16A,
 - Polarization:
 - Norm:
- | | |
|------------|-------|
| Uout: | 12VDC |
| fsw: | 90kHz |
| Horizontal | |
| EN55022A | |



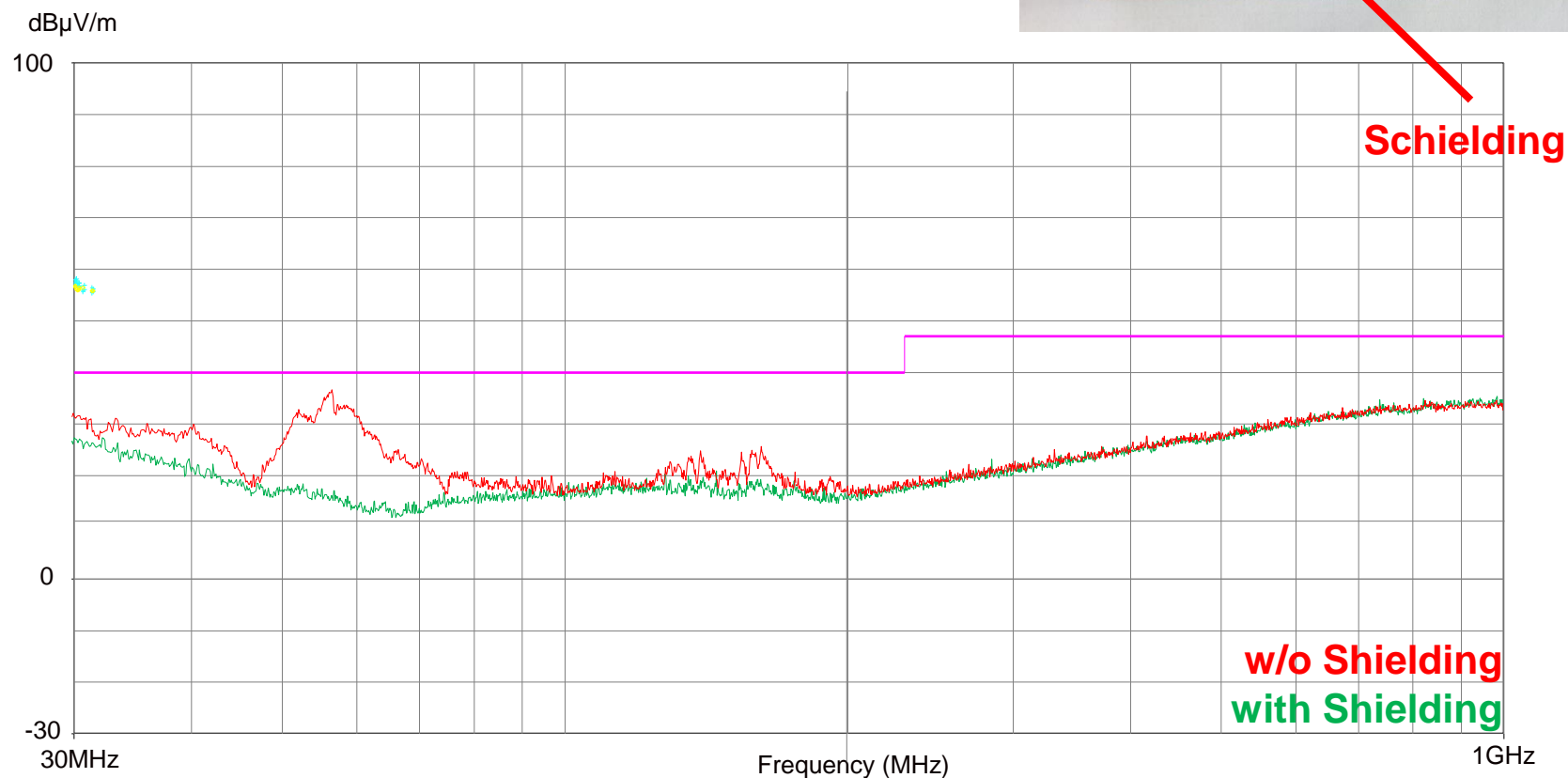
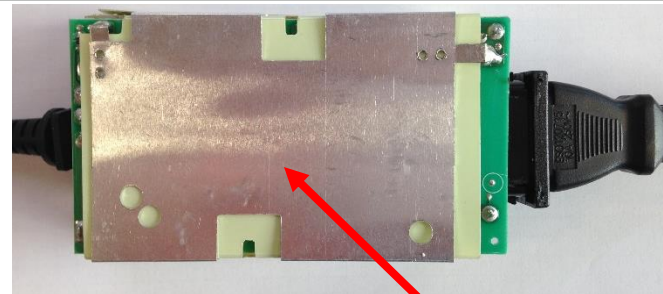
Radiated Emissions made by AC-DC Converter with Ferrite bead and with Y –Caps + Snap Ferrite

- **Uin:** 230VAC,
 - **Iout:** 4,16A,
 - **Polarization:**
 - **Norm:**
- Uout:** 12VDC
fsw: 90kHz
Horizontal
EN55022A



Radiated Emissions made by AC-DC Converter with Ferrite bead and with Y –Caps + Shielding

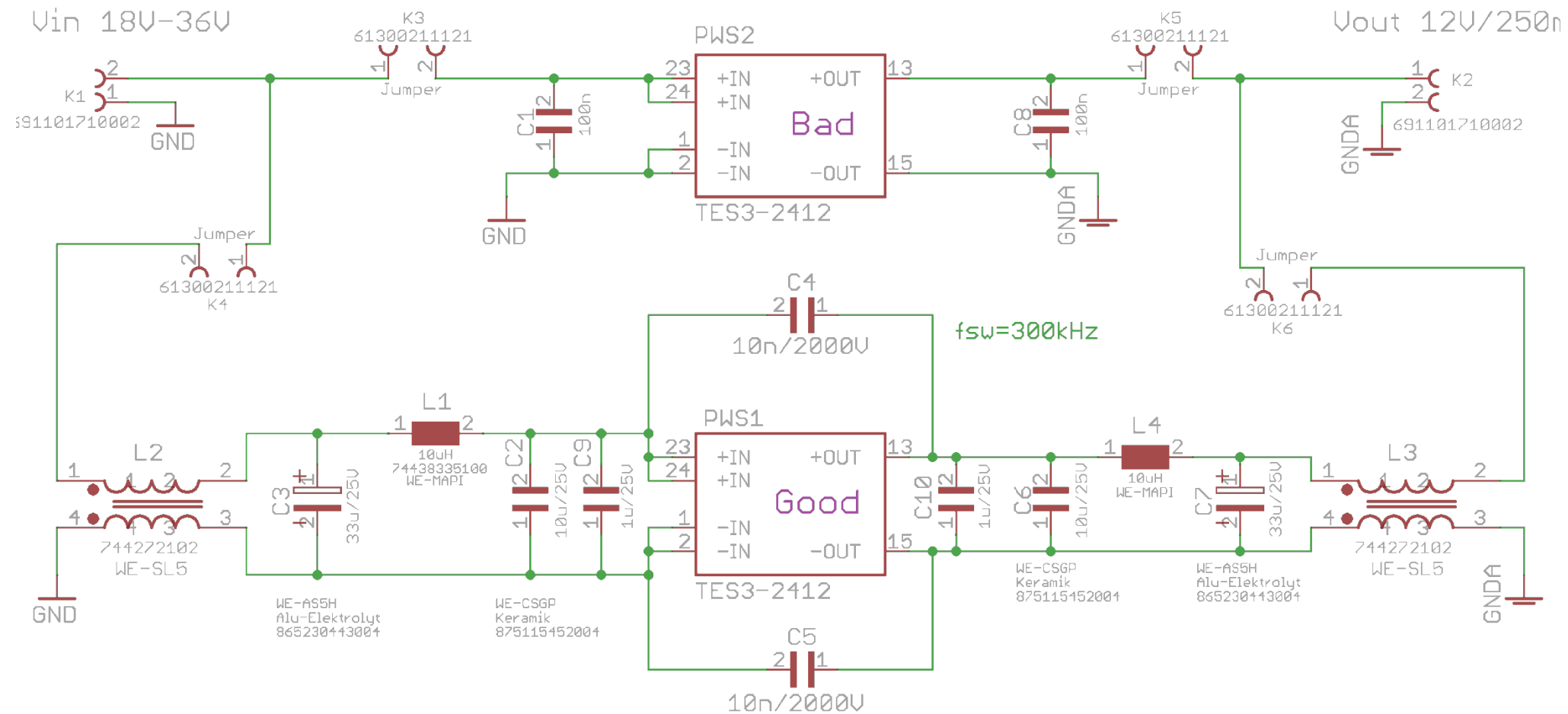
- U_{in} : 230VAC,
 - I_{out} : 4,16A,
 - Polarization: Horizontal
 - Norm: EN55022A
- U_{out} : 12VDC
 f_{sw} : 90kHz



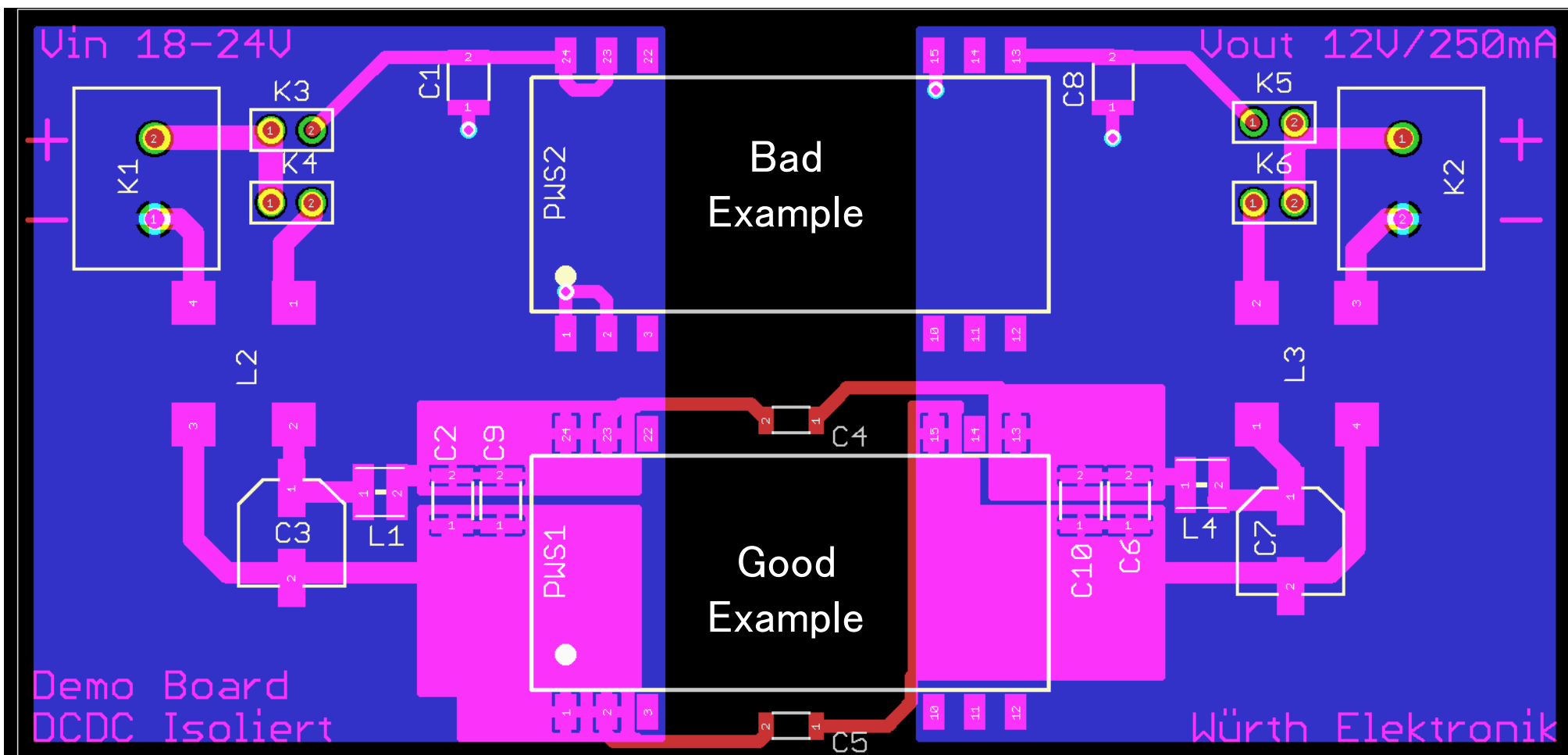


OTHER EMC SITUATION FOR A DC/DC CONVERTER

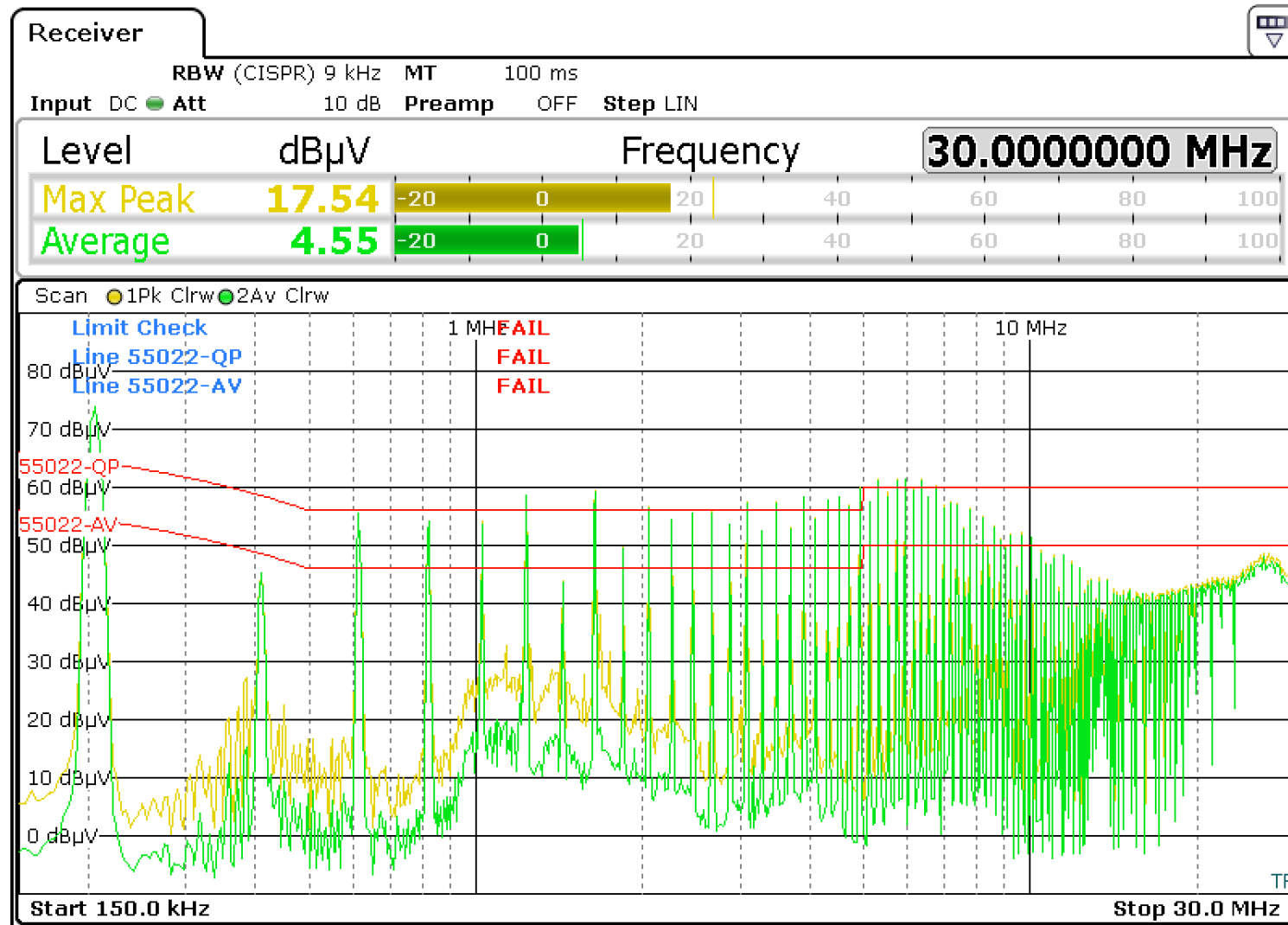
DC/DC Converter with galvanic separation



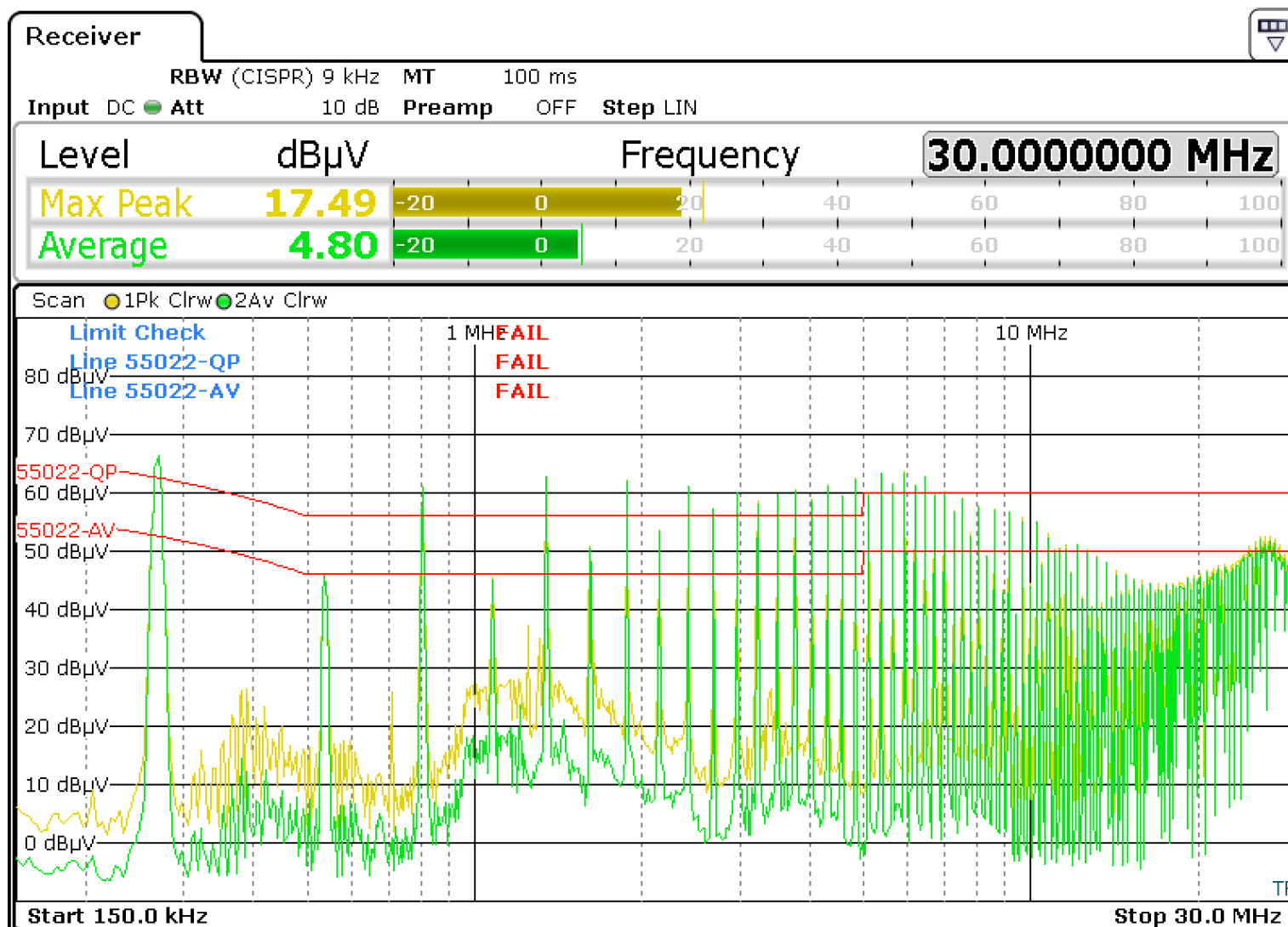
DC/DC Converter with galvanic separation



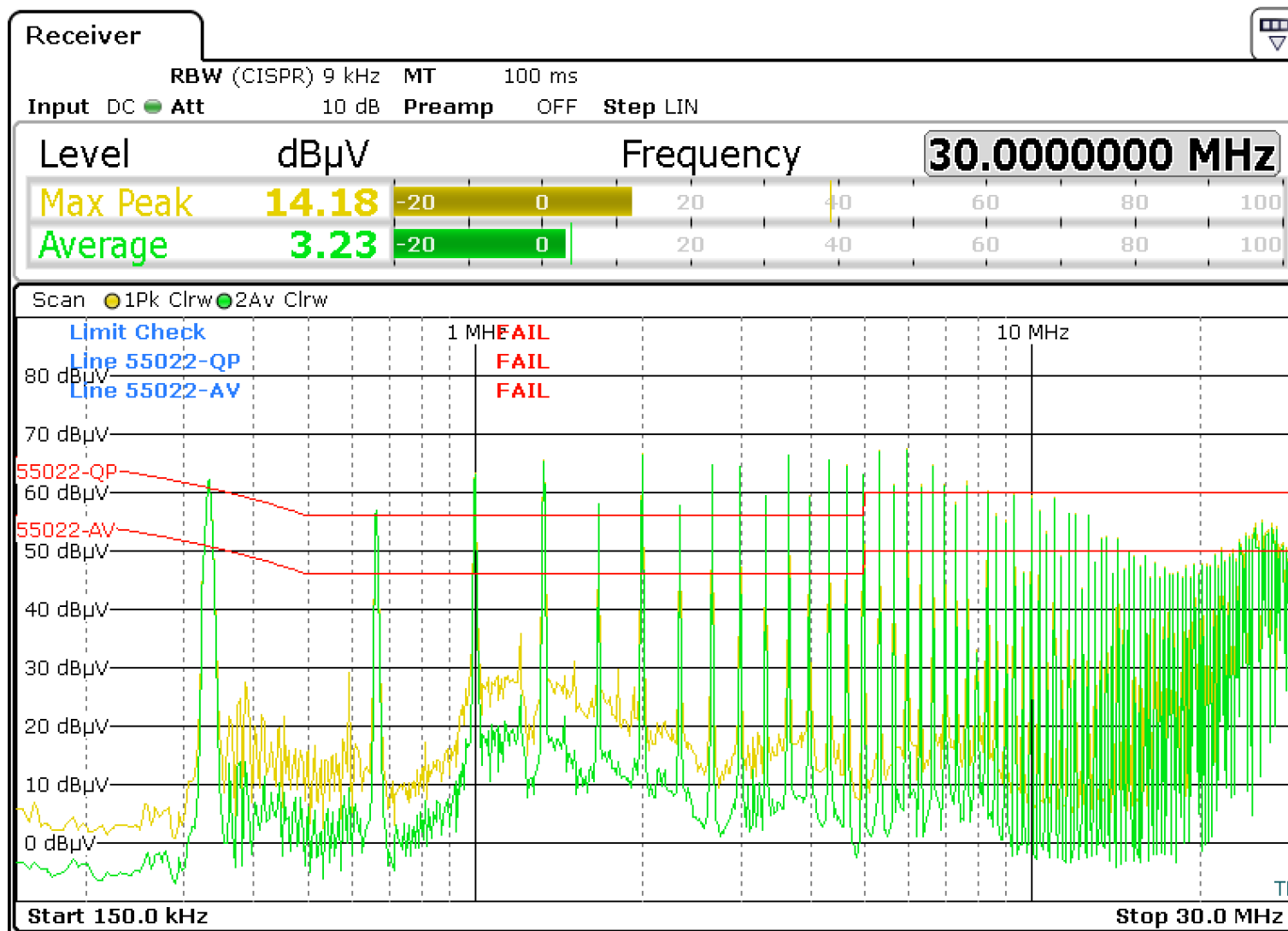
DC/DC Converter with galvanic separation 18V input no filter



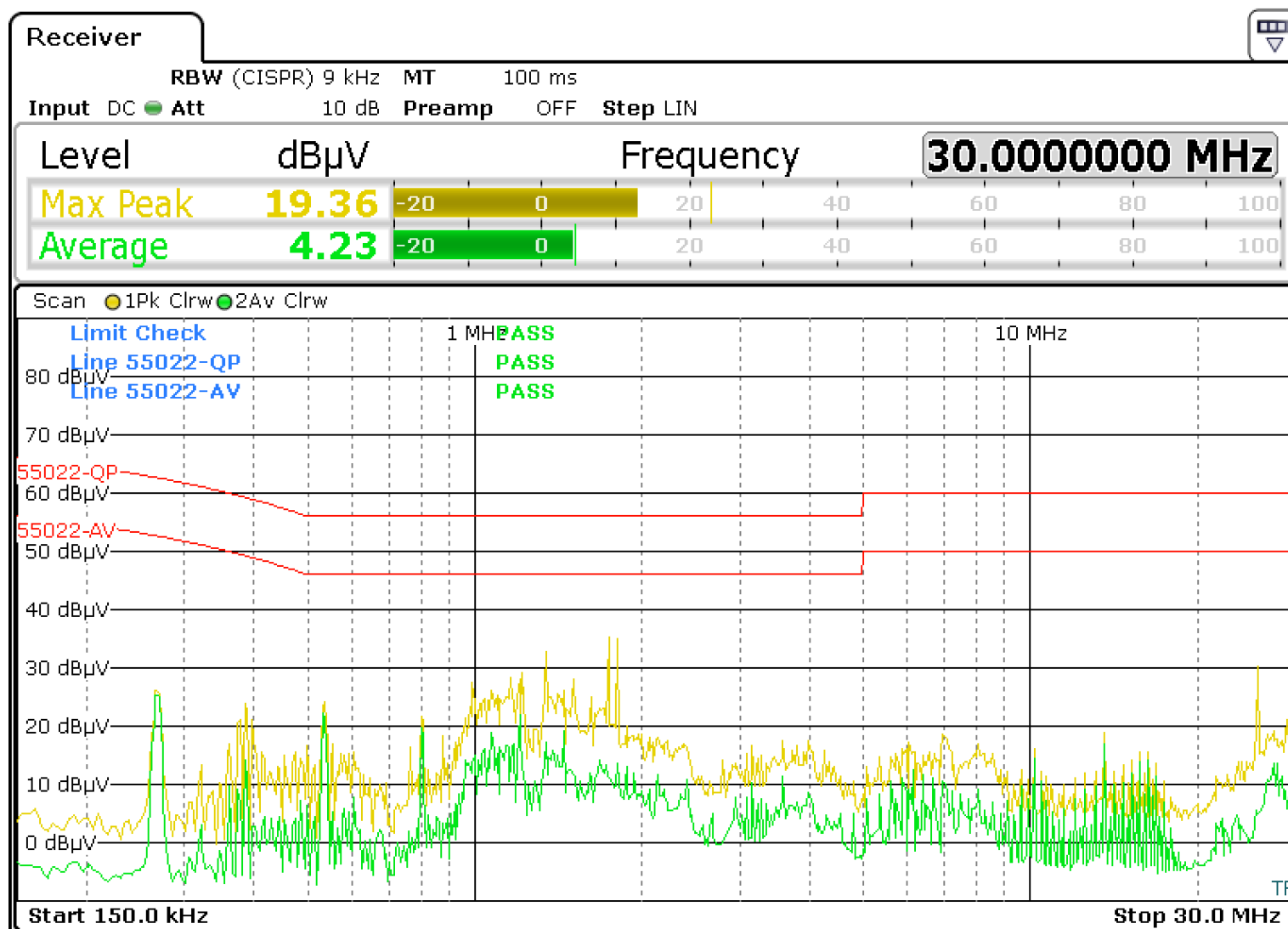
DC/DC Converter with galvanic separation 24V input no filter



DC/DC Converter with galvanic separation 32V input no filter



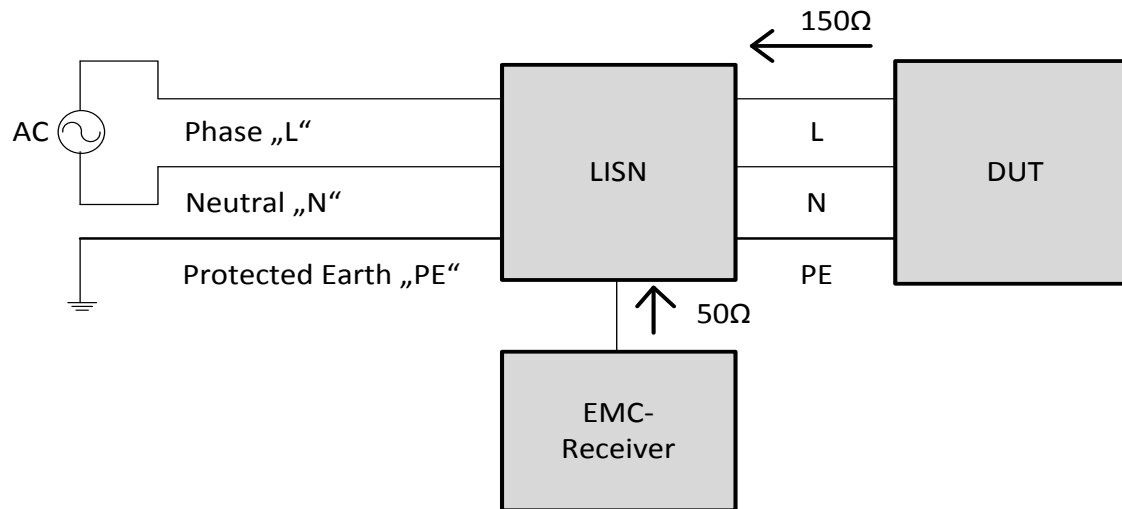
DC/DC Converter with galvanic separation 24V input with input filter





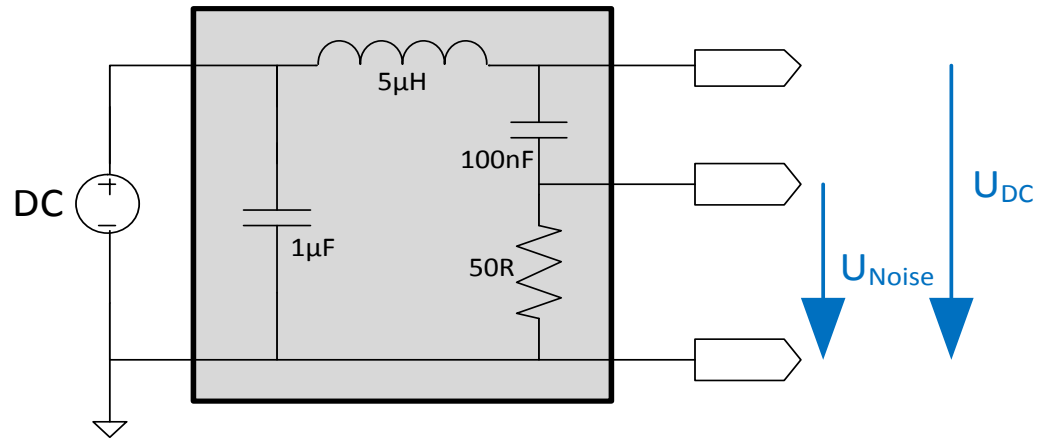
MEASUREMENT TECHNIQUES

Conducted Emission test setup



- **LISN: „Line Impedance Stabilization Network“**
 - **Create known impedance on power lines for DUT**
 - **Filter mains voltage and cut higher frequency**
 - **Transfer conducted emission noise to EMC-Receiver**
- **EMC-chamber is recommended but not required**

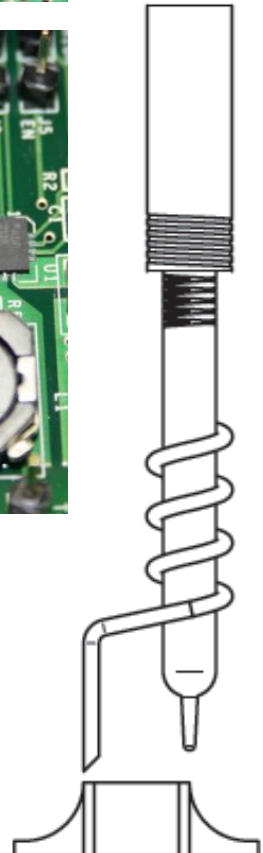
DC-LISN



- DC-LISN allow measurement of conducted emission at DC/DC converters
- Decouple the DC from the EMC receiver
- Creates 50Ω impedance for EMC-receiver
- Just differential noise measureable

Ripple-Measurement

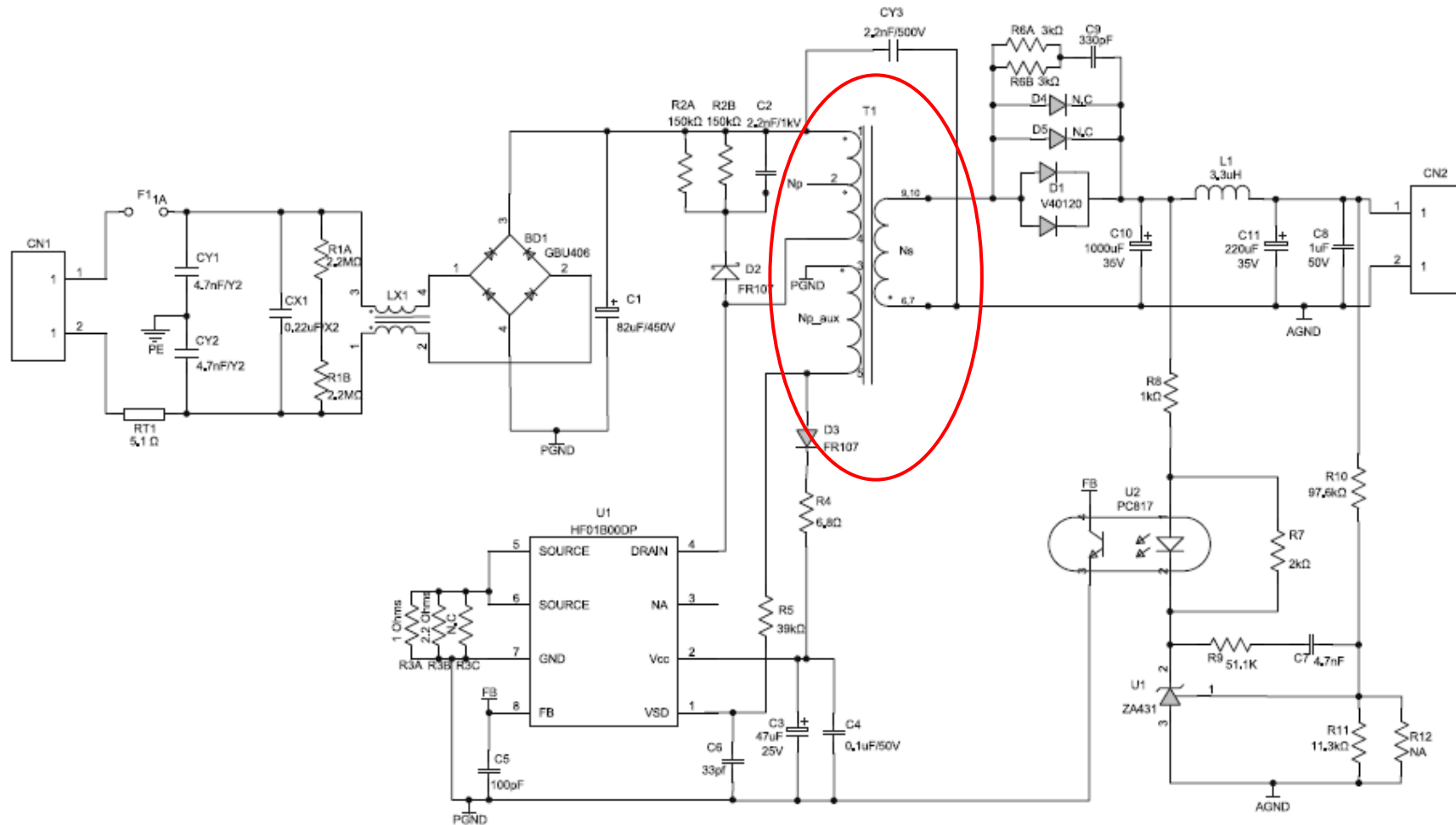
- for a clean external connection





OPTIMIZED TRANSFORMER

Schematic

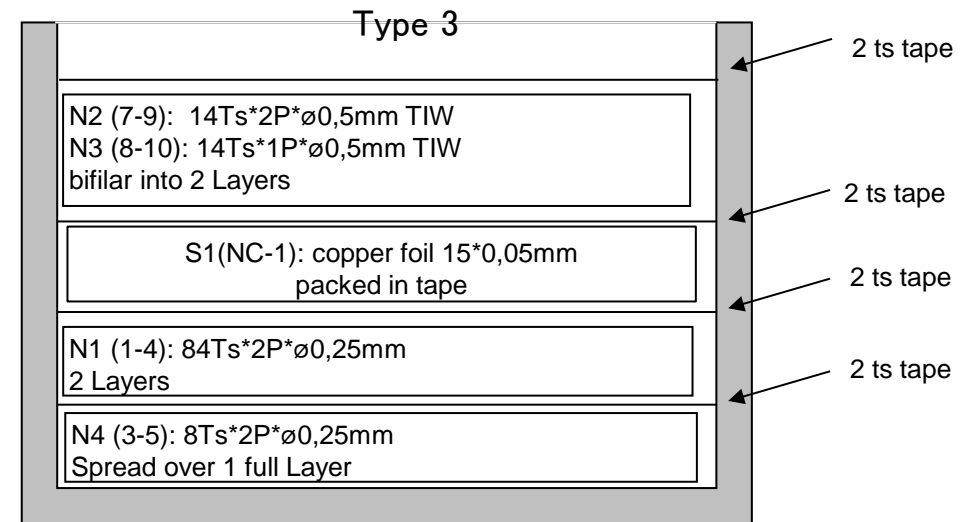
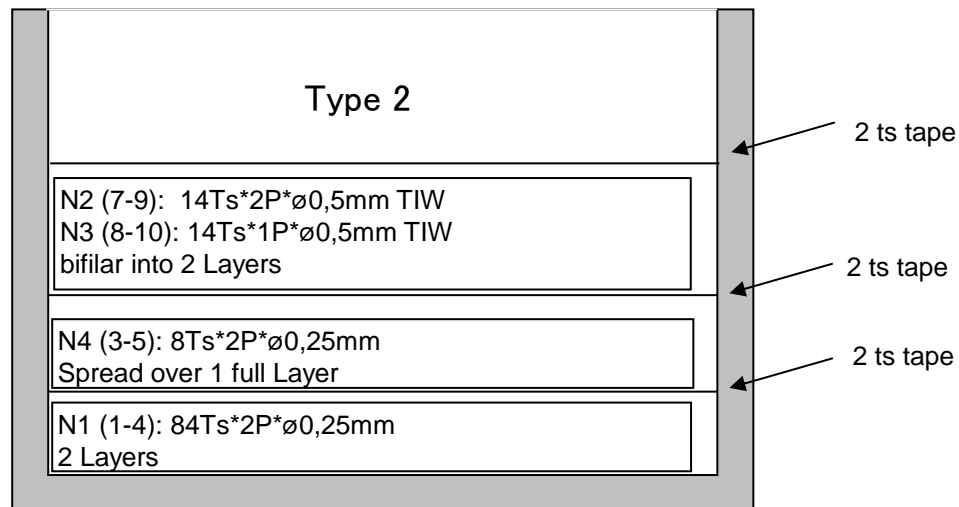
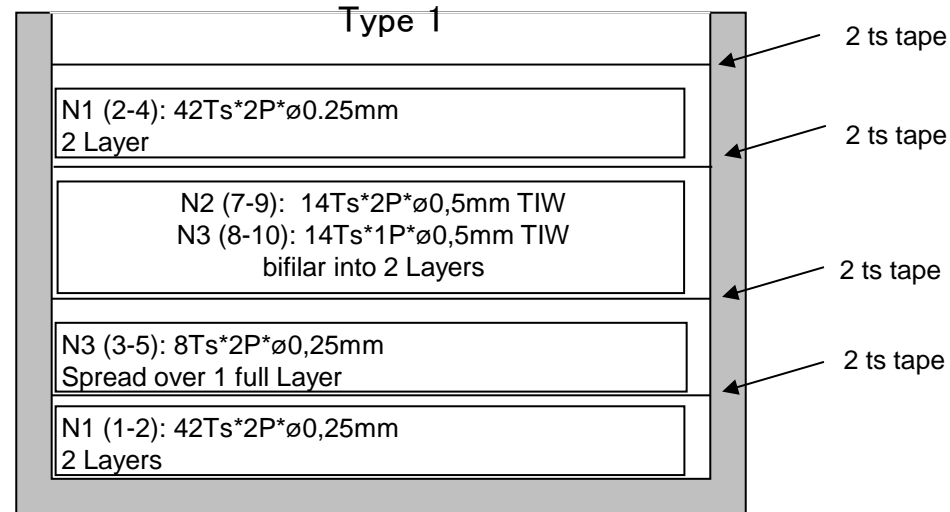


Quelle: MPS EVHF01B00DB-00A

Compare Transformers

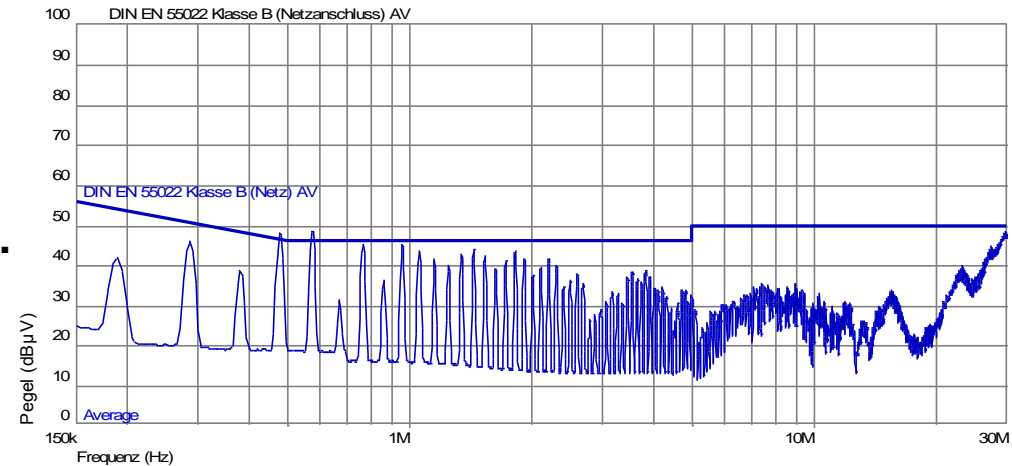
	Type 1	Type 2	Type 3
Inductance	891 μH	907 μH	933 μH
Leakage Inductance	13,2 μH	20,7 μH	26,6 μH
Winding Capacity	53,1 pF	29,0 pF	64,9 pF

Transformer Construction

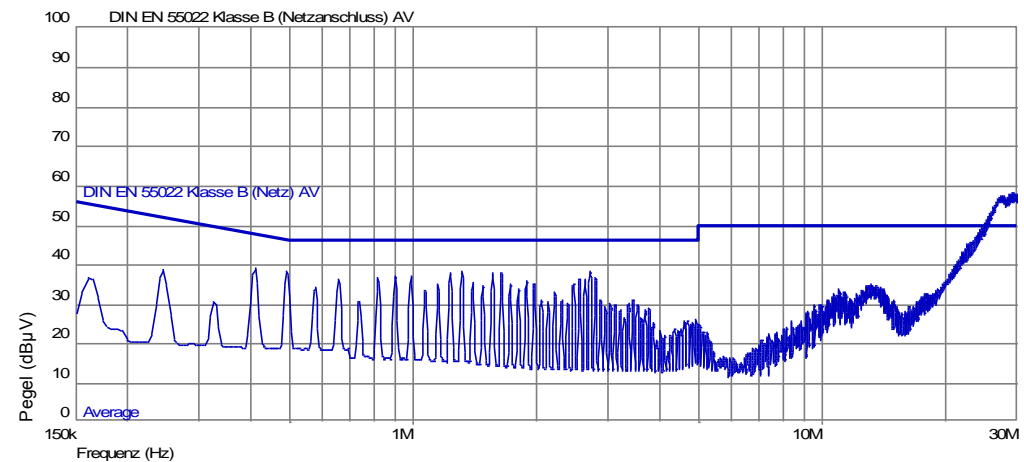


Conducted Emissions with diff Transformers

Type 1:
Transformer with lowest Leakage Ind.

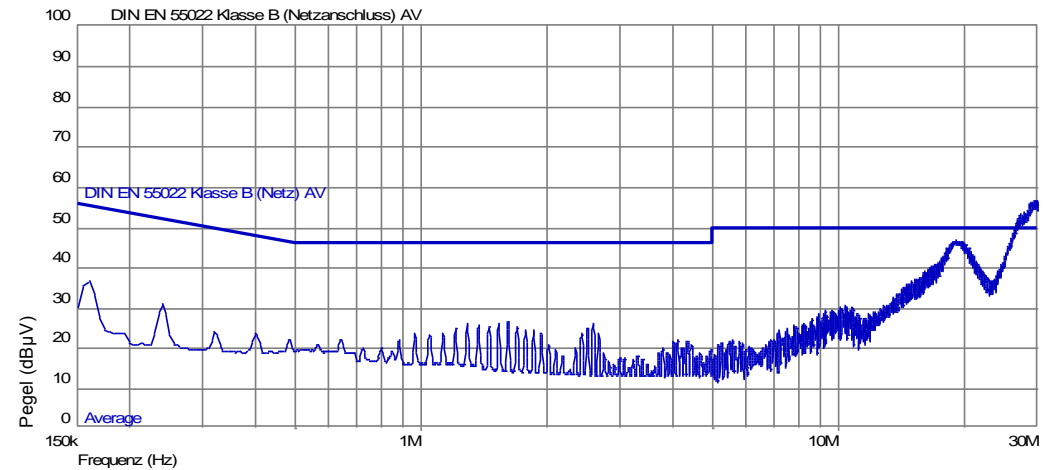


Type 2:
Transformer optimized Lowest Cost

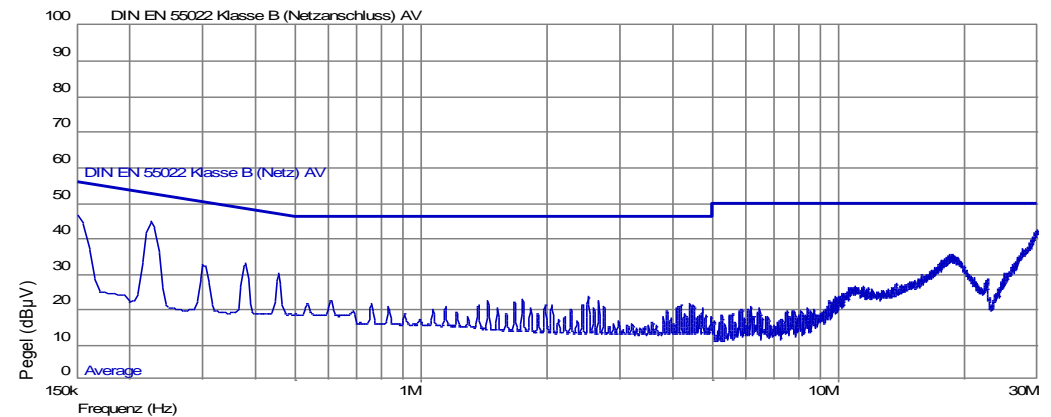


Conducted Emissions with diff Transformers

Type 3: Transformer with Shielding Winding



Transformer Type 3 With additional CMC and Ycap

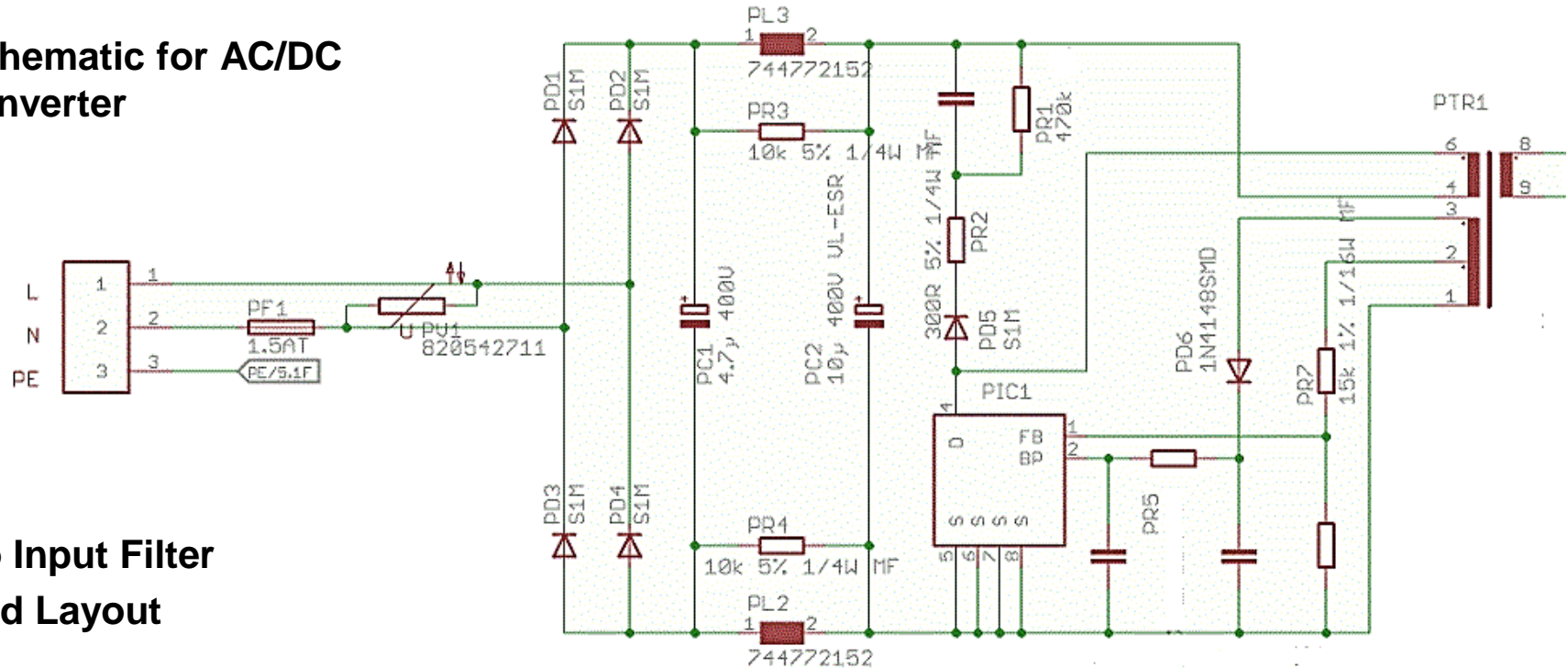




EXAMPLES FOR BAD DESIGN

Example for Bad Design

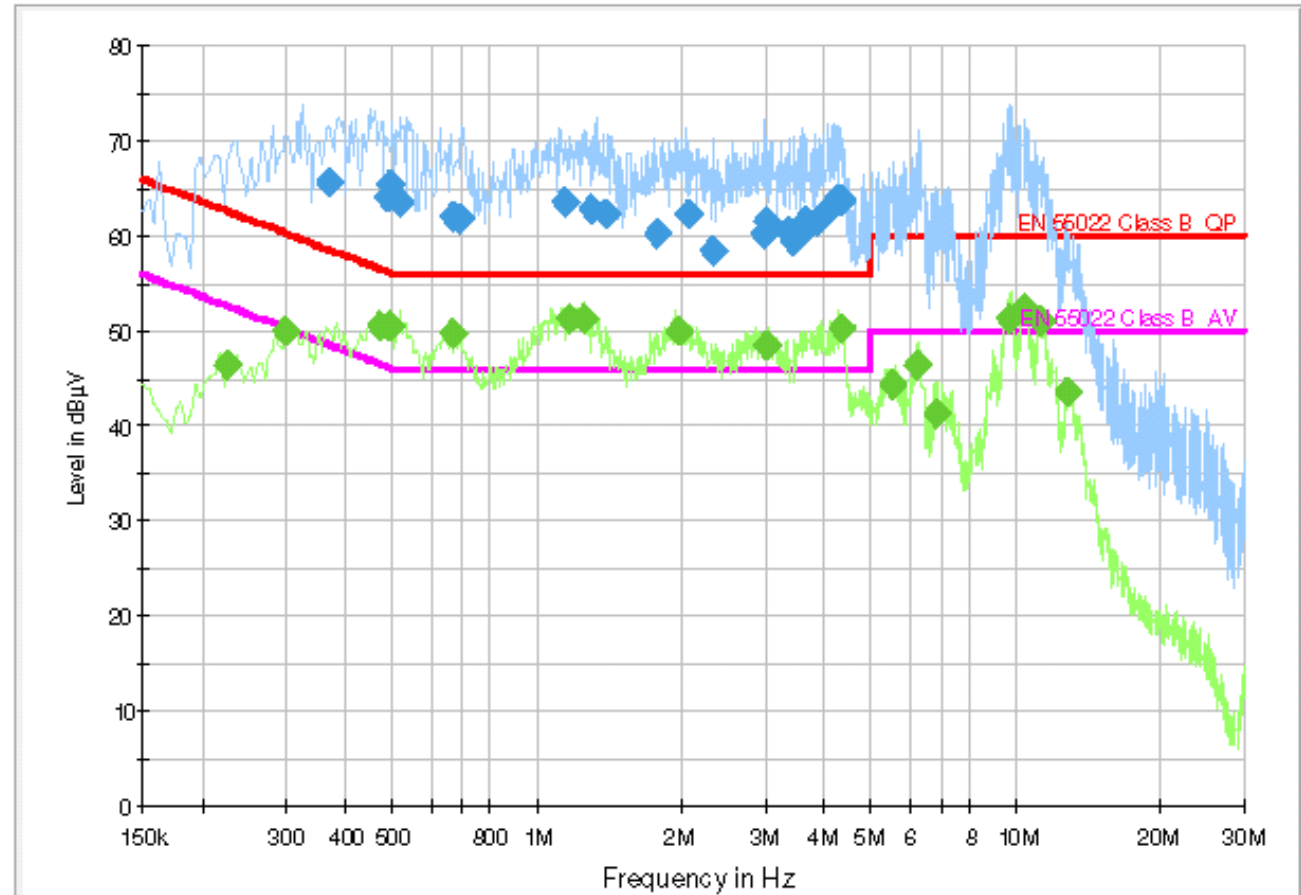
- **Schematic for AC/DC converter**



- **No Input Filter**
- **Bad Layout**

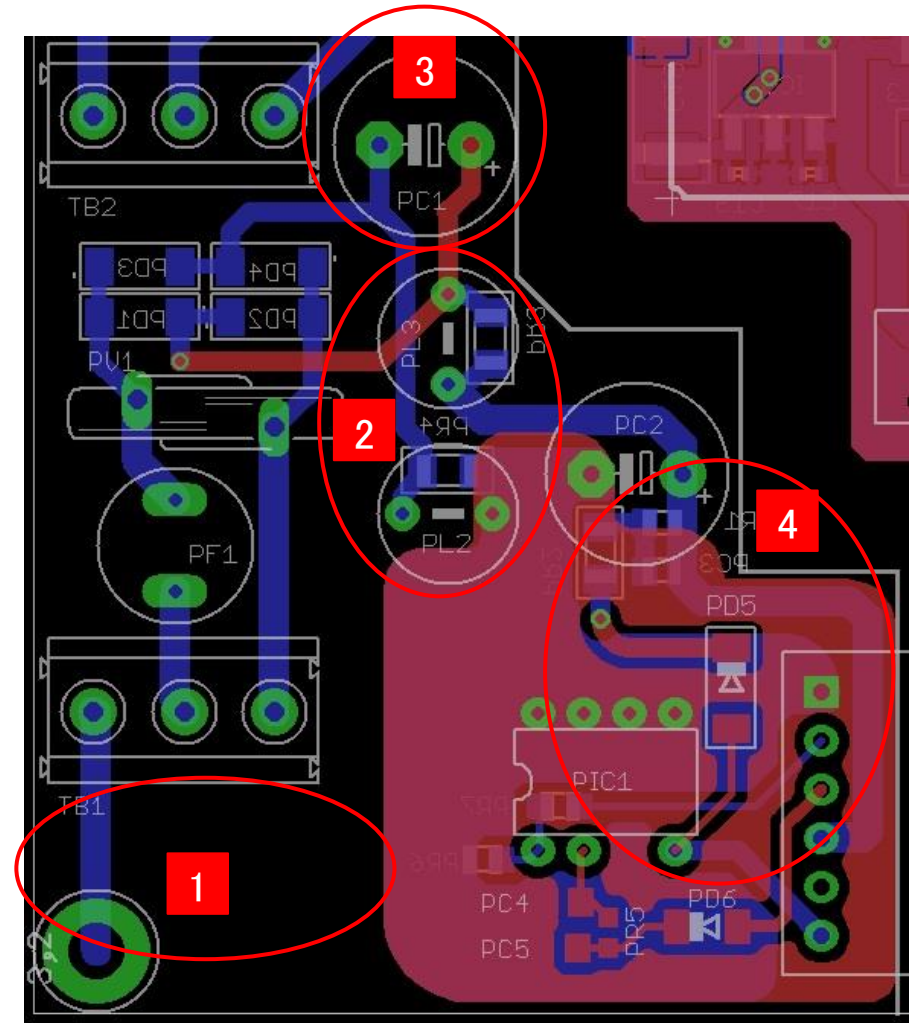
Example for Bad Design

- High Emissions for Conducted
- QP & AV limits exceed



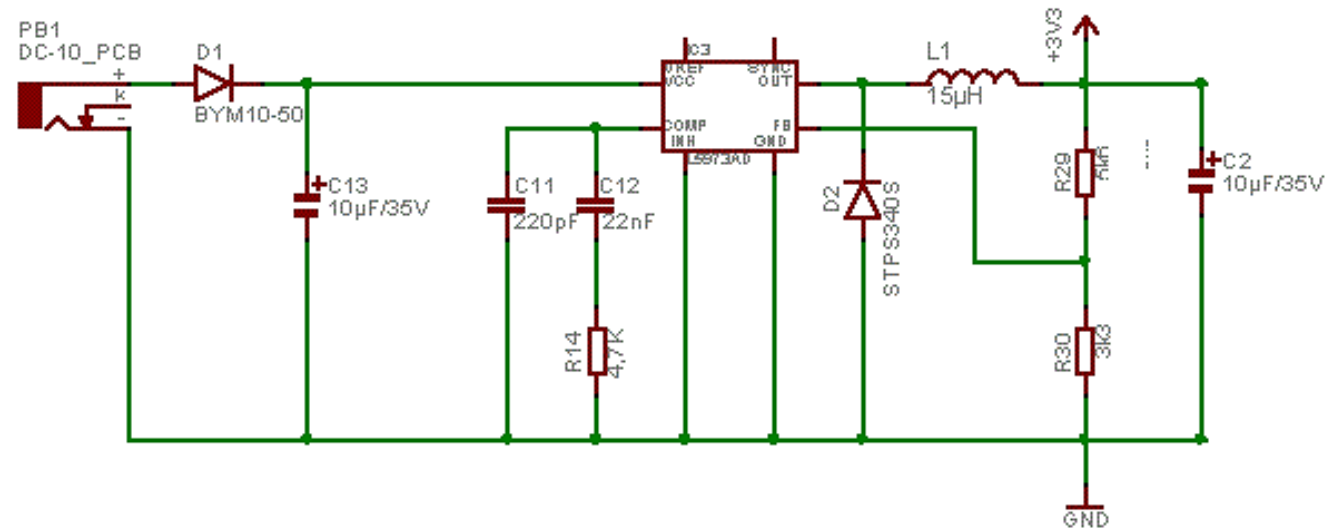
Example for Bad Design

- No Input Filter
- Simple Pi Filter → Layout mistake!
- Wrong positioned Filter
- Simple 2 Layer
- Bad routing



Example for Bad Design

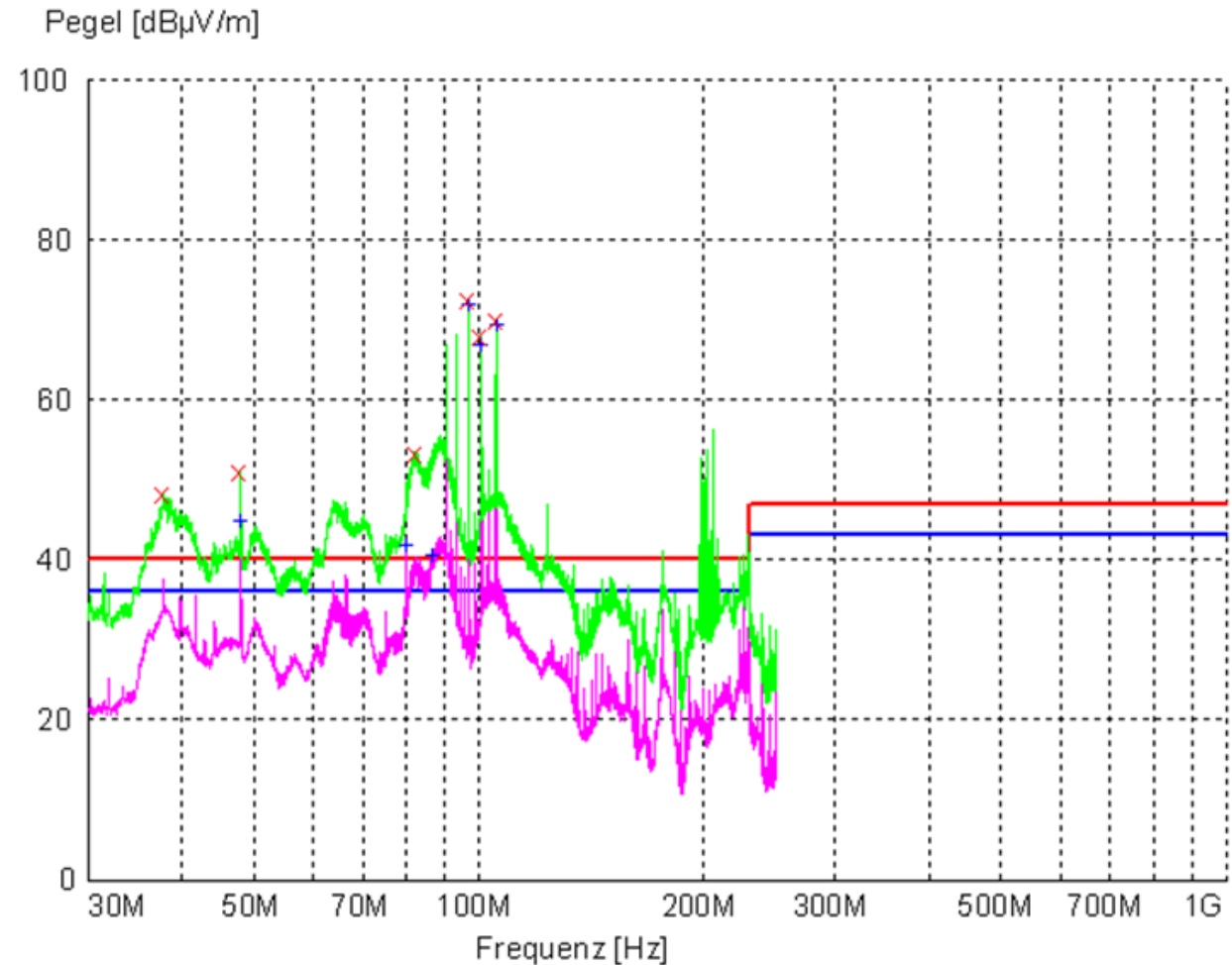
- Un insulated DC/DC converter



- No input filter
- Bad Layout

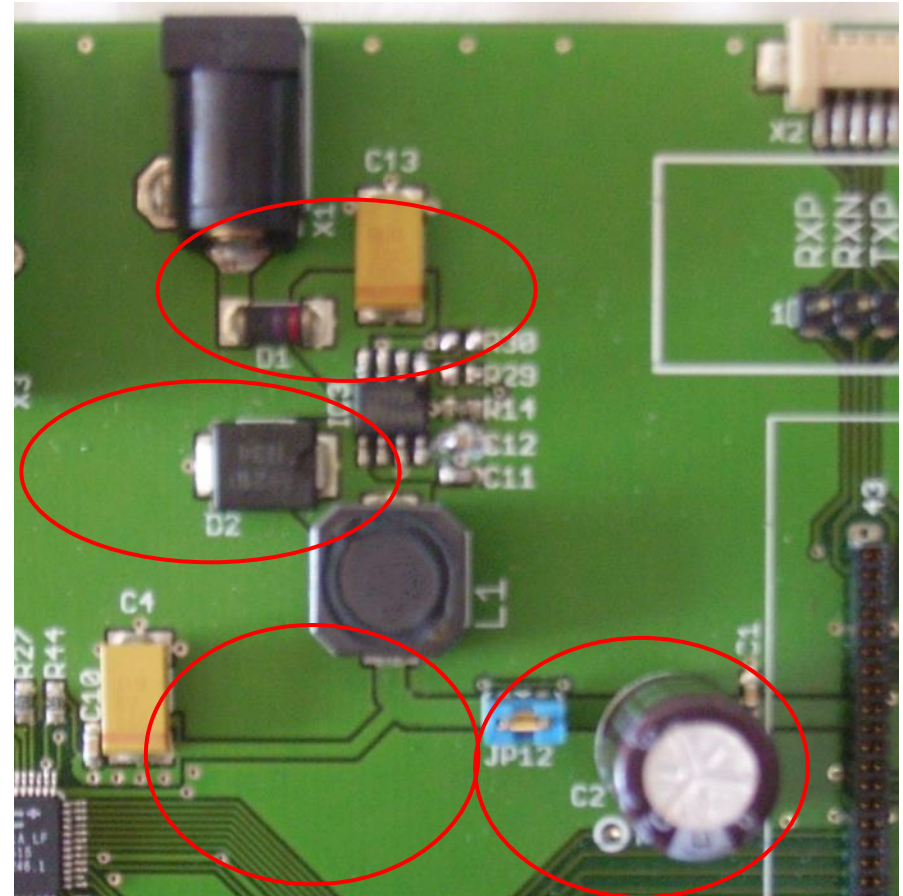
Example for Bad Design

- High emissions for radiated
- Limits over shooted



Example for Bad Design

- No input filter
- Simple 2 layer
- Wrong position for output capacity
- Bad Ground routing



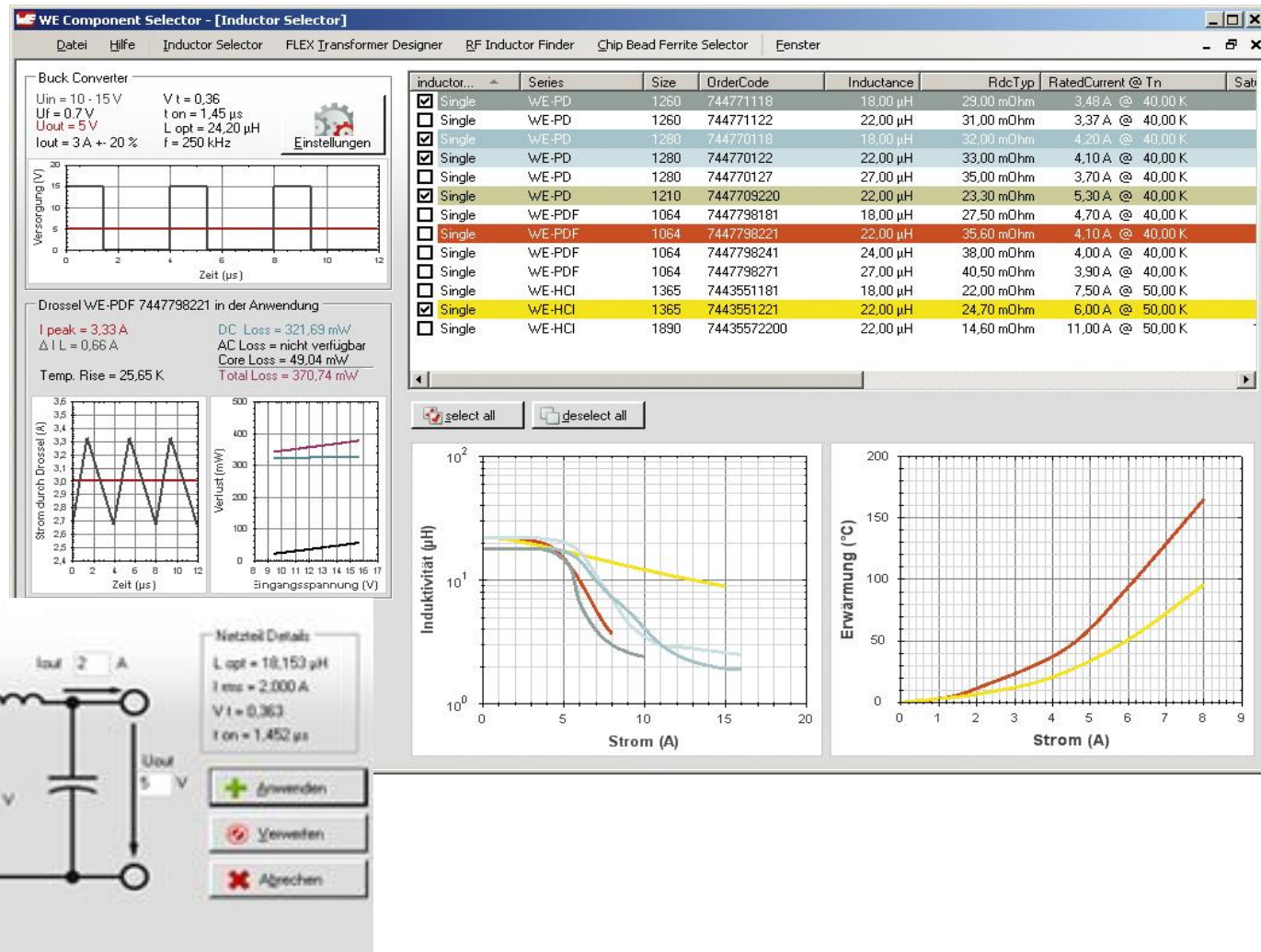


DESIGN TOOLS

WE Component Selector



WE Component Selector



REDEXPERT


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[Start](#) [Würth Elektronik Group](#) [Sign in](#) [English](#)

more than you expect

Inductors **REDEXPERT®**

100 items

Filters: Type = Single ☒ $I_R \geq 2.00 \text{ A}$ ☒ $I_{sat} \geq 2.40 \text{ A}$ ☒ $5.28 \mu\text{H} \leq L \leq 9.80 \mu\text{H}$ ☒

✓	Series	Order Code	Spec	Type	L	R _{DC,typ}	I _R	I _{sat}	Size	Length	Width	Ht
✓	WE-MAPI	74438356056		Single	5.60 μH	68.0 m Ω	2.80 A	4.60 A	4020	4.1 mm	4.1 mm	
✓	WE-TPC	744071056		Single	5.60 μH	20.0 m Ω	4.00 A	4.00 A	8043	8.0 mm	8.0 mm	
✓	WE-TPC	7440650068		Single	6.80 μH	25.0 m Ω	4.20 A	3.60 A	1028	10 mm	10 mm	
✓	WE-TPC	7440650082		Single	8.20 μH	28.5 m Ω	3.80 A	2.80 A	1028	10 mm	10 mm	
✓	WE-TPC	7440660062		Single	6.20 μH	16.5 m Ω	4.30 A	4.50 A	1038	10 mm	10 mm	
✓	WE-SPC	74408943068		Single	6.80 μH	51.0 m Ω	2.00 A	2.70 A	4838	4.8 mm	4.8 mm	

74438356056

WE-MAPI · Single
5.60 μH · 68.0 m Ω
2.80 A · 4.60 A

744071056

WE-TPC · Single
5.60 μH · 20.0 m Ω
4.00 A · 4.00 A

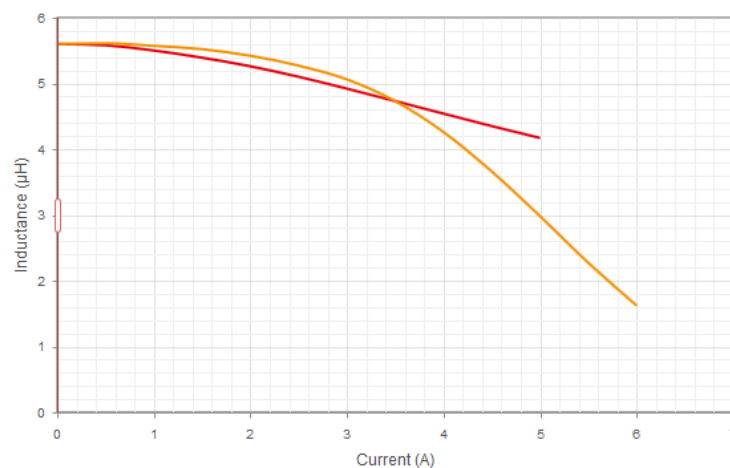
Please, register to add more parts

Share

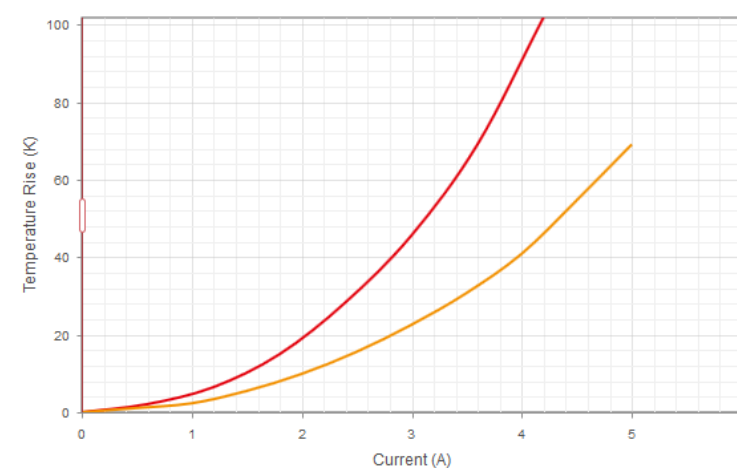
Free Samples

Tidy Up

Inductance / Current



Temperature Rise / Current



Simulation – WEBENCH



- http://www.we-online.de/web/de/electronic_components/toolbox_pbs/webench.php

My Designs/Projects English | 日本語 | 简体中文 | 繁體中文 | 한국어 | Русский Язык | Português | Deutsch | Welcome

WE WÜRTH ELEKTRONIK

New Solutions Visualizer Assistant

RECOMMENDED PARTS

Try WEBENCH Visualizer
Compare all designs in seconds

Size = Total BOM Cost

Footprint (mm²)

Efficiency

WEBENCH® Visualizer

WEBENCH® Optimizer

Lowest BOM Cost

Smallest Footprint

Highest Efficiency

WEBENCH® Visualizer

Switching Regulator
LM3102

Open Design

Design Note	High Efficiency
Topology	Buck
Footprint (mm²)	340
Efficiency (%)	87%
Frequency (kHz)	315
BOM Cost	\$4.59

Switching Regulator
LM2576

Open Design

Design Note	Fast Transient R...
Topology	Buck
Footprint (mm²)	278
Efficiency (%)	85%
Frequency (kHz)	361
BOM Cost	\$5.69

Switching Regulator
LM22670-ADJ

Open Design

Design Note	Adjustable for V...
Topology	Buck
Footprint (mm²)	348
Efficiency (%)	83%
Frequency (kHz)	388
BOM Cost	\$4.51

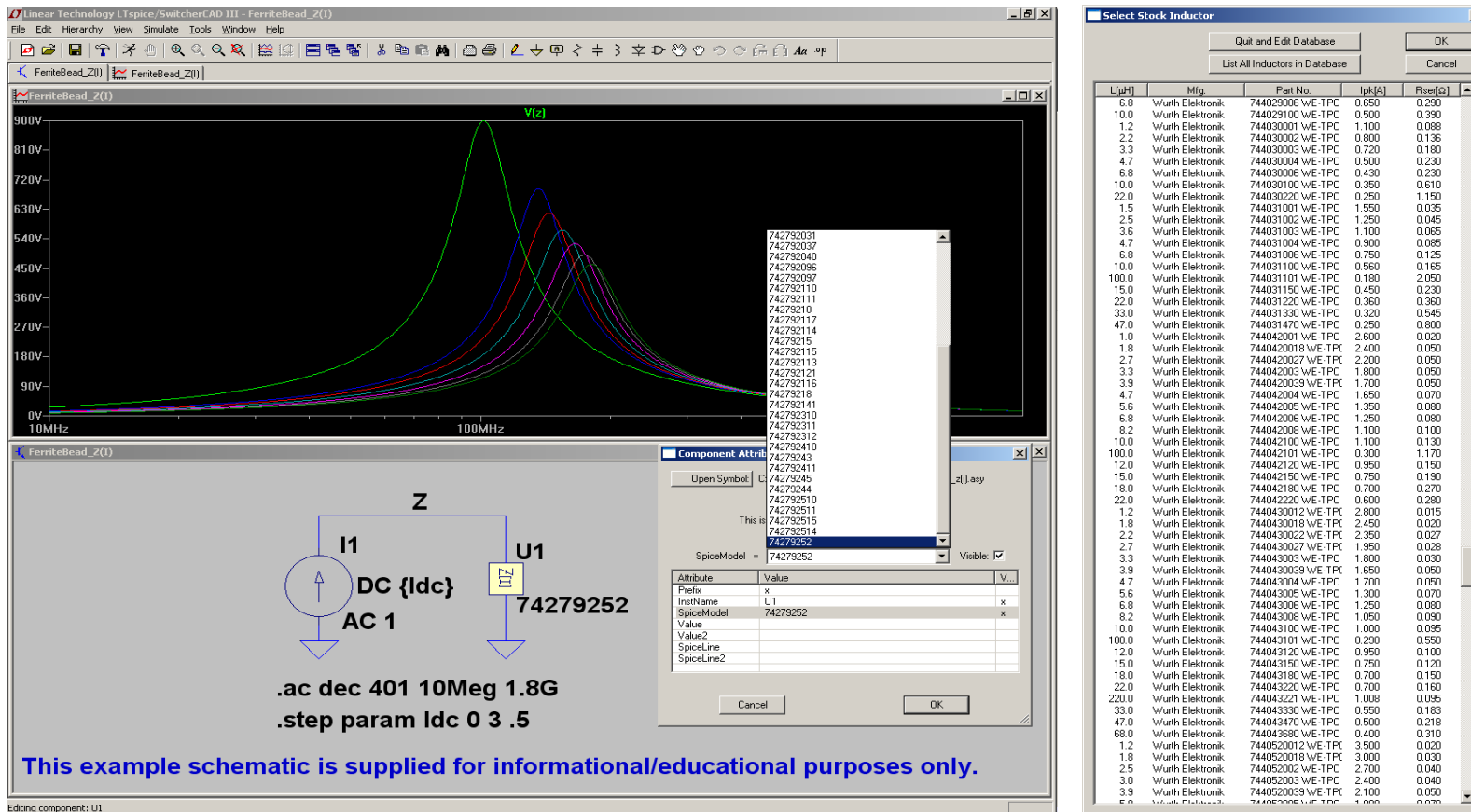
Switcher Solutions

Switcher Solutions: (146 found) ☐ Show All Columns ☐ Show Alternate Topologies ☐ Show Only Modules

Part	Create	WEBENCH® Tools	Topology	Efficiency (%)	Footprint (mm²)	Frequency (kHz)	Vout p-p (mV)	Cross Freq (kHz)	Phase Margin (deg)	BOM Cost	BOM Count	Iout Max (A)	Design Considerations
LM3151-3.3	Open Design		Buck	91%	524	245	6.01	NA	NA	\$5.15	10	12.00	SIMPLE SWITCHER(r) Controller
LM43602	Open Design		Buck	89%	326	350	1.90	14	75	\$4.17	13	2.00	SIMPLE SWITCHER Buck Regulator
LM3150	Open Design		Buck	93%	443	255	5.52	NA	NA	\$5.89	15	15.00	SIMPLE SWITCHER(r) Controller
TPS54339	Open Design		Buck	88%	285	646	4.69	NA	NA	\$2.83	12	3.00	Wide Vin Buck Converter with EcoMode
TPS54239E	Open Design		Buck	88%	285	646	4.72	NA	NA	\$2.73	12	2.00	Wide Vin Buck Converter with EcoMode
TPS54335A	Open Design		Buck	88%	340	270	1.66	16	59	\$3.43	13	3.00	28V, 3A, Low Iq, Synchronous, monolithic buck converter with Eco-mode
LM43603	Open Design		Buck	89%	230	350	3.34	12	74	\$5.53	13	3.00	SIMPLE SWITCHER Buck Regulator

Simulation – LTSpice IV

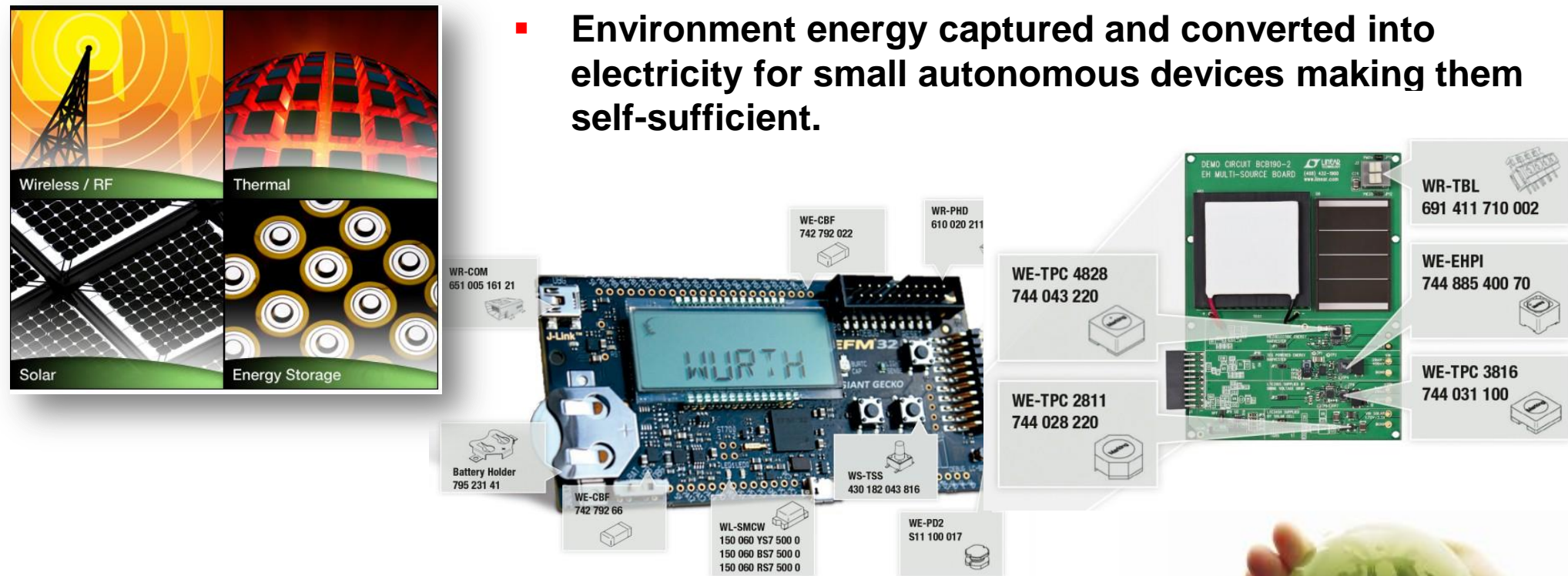
- <http://www.linear.com/designtools/software/#LTspice>



Editing component: U1

Energy Harvesting to Go kit

- Environment energy captured and converted into electricity for small autonomous devices making them self-sufficient.

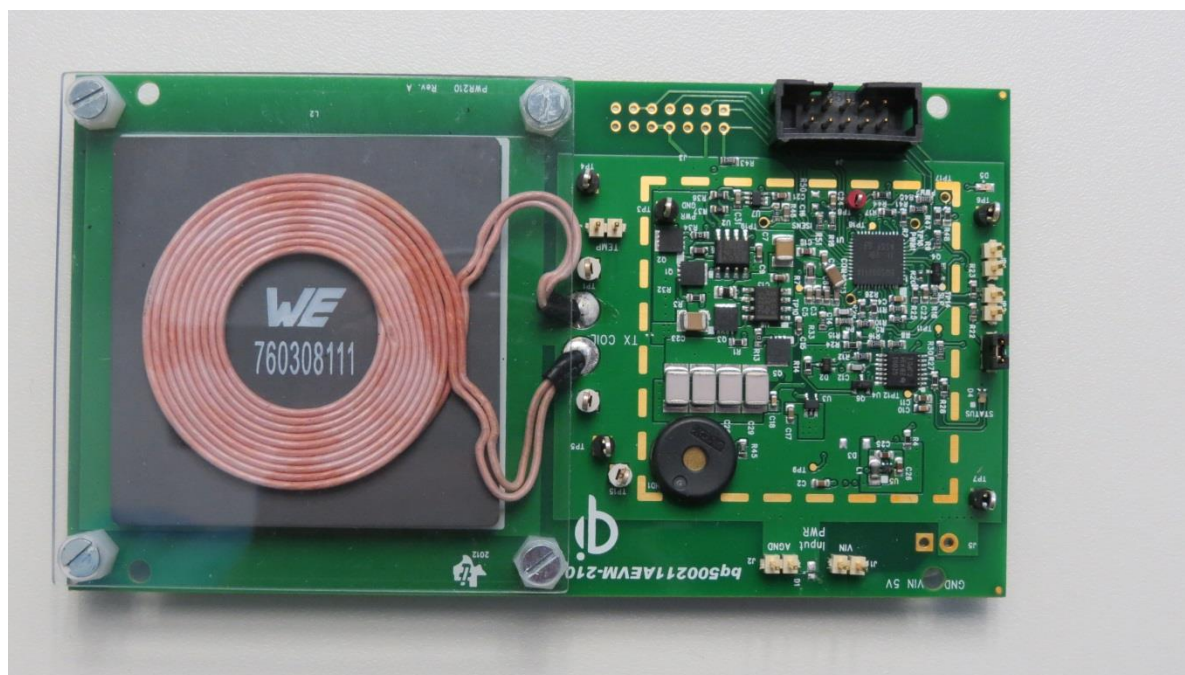


- ❖ Thermo Electric Generator (heat)
- ❖ Piezo Electric (vibration/strain)
 - ❖ Photovoltaic (light)
 - ❖ Galvanic (chemical)
 - ❖ Induction (motion)

Energy
Management &
Storage

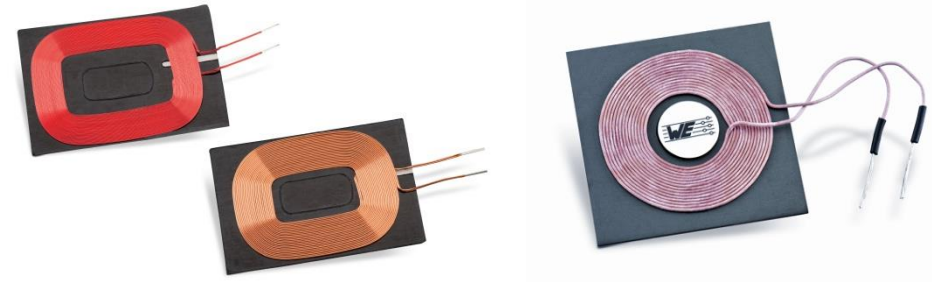
Regulated Voltage

WE WPCC Demo Kit



Wireless Power Coils WE-WPCC- Tx/Rx coils

- Fully compliant to WPC Qi standard
- Efficiency up to 85%
- Supreme shielding characteristics for low leakage inductance
- Outstanding performance due to usage of Litz wire:
 - ✓ lowest R_{DC}
 - ✓ highest Q values



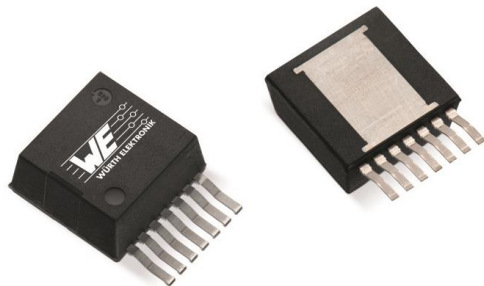
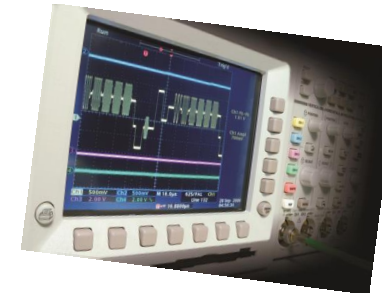
Magl³C Power Module

DC/DC Step Down Converter with integrated inductor

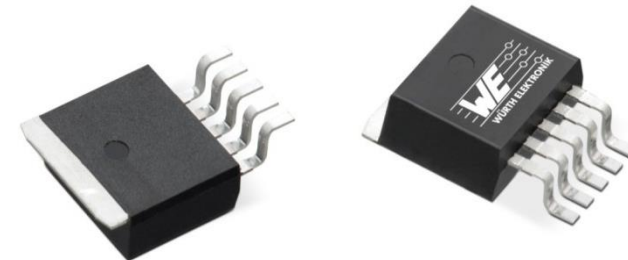


- **Branch:**
 - Industrial
 - Communication
 - Audio/Video equipment
 - Test & Measurement equipment
 - Medical

- **Application area:**
 - Voltage supply
 - Multi-Voltage Systems
 - Software developers with no hardware knowledge



5 types with variable output voltage



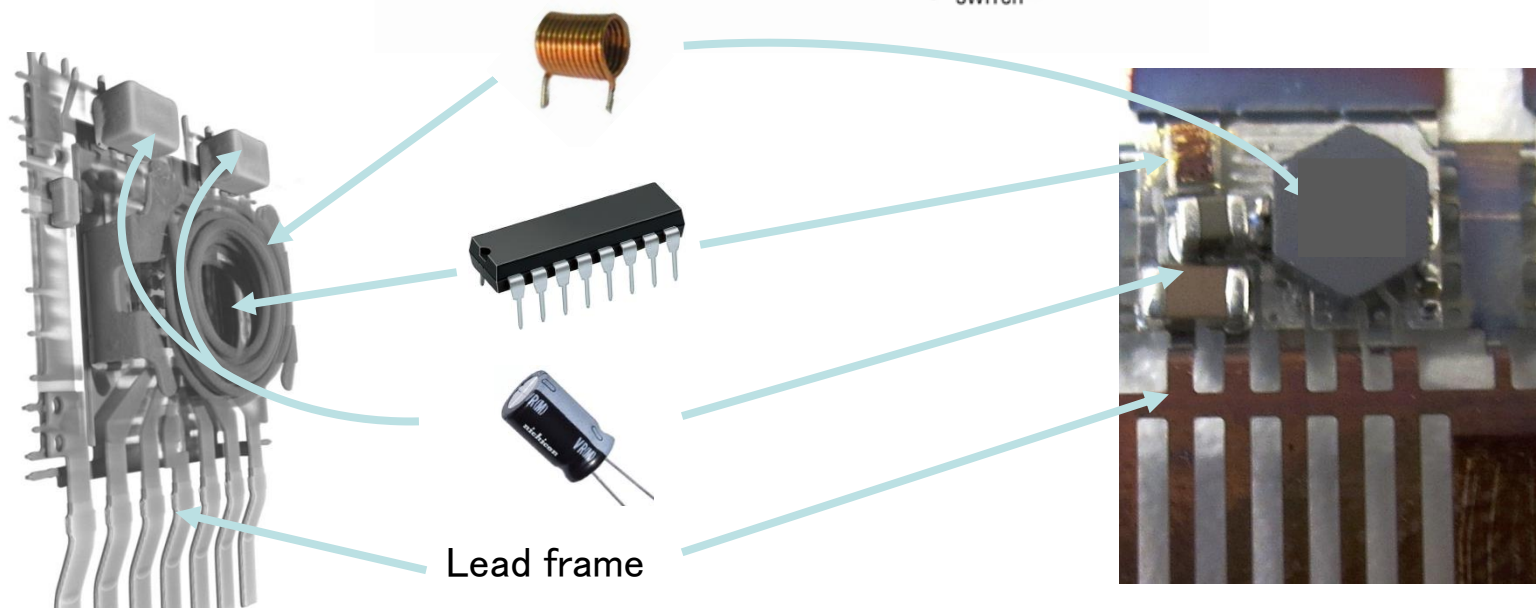
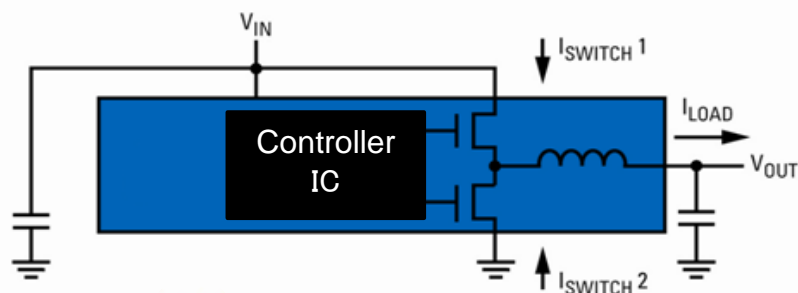
2 types with variable output voltage

MagI³C Power Module

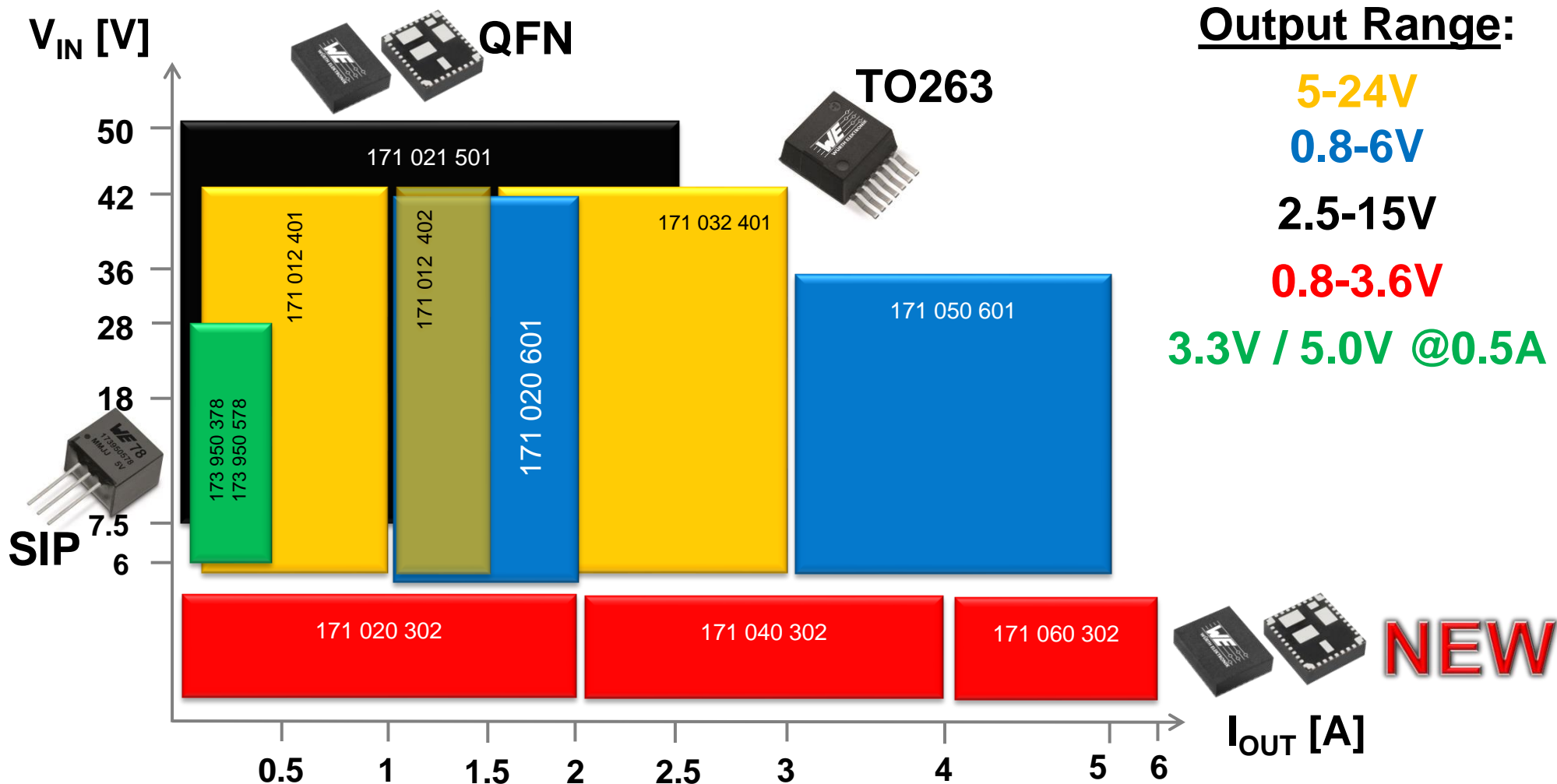
DC/DC Step Down Converter with integrated inductor



Power Module Concept



Magl³C Step-Down Regulator Power Module Family



Trilogy of Magnetics



- 1. LTspice Book
 - How to use and build spice models
- 2. Trilogy of Magnetics
 - Design Guide for EMI Filter Design, SMPS & RF Circuits
- 3. Trilogy of Connectors
 - Basic Principles and Connector Design Explanations
- 4. ABC of Power Modules
 - Functionality, Structure and Handling of a Power Module
- 5. ABC of Capacitors
 - Basic principles, characteristics and capacitor types

!!! Questions ???

- Do you **still think** ...

... that a DC/DC converter “**does not** generate **Conducted Emission**” ???

... that the EMC of a DC/DC converter “**isn’t** affected by the **PCB layout**” ???

... that an “**oscilloscope can’t** help you to carry out any **EMC tests**” ???



If you still have questions?

Just call us: we try to help you

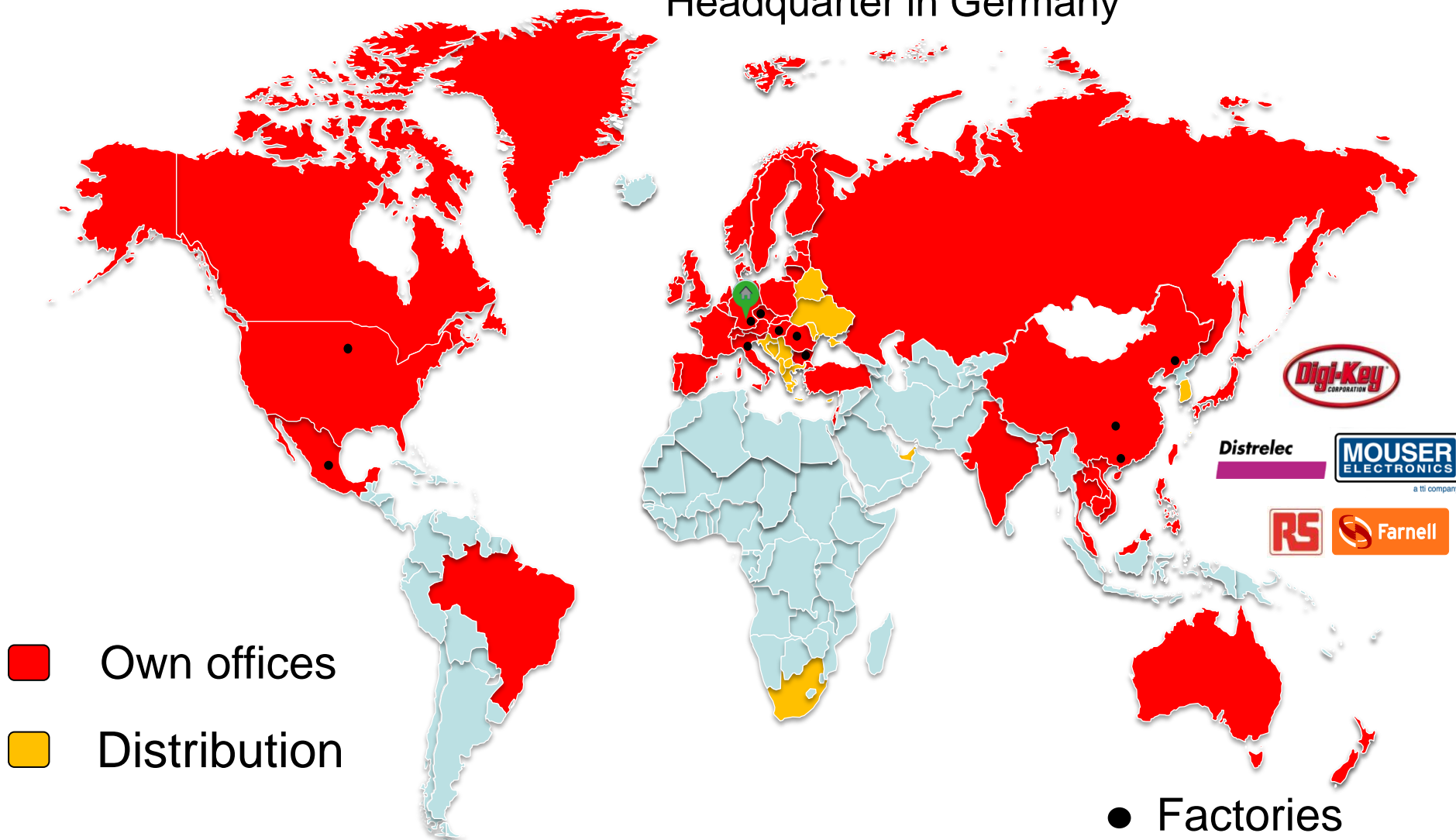
Don't give up !!!



Globally available. Locally present!



Headquarter in Germany



Würth Elektronik eiSos GmbH & Co.KG



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